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DK4

Handel-C Language Reference Manual

For DK version 4

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Contents

1 INTRODUCTION 11
1.1 REFERENCES 11

2 GETTING STARTED WITH HANDEL-C 12
2.1 BASIC CONCEPTS 12
2.1.1 Handel-C programs 12
2.1.2 Parallel programs 12
2.1.3 Channel communication 13
2.1.4 Scope and variable sharing 15

3 LANGUAGE BASICS 16
3.1 PROGRAM STRUCTURE 16
3.2 COMMENTS 17
3.3 STATEMENT SUMMARY 17
3.4 OPERATOR SUMMARY 19
3.5 TYPE SUMMARY 21
3.6 COMPARISON OF HANDEL-C AND ANSI-C 22
3.6.1 Handel-C v C: types and type operators 23
3.6.2 Handel-C v C: floating-point variables 23
3.6.3 Handel-C v C: variable widths and casting 24
3.6.4 Handel-C v C: side effects 25
3.6.5 Handel-C v C: functions 26
3.6.6 Handel-C v C: loop statements 27
3.6.7 Handel-C v C: unions 28
3.6.8 Handel-C v C: data input and output 29
3.6.9 Handel-C v C: memory allocation 29
3.6.10 Handel-C v C: standard library 29
3.6.11 C and Handel-C types and objects 29
3.6.12 Expressions in C and Handel-C 30
3.6.13 Statements in C and Handel-C 32
3.7 HANDEL-C CONSTRUCTS NOT FOUND IN ANSI-C 32

4 DECLARATIONS 36
4.1 INTRODUCTION TO TYPES 36
4.1.1 Handel-C values and widths 36
4.1.2 String constants 37
4.1.3 Constants 37
4.2 LOGIC TYPES 38
4.2.1 int 38
4.2.2 Signed | unsigned syntax 39
4.2.3 Supported types for porting 39
4.2.4 Inferring widths 40

4.2.5 Arrays 41
4.2.6 Array indices 43
4.2.7 struct 43
4.2.8 enum 44
4.2.9 Bit fields 45
4.3 POINTERS 46
4.3.1 Pointers and addresses 48
4.3.2 Pointers to functions 48
4.3.3 Pointers to interfaces 49
4.3.4 Structure pointers 49
4.3.5 address and indirection operators 50
4.4 ARCHITECTURAL TYPES 51
4.5 CHANNELS 51
4.5.1 FIFO code example 53
4.5.2 Arrays of channels 53
4.5.3 Restrictions on channel accesses 53
4.5.4 Timing and latency in FIFOs 55
4.6 INTERFACES: OVERVIEW 56
4.6.1 Interface declaration 56
4.6.2 Interface definition 57
4.6.3 Example interface to external code 58
4.6.4 Interface specifications 59
4.7 RAMS AND ROMS 61
4.7.1 Initialization 61
4.7.2 Inferring size from use 62
4.7.3 Accessing RAMs and ROMs 62
4.7.4 Differences between RAMs and arrays 62
4.7.5 RAM and ROM support on different devices 63
4.7.6 Multidimensional memory arrays 63
4.8 MPRAM (MULTI-PORTED RAMS) 64
4.8.1 Initialization of mprams 66
4.8.2 Mapping of different width mpram ports 66
4.8.3 mprams example 68
4.9 WOM (WRITE-ONLY MEMORY) 69
4.10 SEMA 69
4.11 SIGNAL 70
4.12 STORAGE CLASS SPECIFIERS 71
4.12.1 auto 71
4.12.2 extern (external variables) 72
4.13 EXTERN LANGUAGE CONSTRUCT 72
4.14 REGISTER 75
4.15 INLINE FUNCTIONS 75
4.16 STATIC 76
4.17 TYPEDEF 76
4.18 TYPEOF 77
4.19 CONST 78

4.20 VOLATILE **78**
4.21 COMPLEX DECLARATIONS **78**
4.21.1 Macro expressions in widths 78
4.21.2 < > (type clarifier) 79
4.21.3 Using signals to split up complex expressions 79
4.22 VARIABLE INITIALIZATION **80**

5 STATEMENTS **82**
5.1 SEQUENTIAL AND PARALLEL EXECUTION **82**
5.2 SEQ 83
5.3 REPLICATED PAR AND SEQ **83**
5.4 PRIALT **85**
5.5 USING PRIALT: EXAMPLES **86**
5.6 ASSIGNMENTS **88**
5.6.1 continue 89
5.6.2 goto 90
5.6.3 return [expression] 91
5.6.4 Conditional execution (if ... else) 91
5.6.5 while loops 92
5.6.6 do ... while loops 93
5.6.7 for loops 93
5.6.8 switch 95
5.6.9 break 96
5.6.10 delay 97
5.6.11 try... reset 97
5.6.12 trysema() 99
5.6.13 releasesema() 100

6 EXPRESSIONS **102**
6.1 INTRODUCTION TO EXPRESSIONS **102**
6.1.1 Clock cycles required 102
6.1.2 Breaking down complex expressions 102
6.1.3 Prefix and postfix operators 102
6.2 CASTING OF EXPRESSION TYPES **103**
6.2.1 Restrictions on casting 104
6.3 RESTRICTIONS ON RAMS AND ROMS **104**
6.4 ASSERT **106**
6.5 BIT MANIPULATION OPERATORS **108**
6.5.1 Shift operators 109
6.5.2 Take / drop operators 109
6.5.3 Concatenation operator 109
6.5.4 Bit selection 110
6.5.5 Width operator 111
6.6 ARITHMETIC OPERATORS **111**
6.7 RELATIONAL OPERATORS **113**
6.7.1 Signed/unsigned compares 114

6.7.2 Implicit compares 114
6.8 LOGICAL OPERATORS 114
6.8.1 Bitwise logical operators 115
6.9 CONDITIONAL OPERATOR 116
6.10 MEMBER OPERATORS (. / ->) 116

7 FUNCTIONS AND MACROS 118
7.1 FUNCTIONS AND MACROS: OVERVIEW 118
7.1.1 Functions and macros: language issues 118
7.1.2 Functions and macros: sharing hardware 120
7.1.3 Functions and macros: clock cycles 121
7.1.4 Functions and macros: examples 121
7.1.5 Accessing external names 123
7.1.6 Recursion in macros and functions 124
7.2 INTRODUCTION TO FUNCTIONS 124
7.2.1 Function definitions and declarations 125
7.2.2 Functions: scope 126
7.2.3 Arrays of functions 126
7.2.4 Function arrays: example 127
7.2.5 Function arrays example with static variables 128
7.2.6 Function pointers 129
7.2.7 Function pointers example 130
7.2.8 Simultaneous function calls 134
7.2.9 Multiple functions in a statement 136
7.3 INTRODUCTION TO MACROS 137
7.3.1 Non-parameterized macro expressions 137
7.3.2 Parameterized macro expressions 138
7.3.3 select operator 138
7.3.4 ifselect 139
7.3.5 Recursive macro expressions 140
7.3.6 Recursive macro expressions example 142
7.3.7 Shared expressions 142
7.3.8 Using recursion to generate shared expressions 143
7.3.9 Restrictions on shared expressions 143
7.3.10 let ... in 144
7.3.11 Macro procedures 145
7.3.12 Macro procedures compared to pre-processor macros 146

8 INTRODUCTION TO TIMING 148
8.1 STATEMENT TIMING 148
8.1.1 Example timings 148
8.1.2 Statement timing summary 154
8.2 AVOIDING COMBINATIONAL LOOPS 156
8.3 PARALLEL ACCESS TO VARIABLES 158
8.4 DETAILED TIMING EXAMPLE 159
8.5 TIME EFFICIENCY OF HANDEL-C HARDWARE 161
8.5.1 Reducing logic depth 161

8.5.2 Pipelining 164

9 CLOCKS OVERVIEW 167
9.1 LOCATING THE CLOCK 167
9.1.1 External clocks 168
9.1.2 Internal clocks fed from expressions 168
9.2 CURRENT CLOCK 169
9.3 MULTIPLE CLOCK DOMAINS 169
9.3.1 Channels communicating between clock domains........ 169
9.3.2 Simulating multiple clock domains 183

10 TARGETING HARDWARE AND SIMULATION 184
10.1 INTERFACING WITH THE SIMULATOR 184
10.1.1 Simulator input file format 185
10.1.2 Block data transfers........ 185
10.2 TARGETING FPGA AND PLD DEVICES 186
10.2.1 Summary of supported devices 187
10.2.2 Detecting the current device family........ 189
10.2.3 Targeting specific devices via source code........ 190
10.2.4 Specifying a global reset 193
10.3 USE OF RAMS AND ROMS WITH HANDEL-C 193
10.3.1 Asynchronous RAMs........ 194
10.3.2 Synchronous RAMs 200
10.3.3 Targeting Stratix and Cyclone memory blocks 216
10.3.4 Using on-chip RAMs in Actel devices........ 217
10.3.5 Using on-chip RAMs in Altera devices 218
10.3.6 Using on-chip RAMs in Xilinx devices 219
10.3.7 Using external ROMs........ 219
10.3.8 Connecting to RAMs in foreign code 220
10.3.9 Using other RAMs 224

11 INTERFACING WITH EXTERNAL HARDWARE 225
11.1 INTERFACE SORTS........ 225
11.1.1 Reading from external pins bus_in 226
11.1.2 Registered reading from external pins: bus_latch_in 227
11.1.3 Clocked reading from external pins: bus_clock_in 228
11.1.4 Writing to external pins: bus_out........ 228
11.1.5 Bidirectional data transfer: bus_ts 228
11.1.6 Bidirectional data transfer with registered input: bus_ts_latch_in 229
11.1.7 Bidirectional data transfer with clocked input: bus_ts_clock_in........ 231
11.1.8 Example hardware interface........ 232
11.2 SIMULATING INTERFACES 235
11.3 BUSES AND THE SIMULATOR 237
11.4 MERGING PINS 238
11.4.1 Merging clock pins........ 238
11.4.2 Merging input pins........ 239
11.4.3 Merging tri-state pins........ 239

11.5 TIMING CONSIDERATIONS FOR BUSES 240
11.5.1 Example timing considerations for input buses...................................... 240
11.5.2 Example timing considerations for output buses.................................... 242
11.6 METASTABILITY 242
11.6.1 Techniques to minimize the problem 243
11.6.2 Using interfaces: External resynchronizing example 246
11.7 PORTS: INTERFACING WITH EXTERNAL LOGIC 248
11.8 SPECIFYING THE INTERFACE 249
11.9 TARGETING PORTS TO SPECIFIC TOOLS 250

12 OBJECT SPECIFICATIONS 252
12.1 SUMMARY OF SPECIFICATIONS...252
12.1.1 Compiler atttributes.. 252
12.1.2 Simulator attributes.. 253
12.1.3 Clock attributes.. 254
12.1.4 Channel attributes... 256
12.1.5 Channel and memory attributes 256
12.1.6 Memory attributes... 256
12.1.7 Interface and memory attributes 258
12.1.8 Interface attributes.. 259
12.1.9 Examples 260
12.2 BASE SPECIFICATION 261
12.3 BIND SPECIFICATION..261
12.4 BLOCK SPECIFICATION 263
12.5 BUFFER SPECIFICATION 266
12.6 BUSFORMAT SPECIFICATION 266
12.7 SPECIFYING THE CLOCK PIN FOR SSRAM....................................268
12.8 CLOCKPORT SPECIFICATION 269
12.9 DATA SPECIFICATION (PIN CONSTRAINTS) 271
12.10 DCI SPECIFICATION..272
12.11 EXTINST, EXTLIB, EXTFUNC SPECIFICATIONS 273
12.12 EXTPATH SPECIFICATION 275
12.13 FIFOLENGTH SPECIFICATION...275
12.14 INFILE AND OUTFILE SPECIFICATIONS.......................................276
12.15 INTIME AND OUTTIME SPECIFICATIONS 276
12.16 TIMING CONSTRAINTS EXAMPLE 277
12.17 MINPERIOD SPECIFICATION 280
12.18 OFFCHIP SPECIFICATION 281
12.19 PARANOIA SPECIFICATION 282
12.20 PIN SPECIFICATIONS...283
12.21 PORTS SPECIFICATION 284
12.22 PROPERTIES SPECIFICATION..285
12.23 PULL SPECIFICATION 287

12.24 QUARTUS_PROJ_ASSIGN SPECIFICATION 288
12.25 RATE SPECIFICATION 288
12.26 RCLKPOS, WCLKPOS AND CLKPULSELEN SPECIFICATIONS (SSRAM TIMING) 289
12.27 RESOLUTIONTIME SPECIFICATION 291
12.28 RETIME SPECIFICATION 291
12.29 SC_TYPE SPECIFICATION 292
12.30 SHOW SPECIFICATION 293
12.31 SPEED SPECIFICATION 293
12.32 STANDARD SPECIFICATION 293
12.32.1 Available I/O standards 294
12.32.2 I/O standards supported by different chips 296
12.32.3 I/O standard details 297
12.32.4 Differential I/O standards 301
12.33 STD_LOGIC_VECTOR SPECIFICATION 302
12.34 STRENGTH SPECIFICATION 303
12.35 SYNCHRONOUS SPECIFICATION 304
12.36 UNCONSTRAINEDPERIOD SPECIFICATION 304
12.37 VHDL_TYPE SPECIFICATION 305
12.38 WARN SPECIFICATION 307
12.39 WEGATE SPECIFICATION 307
12.40 WESTART AND WELENGTH SPECIFICATIONS 307

13 HANDEL-C PREPROCESSOR 310
13.1 PREPROCESSOR MACROS 310
13.2 FILE INCLUSION 311
13.3 CONDITIONAL COMPILATION 312
13.4 LINE CONTROL 313
13.5 CONCATENATION IN MACROS 313
13.6 ERROR GENERATION 314
13.7 PREDEFINED MACRO SUBSTITUTION 314
13.8 LINE SPLICING 314

14 LANGUAGE SYNTAX 316
14.1 LANGUAGE SYNTAX CONVENTIONS 316
14.2 KEYWORD SUMMARY 316
14.3 CONSTANT EXPRESSIONS 321
14.3.1 Identifiers: syntax 321
14.3.2 Integer constants: syntax 322
14.3.3 Character constants: syntax 322
14.3.4 Strings: syntax 322
14.3.5 Floating-point constants: syntax 322
14.4 FUNCTIONS AND DECLARATIONS: SYNTAX 323

14.5 MACRO/SHARED EXPRS/PROCS: SYNTAX 324
14.6 INTERFACES: SYNTAX 325
14.7 STRUCTURES AND UNIONS: SYNTAX 326
14.8 ENUMERATED TYPES: SYNTAX 326
14.9 SIGNAL SPECIFIERS: SYNTAX....................................326
14.10 CHANNEL SYNTAX 326
14.11 RAM SPECIFIERS: SYNTAX....................................327
14.12 DECLARATORS: SYNTAX....................................327
14.13 FUNCTION PARAMETERS: SYNTAX 327
14.14 TYPE NAMES AND ABSTRACT DECLARATORS: SYNTAX 328
14.15 STATEMENTS: SYNTAX 328
14.15.1 Compound statements with replicators 331
14.16 REPLICATORS: SYNTAX 331
14.17 EXPRESSIONS: SYNTAX 332

15 INDEX 335

Conventions

A number of conventions are used in this document. These conventions are detailed below.

Warning Message. These messages warn you that actions may damage your hardware.

Handy Note. These messages draw your attention to crucial pieces of information.

Hexadecimal numbers will appear throughout this document. The convention used is that of prefixing the number with '0x' in common with standard C syntax.

Sections of code or commands that you must type are given in typewriter font like this:

```
void main();
```

Information about a type of object you must specify is given in italics like this:

`copy` *SourceFileName DestinationFileName*

Optional elements are enclosed in square brackets like this:

`struct` [type_Name]

Curly brackets around an element show that it is optional but it may be repeated any number of times.

string `::= "{`*character*`}"`

Assumptions & Omissions

This manual assumes that you:

- have used Handel-C or have the Handel-C Language Reference Manual
- are familiar with common programming terms (e.g. functions)
- are familiar with MS Windows

This manual does not include:

- instruction in VHDL or Verilog
- instruction in the use of place and route tools
- tutorial example programs. These are provided in the Handel-C User Manual

1 Introduction

1.1 References

- The C Programming Language 2nd Edition
  Kernighan, B. and Ritchie, D.
  Prentice-Hall, 1988
- Altera Databook
  Altera 2004
  `www.altera.com/literature/lit-index.html`
- Xilinx Data Book
  Xilinx 2004
  `www.xilinx.com/literature/index.htm`
- VHDL for logic synthesis
  Author: Andrew Rushton
  Publisher: John Wiley and Sons
  ISBN: 0-471-98325-X
  Published: May 1998
- IEEE standard 1364 -1995
  IEEE Standard Hardware Description Language Based on the Verilog®
  Hardware Description Language.
  `http://standards.ieee.org/`

2 Getting started with Handel-C

2.1 Basic concepts

Handel-C uses much of the syntax of conventional C with the addition of inherent parallelism. You can write sequential programs in Handel-C, but to gain maximum benefit in performance from the target hardware you must use its parallel constructs. These may be new to some users. If you are familiar with conventional C you will recognize nearly all the other features.

Handel-C programs

- Parallel programs
- Channel communications
- Scope and variable sharing

2.1.1 Handel-C programs

Since Handel-C is based on the syntax of conventional C, programs written in Handel-C are implicitly sequential. Writing one command after another indicates that those instructions should be executed in that exact order. To execute instructions in parallel, you must use the `par` keyword.

Handel-C provides constructs to control the flow of a program. For example, code can be executed conditionally depending on the value of some expression, or a block of code can be repeated a number of times using a loop construct.

You can express your algorithm in Handel-C without worrying about how the underlying computation engine works. This philosophy makes Handel-C a programming language rather than a hardware description language. In some senses, Handel-C is to hardware what a conventional high-level language is to microprocessor assembly language.

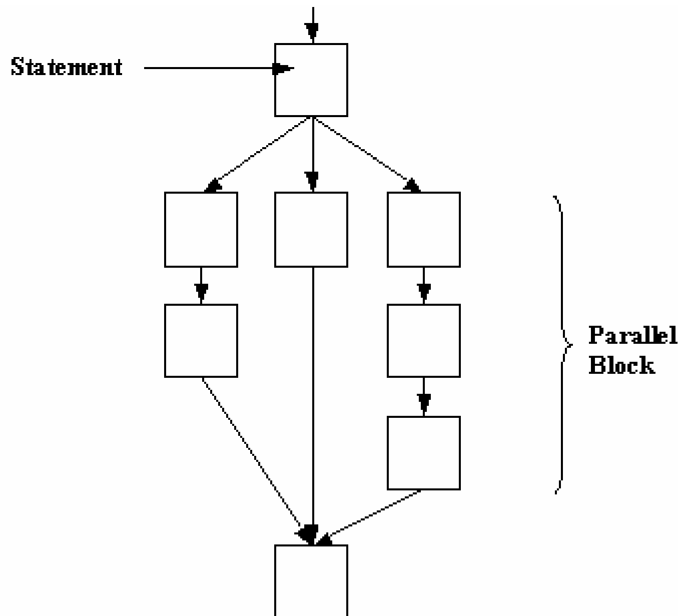
The hardware design that DK produces is generated directly from the Handel-C source program. There is no intermediate 'interpreting' layer as exists in assembly language when targeting general-purpose microprocessors. The logic gates that make up the final Handel-C circuit are the assembly instructions of the Handel-C system.

2.1.2 Parallel programs

The target of the Handel-C compiler is low-level hardware. This means that you get massive performance benefits by using parallelism. It is essential for writing efficient programs to instruct the compiler to build hardware to execute statements in parallel. Handel-C parallelism is true parallelism, not the time-sliced parallelism familiar from general-purpose computers. When instructed to execute two instructions in parallel,

those two instructions will be executed at exactly the same instant in time by two separate pieces of hardware.

When a parallel block is encountered, execution flow splits at the start of the parallel block and each branch of the block executes simultaneously. Execution flow then re-joins at the end of the block when all branches have completed. Any branches that complete early are forced to wait for the slowest branch before continuing.



This diagram illustrates the branching and re-joining of the execution flow. The left hand and middle branches must wait to ensure that all branches have completed before the instruction following the parallel construct can be executed.

2.1.3 Channel communication

Channels provide a link between branches executing in parallel. One parallel branch outputs data onto the channel and the other branch reads data from the channel.

Channels can be constructed with and without FIFO capacities

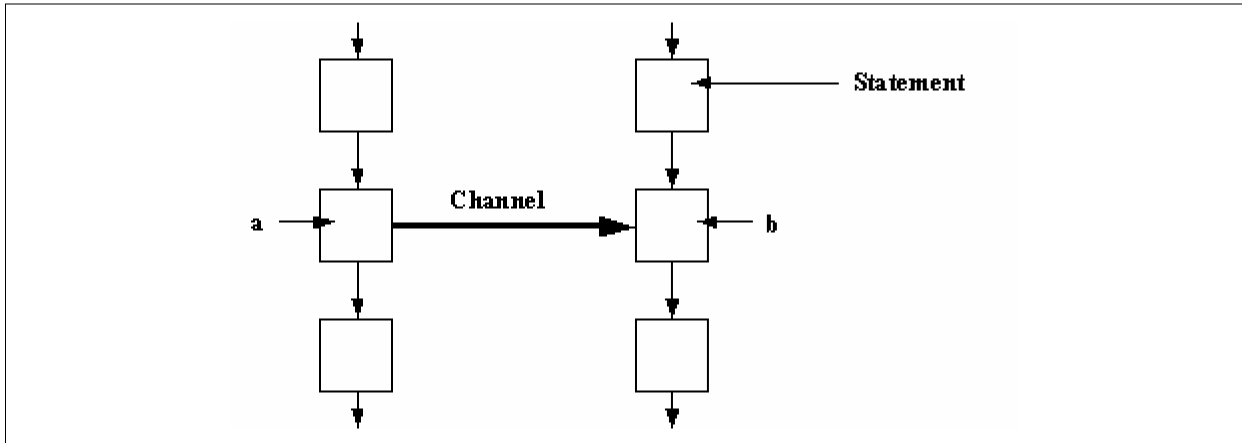
- Channels constructed as FIFOs

  A channel can be constructed as a FIFO queue. In this case, the data is written to the head of the FIFO and is read from the tail. If the FIFO is full, a write blocks until an element is read from the FIFO. If the FIFO is empty, a read blocks until there is data ready to be read.

- Channels constructed without FIFO capacity

  These channels provide synchronization between parallel branches because the data transfer can only complete when both the transmitter and the receiver are ready. If one side is not ready, the other must wait.

Channel synchronization

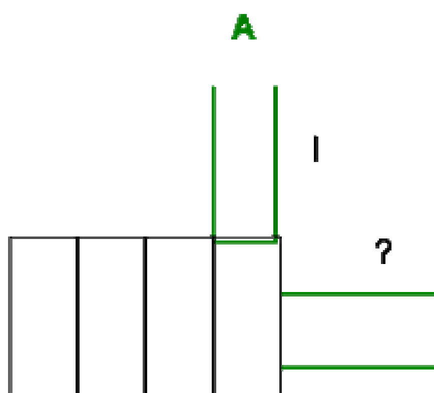


SYNCHRONIZATION OF NORMAL CHANNELS

Here, the channel is shown transferring data from the left branch to the right branch. If the left branch reaches point a before the right branch reaches point b, the left branch waits at point a until the right branch reaches point b.

Communication without synchronization

If you are using a channel FIFO, the left branch will not have to wait at point a if there is space in the FIFO. Instead, it can write to the FIFO once per clock tick until the FIFO is full. Only then will it have to wait. Each time the right branch reads from the FIFO at point b, the data at the head of the FIFO is read, and the next piece of data becomes the head. The right branch must wait if the FIFO is empty.



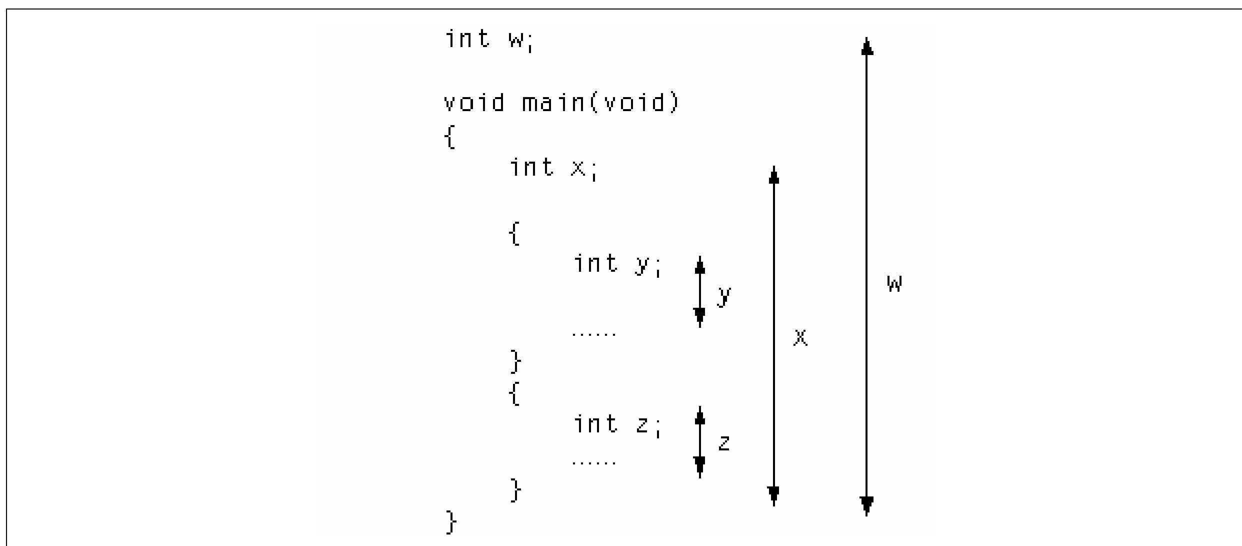
In this case, the two branches will not be synchronized after every read and write.

2.1.4 Scope and variable sharing

The scope of declarations is based around code blocks. A code block is denoted with {...} brackets. This means that:

- Global variables must be declared outside all code blocks
- An identifier is in scope within a code block and any sub-blocks of that block.

The scope of variables is illustrated below:



Since parallel constructs are simply code blocks, variables can be in scope in two parallel branches of code. This can lead to resource conflicts if the variable is written to simultaneously by more than one of the branches. Handel-C states that a single variable must not be written to by more than one parallel branch but may be read from by several parallel branches.

If you wish to write to the same variable from several processes, the correct way to do so is by using channels which are read from in a single process. This process can use a `prialt` statement to select which channel is ready to be read from first, and that channel is the only one which will be allowed to write to the variable.

```
while(1)
    prialt
        {
            case chan1 ? y:
                break;
            case chan2 ? y:
                break;
            case chan3 ? y:
                break;
        }
```

In this case, three separate processes can attempt to change the value of `y` by sending data down the channels, `chan1`, `chan2` and `chan3`. `y` will be changed by whichever process sends the data first.

A single variable should not be written to by more than one parallel branch.

3 Language basics

3.1 Program structure

Sequential structure

As in a conventional C program, a Handel-C program consists of a series of statements which execute sequentially. These statements are contained within a `main()` function that tells the compiler where the program begins. The body of the `main` function may be split into a number of blocks using `{...}` brackets to break the program into readable chunks and restrict the scope of variables and identifiers.

Handel-C also has functions, variables and expressions similar to conventional C. There are restrictions where operations are not appropriate to hardware implementation and extensions where hardware implementation allows additional functionality.

Parallel structure

Unlike conventional C, Handel-C programs can also have statements or functions that execute in parallel. This feature is crucial when targeting hardware because parallelism is the main way to increase performance by using hardware. Parallel processes can communicate using channels. A channel is a point-to-point link between two processes.

Overall structure

The overall program structure consists of one or more `main` functions, each associated with a clock. This is unlike conventional C, where only one `main` function is permitted. You would only use more than one main function if you needed parts of your program to run at different speeds (and so use different clocks). A `main` function is defined as follows:

```
Global Declarations

Clock Definition
void main(void)
{
    Local Declarations

    Body Code
}
```

The `main()` function takes no arguments and returns no value. This is in line with a hardware implementation where there are no command line arguments and no environment to return values to. The ***argc***, ***argv*** and ***envp*** parameters and the return value familiar from conventional C can be replaced with explicit communications with an external system (e.g. a host microprocessor) within the body of the program.

3.2 Comments

Handel-C uses the standard /* ... */ delimiters for comments. These comments may not be nested. For example:

```
/* Valid comment */

/* This is /* NOT */ valid */
```

Handel-C also provides the C++ style // comment marker which tells the compiler to ignore everything up to the next new line. For example

```
x = x + 1;  // This is a comment
```

3.3 Statement summary

Statement	Meaning
`par {...}`	Parallel execution
`seq {...}`	Sequential execution
`par (`*Init* `;` *Test* `;` *Iter*`){...}`	Parallel replication
`seq (`*Init* `;` *Test* `;` *Iter*`){...}`	Sequential replication
`Variable =` *Expression*`;`	Assignment
Variable `++;`	Increment
Variable `--;`	Decrement
`++` *Variable*`;`	Increment
`--` *Variable*`;`	Decrement
Variable `+=` *Expression*`;`	Add and assign
Variable `-=` *Expression*`;`	Subtract and assign
Variable `*=` *Expression*`;`	Multiply and assign
Variable `/=` *Expression*`;`	Divide and assign
Variable `%=` *Expression*`;`	Modulo and assign
Variable `<<=` *Expression*`;`	Shift left and assign
Variable `>>=` *Expression*`;`	Shift right and assign
Variable `&=` *Expression*`;`	Bitwise AND and assign
Variable `|=` *Expression*`;`	Bitwise OR and assign
Variable `^=` *Expression*`;`	Bitwise XOR and assign
Channel `?` *Variable*`;`	Channel input
Channel `!` *Expression*`;`	Channel output
`if (`*Expression*`) {`*statement*`} [else {`*statement*`}]`	Conditional execution
`ifselect (`*Expression*`) {`*statement*`} [else {`*statement*`}]`	Conditional compilation
`while (`*Expression*`) {`*statement*`}`	Iteration
`do {...} while (`*Expression*`);`	Iteration
`for (`*Init* `;` *Test* `;` *Iter*`) {...}`	Iteration
`break;`	Loop, switch and prialt termination
`continue;`	Resume execution
`return[([`*Expression*`])];`	Return from function
`goto` *label*`;`	Jump to label
`switch (`*Expression*`) {`*statement*`}`	Selection
`prialt {`*statement*`}`	Channel alternation

`releasesema()`	Make semaphore available after use of `trysema` expression
`try{...}` `reset(`*`Condition`*`){`*`statement`*`}`	Perform statements on reset condition
`delay;`	Single cycle delay

Note: RAM and ROM elements, signals and array elements are included in the set of variables above. However,

```
ram x [3];
```

```
x[0]++;
```

is invalid.

The assignment group of operations and the increment and decrement operations are included as statements to reflect the fact that Handel-C expressions cannot contain side effects.

3.4 Operator summary

The following table lists all operators. Entries at the top have the highest precedence and entries at the bottom have the lowest precedence. Entries within the same group have the same precedence. Precedence of operators is as expected from conventional C. For example:

```
x = x + y * z;
```

This performs the multiplication before the addition. Brackets may be used to ensure the correct calculation order as in conventional C.

Note that assignments are not true operators in Handel-C.

Operator	Meaning
trysema	Test if semaphore owned. Take if not
select(Constant, Expr, Expr)	Compile-time selection
Expression [Expression]	Array or memory subscripting
Expression [Constant]	Bit selection
Expression [Constant: Constant]	Bit range extraction. One of the two constants may be omitted (but not both).
functionName (Arguments)	Function call
pointerToStructure->member	Structure reference
structureName.member	Structure reference
! Expression	Logical NOT
~ Expression	Bitwise NOT
- Expression	Unary minus
+ Expression	Unary plus
& object	Yields pointer to operand
* pointer	Yields object or function that the operand points to
width(Expression)	Width of expression
(Type) Expression	Type casting
Expression <- Constant	Take LSBs
Expression \\ Constant	Drop LSBs
Expression * Expression	Multiplication
Expression / Expression	Division
Expression % Expression	Modulo arithmetic
Expression + Expression	Addition
Expression - Expression	Subtraction
Expression << Expression	Shift left
Expression >> Expression	Shift right
Expression @ Expression	Concatenation
Expression < Expression	Less than
Expression > Expression	Greater than
Expression <= Expression	Less than or equal
Expression >= Expression	Greater than or equal
Expression == Expression	Equal
Expression != Expression	Not equal
Expression & Expression	Bitwise AND
Expression ^ Expression	Bitwise XOR

`Expression | Expression`	Bitwise OR
`Expression && Expression`	Logical AND
`Expression || Expression`	Logical OR
`Expression ? Expr : Expr`	Conditional selection
`assert`	diagnostic macro to print to stderr

3.5 Type summary

The most common types that may be associated with a variable, and the prefixes for architectural and compound types are listed below.

Common logic types

Type	Width
`int`	See *Note 1
`[signed | unsigned] int n`	n bits
`[signed | unsigned] int undefined`	Compiler infers width
`[signed | unsigned] char`	8 bits
`[signed | unsigned] short`	16 bits
`[signed | unsigned] long`	32 bits
`[signed | unsigned] int32`	32 bits
`[signed | unsigned] int64`	64 bits
`typeof (Expression)`	Yields type of object

*Note 1: Width will be inferred by compiler unless the '`set intwidth = n`' command appears before the declaration.

Architectural types

Prefixes to the above types for different architectural object types are:

Prefix	Object
chan	Channel
chanin	Simulator channel
chanout	Simulator channel
ram	Internal or external RAM
rom	Internal or external ROM
signal	Wire
wom	WOM within multi-port memory

Compound types

The compound types are:

Prefix	Object
struct	Structure
mpram	Multi-port memory

Special types

Type	Object
interface	Interface to external logic or device
sema	Semaphore. Has no width or logic type

Interfaces connect to logic beyond the Handel-C design, whether on the same or a different device.

3.6 Comparison of Handel-C and ANSI-C

Handel-C has many similarities to ANSI-C (ISO-C). However, Handel-C is a language for digital logic design, which means that the way in which DK interprets it may different to the way in which compilers interpret ANSI-C for software design. Handel-C has some extensions to ANSI-C, to allow additional functionality for hardware design. It also lacks some ANSI-C constructs which are not appropriate to hardware implementation.

This section summarizes the differences between Handel-C and ANSI-C. It is not a definitive list. Refer to specific sections to see how DK implements each of the language constructs.

3.6.1 Handel-C v C: types and type operators

Handel-C supports all ANSI-C types apart from `float`, `double` and `long double`. You can still perform floating-point arithmetic.

`char`, `short` and `long` are supported to help the porting of code from ANSI-C. However, it can be better (more efficient in hardware terms) to re-express these as a `signed` or `unsigned int` of a specific width. In Handel-C, `int`s are not limited to 64 bits.

Handel-C has a range of additional types for creating channels and interfaces between different hardware blocks, and for specifying memories and signals. The Celoxica wide number library provides `signed` and `unsigned` compiler-independent implementations of `int32` and `int64`.

Handel-C also allows all ANSI-C storage class specifiers and type qualifiers, but `volatile` and `register` have no meaning in hardware terms, and are accepted for compatibility only.

You have to specify the size of an array in Handel-C. For example, you couldn't write:

```
int ai[SIZE]
```

and then `# define SIZE`.

Handel-C variables can only be initialized if they are `static`, `const` or global. Otherwise, you must assign a value to them in a statement.

```
int a = 8 //not allowed
```

```
int a;
a = 8; // OK
```

```
static int a = 8; // OK
```

The Handel-C `typeof` operator allows you to determine the type of an object at compile time.

3.6.2 Handel-C v C: floating-point variables

There are no floating-point types (`float`, `double` or `long double`) in Handel-C.

Floating-point arithmetic is more complex than integer or fixed-point arithmetic and tends to require more hardware. If you are porting C code to Handel-C, check if there is a way to avoid using floating-points. For example, you might be able to use fixed-point values (which have a binary point), or to change the units to remove the decimal places (e.g. use pence or cents instead of pounds or dollars).

If you do need to use floating-point arithmetic, use the Celoxica floating-point library. This allows you to specify the exact width of the mantissa and exponent. You can download the floating-point library from the downloads section of the Celoxica support web site. If you can use fixed-point arithmetic, use the Celoxica fixed-point library. This is provided in the Platform Developer's Kit.

3.6.3 Handel-C v C: variable widths and casting

Handel-C widths

Handel-C types are not limited to specific widths. When you define a Handel-C variable, you should specify the minimum width required, to minimize hardware usage. For example, if you have a variable, `x`, that can hold a value between 1 and 20, use a 5-bit `int`:

```
int 5 x;
```

Casting

There is no automatic conversion between `signed` and `unsigned` values in Handel-C, you have to explicitly cast them:

```
int 12 x;
unsigned int 12 y;
y = x; //not allowed
y = (unsigned) x; //OK
```

Similarly, there is no automatic type conversion. If you wanted to add an `int 5` and a `long` together, you would need to pad the `int` to 32 bits by using the concatenation operator. However, it would be more usual to perform arithmetic on `int`s of specific widths.

Pointers can be cast to void and back, to another pointer of the same type except for the addition or removal of a type qualifier, between `signed` and `unsigned`, and between similar structs (e.g. a `struct` with identical elements except for the width of the types).

You cannot perform the following casts in Handel-C:

- from a pointer of one type to a pointer of another type (except for those listed above)
- from a pointer to an integral type
- from an integral type to a pointer
- from a pointer to a function to a pointer to another function type

Arithmetic and comparisons on variables of different width

In Handel-C you need to use the concatenation operator or the `take` operator when performing arithmetic or comparisons on variables of different width. For example:

```
int 12 x;
int 8 y;

x = y; // not allowed
y = x; //not allowed
x = y[7] @ y[7] @ y[7] @ y[7] @ y // OK
y = x <-8; // OK; preserves the sign and copies the 7 LSBs
```

Alternatively you can use the width adjustment macros in the Celoxica standard macro library, `stdlib.hcl`. The `adju()` macro adjusts the width of unsigned numbers and the `adjs()` macro adjusts the widths of signed numbers. The standard library is now provided as part of the Platform Developer's Kit (PDK). If you do not already have a copy of PDK, you can download it from the support section of the Celoxica web site.

sizeof

There is no `sizeof` in Handel-C. For simple types (`signed` and `unsigned char`, `int`, `long` and `short`), you can use the width operator. For example, `sizeof long` in C is equivalent to `width long` in Handel-C, except that the number of bytes is returned in C and the number of bits is returned in Handel-C.

3.6.4 Handel-C v C: side effects

There are restrictions on how you can use side-effects in Handel-C, because each statement must only take one clock cycle. Each statement can only contain a single assignment, or an increment or a decrement.

This means that:

- Shortcut assignments (e.g. +=) must appear as standalone statements.
- The initialization and iteration phases of `for` loops must be statements, not expressions.

If you are porting ANSI-C code, complex statements have to be re-written as multiple single statements. It is often more efficient to run these statements in parallel. You cannot use comma operators in Handel-C.

If you had the following expression written in ANSI-C:

```
a = b = ++c, d+e;
```

this could be separated into single statements in Handel-C:

```
seq
{
    ++c;
    b = d + e;
    a = b;
}
```

However, you could rewrite the same code to run all the statements in parallel:

```
par
{
    ++c;
    a = d + e;
    b = d + e;
}
```

3.6.5 Handel-C v C: functions

There are a number of differences in the way in which functions can be used in ANSI-C and Handel-C.

In Handel-C:

- Functions may not be called recursively, since all logic must be expanded at compile-time to generate hardware.
- You can only call functions in expression statements. These statements must not contain any other calls or assignments.
- Variable length parameter lists are not supported.
- Old-style ANSI-C function declarations (where the type of the parameters is not specified) are not supported.
- `main()` functions take no arguments and return no values.
- You can have more than one `main()` function. Each `main()` function is associated with a clock. If you have more than one `main()` function in the same source file, they must all use the same clock.
- You can have arrays of functions and `inline` functions. These are useful when you are writing parallel code.

Re-writing recursive functions

If you want to port code that uses recursive functions to Handel-C, the options for rewriting it include:

- Using recursive macro expressions or recursive macro procedures. (It must be possible to determine the depth of recursion at compile-time.)
- Creating multiple copies of a function.

- Re-writing the function to create iterative code. This is relatively easy if the function is calling itself (simple recursion), and the recursive call is the last item within the function definition (tail recursion).

The following ANSI-C function has simple tail recursion:

```
unsigned long Factorial (unsigned long n)
{
    if (n==0)
        return 1;
    else
        return n * Factorial (n-1);
}
```

It can be re-written in Handel-C as:

```
unsigned int 32 Factorial (unsigned int 32 n)
{
    unsigned int 32 nfact;
    nfact = 1;
    if (n == 0)
        delay;
    else
   {
        while (n != 0)
        {
            nfact *= n;
            --n;
        }
    }
    return nfact;
}
```

Note that the `if...else` is required to prevent the possibility of a combinatorial loop if the `while` loop is not executed.

3.6.6 Handel-C v C: loop statements

`for` loops in Handel-C are slightly different to those in ANSI-C: the initialization and iteration steps are written as statements rather than expressions. This is because of restrictions on side effects in expressions in Handel-C.

You need to ensure that loop statements take at least one clock cycle in Handel-C. This means that:

- you cannot have empty loops in Handel-C

- you need to ensure that the body of a loop will always execute at least once, or else provide an alternative execution point using an `if...else`.

For example, if you had the following ANSI-C code:

```
while ((--i) != 0)
{
    MyFunction (i);
}
```

The `while` loop would not be executed if `i` was equal to 0. You could re-write this in Handel-C as:

```
--i;
if (i != 0)
    while (i != 0)
    {
        MyFunction (i);
        --i;
    }
else
    delay;
```

Note that you need to decrement `i` before you enter the while body to preserve the order dependency of the ANSI-C code.

3.6.7 Handel-C v C: unions

If there is no relationship between members of the union, you can use a `struct` instead.

If the members of the union are of related types (e.g. `int`, `long` and `char`), you can "reuse" a single variable which is the width of the widest variable in the union. For example, if you have the following union in your C code:

```
union
{
    unsigned long ul;
    unsigned char uc;
    short ss;
} u;
```

you could use a single variable of the same width as the long:

```
unsigned int 32 i;
```

You could then get values equivalent to `ul`, `ss` and `ul` by casting and using the `take` operator:

u.ul would be written as i

u.uc would be written as i <-8

u.ss would be written as (signed) (i<-16)

Note that in ANSI-C there is no guarantee about whether ul, uc and ss would share storage, and so the Handel-C code above might not exactly reproduce the behaviour of the ANSI-C code in your C compiler.

3.6.8 Handel-C v C: data input and output

Handel-C does not have functions equivalent to scanf() and printf(). You can use scanf() and printf() when you are simulating a design, as Handel-C allows you to make calls to Handel-C functions. Alternatively, you can use the Handel-C infile and outfile specifications. Both these methods allow you to debug an algorithm before you build it in hardware.

When you are targeting hardware, data is passed between different parts of your Handel-C design using channels. If your Handel-C design will receive data from or send data to external components, you need to specify an interface. These external components might be written in EDIF, Verilog or VHDL, or they could be an additional component specified in Handel-C.

3.6.9 Handel-C v C: memory allocation

Memory allocation is not relevant when you are targeting hardware, so Handel-C has no equivalent of malloc and free.

You can use Handel-C to create RAM or ROM blocks on an FPGA or PLD, or interface to off-chip memory.

3.6.10 Handel-C v C: standard library

The standard library in Handel-C is called stdlib.hcl. This has no relationship to the C library, stdlib.lib or to stdio.lib.

stdlib.hcl contains bit manipulation and arithmetic macros.

The standard library is now provided as part of the Platform Developer's Kit (PDK). If you do not already have a copy of PDK, you can download it from the support section of the Celoxica web site.

3.6.11 C and Handel-C types and objects

In both	Conventional C only	Handel-C only
int	double	chan
unsigned	float	ram
char	union	rom
long		wom
short		mpram
enum		signal
register		chanin
static		chanout
extern		undefined
struct		interface
volatile		<>
void		inline
const		typeof
auto		
signed		
typedef		

3.6.12 Expressions in C and Handel-C

<table>
<tr><th>In both</th><th>Conventional C only</th><th>Handel-C only</th></tr>
<tr><td>* (pointer indirection)</td><td>sizeof</td><td>select(…)</td></tr>
<tr><td>& (address of)</td><td></td><td>width(…)</td></tr>
<tr><td>-</td><td></td><td>@</td></tr>
<tr><td>+</td><td></td><td>\\</td></tr>
<tr><td>* (multiplication)</td><td></td><td><-</td></tr>
<tr><td>/</td><td></td><td>[:]</td></tr>
<tr><td></td><td></td><td>let…in</td></tr>
<tr><td>%</td><td></td><td></td></tr>
<tr><td><<</td><td></td><td></td></tr>
<tr><td>>></td><td></td><td></td></tr>
<tr><td>></td><td></td><td></td></tr>
<tr><td><</td><td></td><td></td></tr>
<tr><td>>=</td><td></td><td></td></tr>
<tr><td><=</td><td></td><td></td></tr>
<tr><td>==</td><td></td><td></td></tr>
<tr><td>!=</td><td></td><td></td></tr>
<tr><td>& (bitwise and)</td><td></td><td></td></tr>
<tr><td>^</td><td></td><td></td></tr>
<tr><td>|</td><td></td><td></td></tr>
<tr><td>? :</td><td></td><td></td></tr>
<tr><td>[]</td><td></td><td></td></tr>
<tr><td>!</td><td></td><td></td></tr>
<tr><td>&&</td><td></td><td></td></tr>
<tr><td>~</td><td></td><td></td></tr>
<tr><td>||</td><td></td><td></td></tr>
<tr><td>-></td><td></td><td></td></tr>
</table>

3.6.13 Statements in C and Handel-C

In both	Handel-C only
`{;}`	`par`
`switch`	`delay`
`do … while`	`?`
`while`	`!`
`if … else`	`prialt`
`for (;;)`	`seq`
`break`	`ifselect`
`continue`	
`return`	
`goto`	
`assert`	assert is an expression in Handel-C and not the same as in ANSI-C

3.7 Handel-C constructs not found in ANSI-C

Handel-C is designed to target hardware. It allows you to specify timing and to target components such as memory, ports, buses and wires. One of the most important differences to ANSI-C is the ability to create code that executes in parallel.

Handel-C constructs that are not found in ANSI-C are listed below.

Parallelism

The `par` keyword specifies that a block of code should execute in parallel. Each statement within the block is executed in the same clock cycle. If the `par` keyword is not used, statements within a code block are executed sequentially. You can use the `seq` keyword to make this more explicit.

Channels allow communication between parallel branches of code. They are specified using the `chan` keyword, or by `chanin` and `chanout` when you are simulating code. You can read from and write to channels using statements of the form

```
Channel ? Variable; //reads from a channel
Channel ! Expression;  //writes to a channel
```

`prialt` statements are used with multiple channels, to select the first one that is ready for a read or write.

Semaphores (`sema`) allow you to coordinate the use of resources that are shared between parallel branches of code. The `trysema()` construct tests to see if the sema is owned.

The `releasesema()` construct frees a semaphore once it is no longer needed by a resource.

inline functions, arrays of functions, macro procedures and macro expressions help you to create multiple copies of functions. You need copies of a function if it is to be accessed by parallel branches of code.

Timing

The `set clock` construct specifies the clock source for each `main()` function. You can have more than one clock interfacing with your design by specifying more than one `main()` function. If you want to simulate code, you can set a "dummy" clock. You can specify the frequency of a clock using the `rate` specification. The `clockport` specification can be used to assign a dedicated clock input resource on your target device. You can also use it to specify that a port on an interface is used to drive the Handel-C clock.

Assignments and `delay` take one clock cycle in Handel-C. Everything else is "free". The `delay` statement does nothing, but takes one clock cycle. This can be used to avoid timing conflicts, such as combinational loops.

The `intime` and `outtime` specifications can be used to specify the maximum delay between an interface and an element interacting with an interface, (e.g. the port reading data into a RAM).

Compile-time selection and expansion and generic code

When you write code to target hardware, all logic needs to be expanded at compile time. This means that you cannot use recursive functions. However, macro procedures, macro expressions and shared expressions allow compile-time recursion in combination with the `select`, `ifselect` and `let...in` constructs.

The `select` operator allows you to select between expressions at compile time. It is similar to the conditional operator (`cond ? expr1: expr2`), but no hardware is generated for the conditional.

The `ifselect` construct is similar to an `if...else`, but selects between alternative blocks of code at compile time.

The `typeof` operator allows the type of an object to be determined at compile time. The `undefined` keyword specifies that the compiler should infer the width of a variable. These constructs allow you to create parameterizable code. For example, the Celoxica fixed-point library uses macros to pass the integer width and fraction width of a fixed-point number into code that creates a `struct` to hold the number.

Targeting hardware; FPGAs and PLDs

The `set family` and `set part` constructs allow you to specify the device you want to target in your source code. You can also set the device using the DK GUI.

Targeting hardware; memory

The `ram` and `rom` keywords allow you to create on-chip RAM and ROM, and to interface to external RAM and ROM. If you want to create a block RAM, use the `block` specification.

To interface to off-chip RAMs or ROMs, use the `offchip` specification. The `addr`, `data`, `we`, `cs`, `oe` and `clk` specifications define the pins used between the FPGA/PLD and external RAM or ROM.

An `mpram` is a multi-ported RAM. This allows you to read from and write to a RAM within the same clock cycle, or to make two read or two write accesses. Individual ports can be specified as read/write, read-only and write-only using the `ram`, `rom` and `wom` keywords.

If you want to interface to a dedicated memory resource on the FPGA/PLD, use the `ports` specification.

The `clkpulselen`, `rclkpos` and `wclkpos` specifications allow you to synchronize a RAM clock with the Handel-C clock. The `westart`, `welength` and `wegate` specifications allow you to specify timing of a RAM clock that is asynchronous to the Handel-C clock.

Targeting hardware; wires

If you specify a `signal` in Handel-C, this creates a wire in hardware. A signal takes on the value assigned to it but only for that clock cycle. The value assigned to it can be read back during the same clock cycle.

Targeting hardware; resets

`set reset` allows you to reset your device into a known state. It can also be used to configure devices that are not in a known state at start up.

`try...reset` allows you to specify some actions that occur if a particular condition becomes true within a particular block of hardware.

Interfacing to existing modules and to peripherals

Handel-C interfaces can be used to connect to external devices or to external logic on your target FPGA/PLD, such as other programs written in Handel-C, VHDL or Verilog.

Port-type interfaces allow you connect to external logic. The `bind`, `properties` and `std_logic_vector` specifications allow you to parameterize interfaces connecting to external code.

Bus-type interfaces connect to pins connected to peripheral devices. The `standard` specification selects the I/O standard for interface pins and the `strength` specification determines the drive current. You can use the `dci` specification if you want to use digital controlled impedance. The `pull` specification allows you to create a pull up or pull down resistor for bus pins. The `speed` specification allows you to specify the slew rate for the output buffer on pins.

The `extern "`*`language`*`"` construct is the same as that found in C++. It allows you to connect to blocks of ANSI-C or C++ code for co-simulation.

Bit manipulation

Handel-C types are not constrained to a specific width, so you can specify the exact width needed for a variable to minimize hardware usage. Bit manipulation is required to connect objects of different widths. In addition to the ANSI-C bit manipulation operators,

Handel-C provides the `take` and `drop` operators, which take and drop the least significant bits of a variable, and the concatenation operator, to extend variable width. The bit selection operator, allows you to select individual bits of a variable.

4 Declarations

4.1 Introduction to types

Handel-C uses two kinds of objects: logic types and architecture types. The logic types specify variables. The architecture types specify variables that require a particular sort of hardware architecture (e.g., ROMs, RAMs and channels).

Both kinds are specified by their scope (`static` or `extern`), their size and their type. Architectural types are also specified by the logic type that uses them.

Both types can be used in derived types (such as structures, arrays or functions) but there may be some restrictions on the use of architectural types.

Specifiers

The type specifiers `signed`, `unsigned` and `undefined` define whether the variable is signed and whether it takes a default defined width.

You can use the storage class specifiers `extern` and `static` to define the scope of any variable.

Functions can have the storage class `inline` to show that they are expanded in line, rather than being shared.

Type qualifiers

Handel-C supports the type qualifiers `const` and `volatile` to increase compatibility with ANSI-C. These can be used to further qualify logic types.

Disambiguator

Handel-C supports the extension < >. This can be used to clarify complex declarations of architectural types.

4.1.1 Handel-C values and widths

A crucial difference between Handel-C and conventional C is Handel-C's ability to handle values of arbitrary width. Since conventional C is targeted at general-purpose microprocessors it handles 8, 16 and 32 bit values well but cannot easily handle other widths. When targeting hardware, there is no reason to be tied to these data widths and so Handel-C has been extended to allow types of any number of bits.

Handel-C has also been extended to cope with extracting bits from values and joining values together to form wider values. These operations require no hardware and can provide great performance improvements over software.

When writing programs in Handel-C, care should be taken that data paths are no wider than necessary to minimize hardware usage. While it may be valid to use 32-bit values for all items, a large amount of unnecessary hardware is produced if none of these values exceed 4 bits.

Care must also be taken that values do not overflow their width. This is more of an issue with Handel-C than with conventional C because variables should be just wide enough to contain the largest value required (and no wider).

You cannot cast a variable or expression to a type with a different width. Use the concatenation operator to zero pad or sign extend a variable to a given width.

4.1.2 String constants

String constants are allowed in Handel-C. A string constant consists of a string of characters delimited by double quotes ("). They will be stored as a null-terminated array of characters (as in ANSI-C). String constants can contain any of the special characters listed below. Arrays and pointers can be initialized with string constants, and string constants can be assigned to pointers. If a string constant is assigned to a pointer, the storage for the string will be created implicitly.

Special characters:

`\a`	alert
`\b`	backspace
`\f`	formfeed
`\n`	newline
`\r`	carriage return
`\t`	tab
`\v`	vertical tab
`\\`	backslash
`\?`	question mark
`\'`	single quote
`\"`	double quote
`\o`*`number`*	octal number e.g. `\o77`
`\x`*`number`*	hexadecimal number e.g. `\xf3`

4.1.3 Constants

Constants may be used in expressions. Decimal constants are written as simply the number while hexadecimal constants must be prefixed with `0x` or `0X`, octal constants

must be prefixed with a zero and binary constants must be prefixed with `0b` or `0B`. For example:

```
w = 1234;         /* Decimal     */
x = 0x1234;       /* Hexadecimal */
y = 01234;        /* Octal       */
z = 0b00100110;   /* Binary      */
```

The width of a constant may be explicitly given by 'casting'. For example:

```
x = (unsigned int 3) 1;
```

Casting may be necessary where the compiler is unable to infer the width of the constant from its usage.

4.2 Logic types

The basic logic type is an `int`. It may be qualified as `signed` or `unsigned`. Integers can be manually assigned a width by the programmer or the compiler will attempt to infer a width from use.

Enumeration types (`enums`) allow you to define a specified set of values that a variable of this type may hold.

There are derived types (types that are derived from the basic types). These are arrays, pointers, `structs` bit fields, and functions. The non-type `void` enables you to declare empty parameter lists or functions that do not return a value. The `typeof` type operator allows you to reference the type of a variable.

4.2.1 int

There is only one fundamental type for variables: `int`. By default, integers are signed. The `int` type may be qualified with the `unsigned` keyword to indicate that the variable only contains positive integers or 0. For example:

```
int 5 x;
unsigned int 13 y;
```

These two lines declare two variables: a 5-bit signed integer `x` and a 13-bit non-negative integer `y`. In the second example here, the `int` keyword is optional. Thus, the following two declarations are equivalent.

```
unsigned int 6 x;
unsigned 6 x;
```

You may use the `signed` keyword to make it clear that the default type is used. The following declarations are equivalent.

```
int 5 x;
signed int 5 x;
signed 5 x;
```

The range of an 8-bit signed integer is -128 to 127 while the range of an 8-bit unsigned integer is 0 to 255 inclusive. This is because signed integers use 2's complement representation.

You may declare a number of variables of the same type and width simultaneously. For example:

```
int 17 x, y, z;
```

This declares three 17-bit wide signed integers `x`, `y` and `z`.

4.2.2 Signed | unsigned syntax

`Signed | unsigned` is declared in the same way as in ANSI-C except that Handel-C allows the width to be declared. The width may be undefined, an expression, or nothing.

For example:

- `int a;`
- `long b;`
- `unsigned int 7 c;`
- `signed undefined d;`
- `long signed int e;`

4.2.3 Supported types for porting

Handel-C provides support for porting from conventional C by allowing the types `char`, `short` and `long`. For example:

```
unsigned char w;
short y;
unsigned long z;
```

Note that these are fixed-widths in Handel-C, and implementation dependent in ANSI-C. The widths used for each of these types in Handel-C is as follows:

Type	Width
char	8 bits (signed)
short	16 bits
long	32 bits

Smaller and more efficient hardware will be produced by using variables of the smallest possible width.

4.2.4 Inferring widths

The Handel-C compiler can infer the width of variables from their usage. It is therefore not always necessary to explicitly define the width of all variables and the `undefined` keyword can be used to tell the compiler to try to infer the width of a variable. For example:

```
int 6 x;
int undefined y;

x = y;
```

In this example the variable x has been declared to be 6 bits wide and the variable y has been declared with no explicit width. The compiler can infer that y must be 6 bits wide from the assignment operation later in the program and sets the width of y to this value.

If the compiler cannot infer all the undefined widths, it will generate errors detailing which widths it could not infer.

The `undefined` keyword is optional, so the two definitions below are equivalent:

```
int x;
int undefined x;
```

Handel-C provides an extension to allow you to override this behaviour to ease porting from conventional C. This allows you to set a width for all variables that have not been assigned a specific width or declared as `undefined`.

This is done as follows:

```
set intwidth = 16;

int x;
unsigned int y;
```

This declares a 16-bit wide signed integer x and a 16-bit wide unsigned integer y. Any width may be used in the set intwidth instruction, including undefined.

You can still declare variables that must have their width inferred by using the undefined keyword. For example:

```
set clock = external "p1";
set intwidth = 27;

void main(void)
{
    unsigned x;
    unsigned undefined y;
}
```

This example declares a variable x with a width of 27 bits and a variable y that has its width inferred by the compiler. This example also illustrates that the int keyword may be omitted when declaring unsigned integers.

You may also set the default width to be undefined:

```
set intwidth = undefined;
```

4.2.5 Arrays

You can declare arrays of variables in the same way that arrays are declared in conventional C. For example:

```
int 6 x[7];
```

This declares 7 registers each of which is 6 bits wide. Accessing the variables is exactly as in conventional C. For example, to access the fifth variable in the array:

```
x[4] = 1;
```

Note that as in conventional C, the first variable has an index of 0 and the last has an index of *n*-1 where *n* is the total number of variables in the array.

When a variable is used as an array index, as is often done when using a for loop, the variable must be declared unsigned.

Example

This loop initializes all the elements in array ax to the value of index.

```
unsigned int 6 ax[7];
unsigned index;

index=0;
do
{
    ax[index] = (0 @ index);
    index++;
}
while(index <= 6);
```

Note that the width of `index` has to be adjusted in the assignment. This is because its width will be inferred to be 3, from the array dimension (the array has 7 elements, so "index" will only ever need to count as far as 6).

Multidimensional arrays

You can declare multi-dimensional arrays of variables. For example:

```
unsigned int 6 x[4][5][6];
```

This declares 4 * 5 * 6 = 120 variables each of which is 6 bits wide. Accessing the variables is as expected from conventional C. For example:

```
y = x[2][3][1];
```

Pointers to arrays

If you want to declare a pointer to the whole of an array, rather than an individual element, you must enclose the variable name and the "*" in brackets. You must also use brackets when initializing a pointer to an entire array:

```
// Declare an array
unsigned 4 MyArray [2];

// Declare a pointer to an array element
unsigned 4 *pointer_to_array_element;

// Declare a pointer to the entire array - brackets are required
unsigned 4 (*pointer_to_array) [2];

void main(void)
{
    // Initialize pointer to point to an individual array element
    pointer_to_array_element = &MyArray[0];
```

```
    // Initialize pointer to point to the entire array - brackets are
required
    (pointer_to_array) = &(MyArray);
    ...
}
```

If you wanted to view all the referenced values `MyArray` in the Watch window during simulation, you would need to type in "`(*pointer_to_array)`".

4.2.6 Array indices

When an array is declared, the index has the smallest width possible. For instance, in array[8], the index need only go up to seven and will therefore be a three bit number. If a variable is declared to represent the index, it too will be three bits.

4.2.7 struct

`struct` defines a data structure; a grouping together of variables under a single name. The format of the structure can be identified by a type name. The variable members of the structure may be of the same or different types. Once a structure has been declared, its type name can be used to define other structures of the same type. Structure members may be accessed individually using the construct

struct_Name.member_Name

Syntax

A structure type is declared using the format

```
struct [type_Name]
{
    member-list
} [instance_Name {,instance_Name}];
```

member-list is a list of variable definitions terminated by semi-colons.

The use of *instance_Names* declares variables of that structure type. Alternatively, you may declare variables as follows:

```
struct type_Name instance_Name;
```

Storage

- Structures may be passed through channels and signals.
- Structures may be stored in internal memory elements.
- Structures cannot be stored in off-chip RAMs.

If a structure contains a memory element, a channel, or a signal, it cannot be stored in another memory element, it cannot be passed to a function "by value", it cannot be assigned to and it cannot be passed through a channel or a signal.

If a structure contains a memory element, it cannot be assigned (or assigned to) another structure, as the assignment cannot be performed in a single clock cycle.

Whole structures may not be sent directly to interfaces.

Example

```
struct human // Declare human struct type
{
    unsigned int 8 age; // Declare member types
    int 1 sex;
    char name[25];
}; // Define human type

struct human sister;
sister.age = 25;
```

Initialization

You can use a list initializer to initialize static or const structures or structures with global scope. List initializers may be flat or structured.

```
struct Boris
{
    int 12 v[3];
    int 8 a, b;
};
static struct Boris b = {{1, 2, 3}, 4, 5};
```

4.2.8 enum

`enum` specifies a list of constant integer values, for example:

```
enum weekdays {MON, TUES, WED, THURS, FRI};
```

The first name (in this case `MON`) has a value of 0, the next 1, and so on, unless explicit values are specified. If not all values are specified, values increment from the last specified value.

If you do not specify a width for the `enum`, the program must contain information from which the compiler can infer the width.

You can declare variables of a specified `enum` type. They are effectively equivalent to `int undefined` or `unsigned undefined`. The signedness is inferred from use.

To specify `enum` values

```
enum weekdays {MON = 9, TUES, WED, THURS, FRI};
```

To specify the width of an enum

```
enum weekdays {MON = (unsigned 4)9, TUES, WED, THURS, FRI};
```

To declare a variable of type `enum`

```
enum weekdays x;
```

To assign `enum` values to a variable

```
static int x = MON;
```

Example

The example below illustrates how to infer the width of an `enum`. The cast ensures the enumerated variable has a width associated with it.

```
set clock = external "P1";
typedef enum
{
    A,
    B,
    C = 43,
    D
} En;

void main(void)
{
    En num;
    int undefined result;

    num = (int 7)D;

    result = num;
}
```

4.2.9 Bit fields

A bit field is a type of structure member consisting of a specified number of bits. The length of each field is separated from the field name by a colon (:). Each element can be accessed independently. Since Handel-C allows you to specify the width of integers in bits, a bit field is merely another way of specifying a standard structure. In ANSI-C, bit fields are made up of words, and only the specified bits are accessed, the rest are padded. Padding in ANSI-C is implementation dependent. There is no padding in Handel-C, so nothing can be assumed about it.

Syntax

```
struct [tag_name]
{
    field_Type field_Name: field_Width
    ...
} [instance_names] ;
```

Example

This example defines an identical array of flags as a structure and as a bit field.

```
struct structure
{
    unsigned int 1 LED;
    unsigned int 1 value;
    unsigned int 1 state;
}outputs;

struct bitfield
{
    unsigned int LED : 1;
    unsigned int value : 1;
    unsigned int state : 1;
}signals;
```

4.3 Pointers

A pointer declaration consists of *, the name of the pointer and the type of the variable that it points to.

```
type *Name
```

Pointers are used to point to variables in conjunction with the unary operator &, which gives the address of an object. To set a pointer to point to a variable, you assign the address of the variable to the pointer. For example

```
int 8 *ptr;   //declare a pointer to an int 8
int 8 object, x;
object = 6;
x = 10;
ptr = &object; //assigns address of object to ptr
x = *ptr;      // x is now 6
*ptr = 12;     //object is now 12
```

The behaviour of uninitialized pointers is undefined. De-referencing an uninitialized pointer during simulation will result in a run-time error, after which the simulator will terminate.

Casting pointers

In Handel-C, you may only cast void pointers (`void * pointerName`) to a different type. All other pointers may only be cast to change the sign of an object pointed to, and whether it is const or volatile. These restrictions are the standard casting restrictions in Handel-C.

You can change a void pointer's type by casting, assignment or comparison. `Void *` must have a consistent type so:

```
void *p;
int 6 *s;
int 7 *t;
p = s;
p = t; //invalid
```

Pointer arithmetic

You cannot perform arithmetic on a void pointer because the size of the object being pointed to is not known.

- Valid pointer operations are:
- Assign a pointer to another pointer of the same type
- Add a pointer and an integer
- Subtract an integer from a pointer
- Subtract or compare (using <, <=, > or >=) a pointer to an array or memory member with another pointer to a member of the same array or memory
- Compare two pointers for equality (using != or ==)
- Assign or compare a pointer to NULL

The result of subtracting or comparing pointers to members of different arrays or memories or to other objects is undefined.

The behaviour of arithmetic on pointers that moves the pointer beyond the extent of the object is undefined. An exception is that an address one element beyond an array or memory (at the high end) is valid, but it is not valid to dereference a pointer at such an address (the behaviour of the dereference would be undefined). This "one-beyond" address is useful for loops.

Examples

In the examples below, `p` and `q` can point to any part of `Single` or an element of `Array`, `AnotherArray` or `Memory`.

```
int undefined i;
int 4 Single, Array [10], AnotherArray [20];
ram int 4 Memory [10];
int 4 * p, * q;
unsigned int 1 test;

p = & Single;
p += 2; // undefined behaviour (invalid address)
p = & Single; ++ p; // defined (valid address), but ...
* p = 0; // ... undefined behaviour
p = & (Array [4]);
p += 2; // now, p = & (Array [6])
p = Array; q = & (Array [4]);
i = q - p; // meaningful; now, i = 4;
test = (p < q); // meaningful (true in this case)
test = (p == q); // meaningful (false in this case)
p = Array; q = AnotherArray;
i = q - p; // undefined behaviour
test = (p < q); // undefined behaviour
test = (p == q); // meaningful (false for pointers into different objects)
```

4.3.1 Pointers and addresses

Pointers in Handel-C are similar to those in conventional C. They provide the address of a variable or a piece of code. This enables you to access variables by reference rather than by value.

The indirection operator (*) is the same as it is in ANSI-C. It is used to de-reference pointers (i.e. to access objects pointed to by pointers).

The "address of" operator (&) works as it does in ANSI-C.

4.3.2 Pointers to functions

If you point to code (a function), the address operator is optional. The syntax is

```
returnType (*pointerName)(parameter list);
```

The parentheses at the end of the declaration declare the pointer to be a pointer to a function. The * before the *`pointerName`* declares it to be a pointer declaration.

There is the standard C type ambiguity between the declaration of a function returning a pointer and a pointer to a function. To ensure that * is associated with the pointer name rather than the return type, you need to use parentheses

```
int 8 * functionName(); //function returning pointer
```

and

```
int 8 (* pointerName)(); //pointer to function
```

4.3.3 Pointers to interfaces

When declaring pointers to interfaces, you must ensure that you declare a pointer to an interface sort and then assign a defined interface to it (much as when you declare a pointer to a function). You cannot combine the definition of an object with the declaration of a pointer to it.

The members of the interface must have the same name in the declaration of the pointer type as in the definition of the interface object which you assign the pointer to.

Example

```
//declaration of pointer to interface of sort bus_out
interface bus_out() *p(int 2 x);
interface bus_out() b(int 2 x=y); //interface definition
p=&b;  // p now points to b
```

4.3.4 Structure pointers

The structure pointer operator (->) can be used, as in ANSI-C. It is used to access the members of a structure, when the structure is referenced through a pointer.

```
struct S
{
    int 18 a, b;
} s, *sp;
sp = &s;
s.a = 26;
sp->b = sp->a;
```

The last line accesses the member variables of structure `s` through pointer `sp`. Because the pointer is being used to access the structure, the -> operator is used to refer to the member variables.

```
sp->a = (*sp).q
```

You can cast structure pointers between structures with the same member types and names. For example:

```
struct S1
{
    int 6 x;
} st1;

struct S2
{
    int 6 x;
} st2;

set clock = external;
void main (void)
{
    int r;
    struct S1 *structPtr1;
    struct S2 *structPtr2;
    structPtr1 = &st1;
    structPtr2 = (struct S2 *)structPtr1;
    structPtr2->x = 7;
    r = st1.x; //r = 7
}
```

4.3.5 address and indirection operators

The indirection operator * is the same as it is in ANSI-C. It is used to de-reference pointers (i.e. to access objects pointed to by pointers).

The address operator (&) works as it does in ANSI-C.

The following can also be used: pointers to arrays, pointers to channels, pointers to signals, pointers to memory elements, pointers to structures, pointers to pointers, arrays of pointers.

Example: pointer assignment

```
unsigned char cha, chb, *chp;

chp = &cha;
cha = 90;

chb = *chp;
chp = &chb;
```

The first line declares two `unsigned` variables (`cha` and `chb`), and a pointer to an `unsigned` (`chp`). The second line assigns the address of `cha` to pointer `chp`. In other words, pointer `chp` now points to variable `cha`. The third line simply assigns a value to

cha. The fourth line dereferences pointer chp, to access what it's pointing to, which is cha. In other words, chb is assigned the value of the object pointed to by chp. The last line assigns the address of chb to pointer chp. In other words, pointer chp now points to variable chb.

Example: pointer to pointer assignment

```
struct S
{
    int 6 a, b;
} s1, s2, *sp, **spp;

sp = &s1;
spp = &sp;
s2 = **spp;
```

This declares two variables of type struct S (s1 and s2), a pointer to a variable of this type (sp), and a pointer to a pointer to a variable of this type (spp). The next line assigns the address of structure s1 to pointer sp (pointer sp to point to structure s1). The following line assigns the address of pointer sp to pointer spp (pointer spp to point to pointer sp). The last line dereferences pointer spp twice, and it assigns the dereferenced value, which is s1, to structure s2 (i.e. s2 now equals s1).

4.4 Architectural types

The architectural types are:

- channels (used to communicate between parallel processes)
- interfaces (used to connect to pins or provide signals to communicate with external code)
- memories (rom, ram, wom and mpram)
- signal (declares a wire).

The type clarifier < > has been provided to help clarify the definitions of memories, channels and signals.

4.5 Channels

Handel-C provides channels for communicating between branches of code executing in parallel. One branch writes to a channel and a second branch reads from it.

Channels are declared with the chan keyword. For example:

```
chan int 7 link;
```

The width and type of data sent down the channel must be of the same width and type as the channel. The width and type of a channel can sometimes be inferred by the Handel-C compiler, if they are not explicitly declared. The channel can be an entry in an array of channels, or be pointed to by a channel pointer.

If you want to select the first channel that is ready to communicate from a list of channels, use the `prialt` statement.

If you wish to convert the channel into a FIFO, use the `fifolength` specification. This creates a FIFO with the specified number of data stores of the same width as the channel.

If you are simulating code, you may use `chanin` and `chanout` to specify interfaces to the simulator. These do not represent architectural channels but can be addressed in a similar way.

Syntax

`chan [ ` ***`logicType`*** ` ] ` ***`Name`*** ` [with ` ***`specifications`***`];`

Reading from a channel

`Channel` `?` *`Variable`*`;`

This assigns the value read from the channel to the variable. It may also be read to a signal, an array element, RAM element or WOM element.

Writing to a channel

`Channel` `!` *`Expression`*`;`

This writes the value of the expression to the channel. *`Expression`* may be any expression of the correct type.

Example

```
set clock = external;
void main(void)
{
    unsigned 8 Res;
    chan Bill;

    par
    {
        Bill ! 23;
        Bill ? Res;
    }
}
```

4.5.1 FIFO code example

```
chan unsigned 8 ch with { fifolength=2 };
{
  unsigned 8 a,b,c,d ;

  ch!1; // FIFO becomes <1>
  ch!2; // FIFO becomes <1,2>
  ch?a; // FIFO becomes <2>, a becomes 1
  ch!3; // FIFO becomes <2,3>
  ch?b; // FIFO becomes <3>, b becomes 2
  ch?c; // FIFO becomes empty, c becomes 3
  ch!4; // FIFO becomes <4>
  ch!5; // FIFO becomes <4,5>
  ch?d; // FIFO becomes <5>, d becomes 4
  ch!6; // FIFO becomes <5,6>
}
```

4.5.2 Arrays of channels

Handel-C allows arrays of channels to be declared. For example:

```
chan unsigned int 5 x[6];
```

This is equivalent to declaring 6 channels each of which is 5 bits wide. A channel can be accessed by specifying its index. As with variable arrays, the index for the nth element is n-1. For example:

```
x[4] ! 3; // Output 3 on channel x[4]
x[3] ? y; // Input to y from channel x[3]
```

It is also possible to declare multi-dimensional arrays of channels. For example:

```
chan unsigned int 6 x[4][5][6];
```

This declares 4 * 5 * 6 = 120 channels each of which is 6 bits wide. Accessing the channels is similar to accessing arrays in conventional C. For example:

```
x[2][3][1] ! 4; // Output 4 on channel
```

4.5.3 Restrictions on channel accesses

No two statements may simultaneously write to or simultaneously read from a single channel.

```
par
{
    out ! 3 // Undefined: simultaneous send to a channel
    out ! 4
}
```

This code will give an undefined result, as it attempts to write simultaneously to a single channel. Similarly, the following code will not work because an attempt is made to read simultaneously from the same channel:

```
par
{
    in ? x; // Undefined: simultaneous receive from a channel
    in ? y;
}
```

Your code should not rely on the perceived behaviour of multiple simultaneous reads and writes, in either simulation or hardware.

You can detect parallel accesses to channel during simulation using the Detection of simultaneous channel reads/writes option on the Compiler tab in Project Settings, or by using the `-S+parchan` option in the command line compiler.

Simultaneous channel access concealed within prialt

The `prialt` construct negotiates the readiness of the remote (i.e. non-prialt) end of channel. It does not resolve conflicts at the local (i.e. prialt) end of the channel. The programmer must still avoid simultaneous channel accesses, even if the send or receive statements are inside a `prialt` statement.

Examples:

```
int 4 x, y, z;
chan <int 4> ch1, ch2;
unsigned int 1 thing;

// Code that affects thing

par {
   ch2 ! x;
   prialt
   {
   case ch1 ! y:
      break;
   case ch2 ! y:
      // Undefined: simultaneous send
      break;
   }
   if (thing)
      ch1 ? z;
   else
      ch2 ? z;
}
```

If `thing` is false, then channel `ch2` is the only channel that becomes ready to receive, so the prialt tries to send `y` over `ch2` simultaneously with the statement sending `x` over `ch2`, resulting in an illegal simultaneous access.

There is a conflict even when `thing` is true, as `ch2` undergoes readiness negotiations within the `prialt` statement and this also requires access to the channel.

Restrictions on channels accesses between clock domains

If you have channels communicating between clock domains, all writes to a channel must take place within a single clock domain, and all reads must take place within a single clock domain.

For more information on using channels to communicate between clock domains, see ***Channels communicating between clock domains*** (see page 169)

4.5.4 Timing and latency in FIFOs

Note that if `fifolength` is a power of 2, the channel will be implemented in a different way to when it is not, in order to save memory.

Channels with FIFO sizes of a power of 2 may have greater latency.

The latency of channels is dependent on the target architecture and the way the code has been implemented within it.

4.6 Interfaces: overview

All interfaces, except for external (foreign code or off-chip) RAMs are declared with the `interface` keyword in Handel-C. Interfaces are used to communicate with:

- external devices
- external logic, such as other Handel-C programs, programs written in VHDL etc.

You can communicate between blocks of internal logic using channels

The interface definition is in two parts:

- an interface sort: the name of the black box or primitive that the interface connects to
- an instance name: the name of the instance of the interface sort in Handel-C

Interface definitions may be split into declarations and definitions. You must use a declaration if you want to define multiple instances of the same interface sort, or to use forward references.

The declaration gives the sort name and port names and types associated with that interface sort.

The definition gives the instance name, object specifications and the data transmitted for a single instance of the interface sort.

Only `signed` and `unsigned` types may be passed over interfaces.

Your license may not allow you to use interfaces. If this is the case you can only interface to external devices using macros provided in any Celoxica libraries you have licenses for, such as PAL.

4.6.1 Interface declaration

You need to use an interface declaration if you want to define multiple instances of an interface sort, or to use forward references. If you only want a single instance of an interface sort, you only need to use an interface definition.

Interfaces of pre-defined sorts do not need to be declared.

The general format of the interface declaration is:

```
interface Sort (ports_in_to_Handel-C)
   (ports_out_from_Handel-C);
```

Sort	user-defined name or predefined interface sort
ports_in_to_Handel-C	Optional. One or more prototypes of ports bringing data into the Handel-C code.
ports_out_from_Handel-C	Optional. One or more prototypes of ports sending data from the Handel-C code.

A port prototype consists of the port type, and the port name. At least one port (whether to Handel-C or from Handel-C) must be declared. Port declarations are delimited by commas. For example:

```
interface MyInterface (int 5 InPort)
   (int 4 OutPort1, int 4 OutPort2);
```

The name of each port in a `port_in` or `port_out` interface must be different, as they will all be built to the top level of the design.

Once you have declared an interface sort, you can define multiple instances of that sort. The interface definition creates a named instance of the interface sort, assigns data to be transmitted to the output ports, and may also specify properties using interface specifications. You cannot use interface specifications in interface declarations, only in interface definitions.

You can declare pointers to an interface declaration and then assign a defined interface to the pointer.

4.6.2 Interface definition

A Handel-C `interface` definition consists of an interface sort, an instance name and data ports, together with information about each port.

The definition defines a single instance of an interface sort. If you want to define multiple instances, or use forward references to the interface, declare the interface, and then make multiple definitions of that interface sort. (You do not need to declare interfaces of predefined sorts.)

The general format of an interface definition is:

```
interface Sort (ports_in_to_Handel-C)
     InstanceName (ports_out_from_Handel-C )
     with {GeneralSpecs};
```

Sort	Pre-defined interface sort, or used-defined sort. (This should match the sort in the interface declaration, if you are using one.)
ports_in_to_Handel-C	Definitions of one or more ports bringing data into the Handel-C code. (Port definitions are described below.)
InstanceName	User-defined identifier for that instance of the interface. (You can define any number of instances of an interface sort, if you make a declaration of the interface sort.)
ports_out_from_Handel-C	Definitions of one or more ports sending data from the Handel-C code. Each output port should be assigned an expression. The value of the expression will be connected to that port.
GeneralSpecs	Handel-C interface specifications. These specify hardware details of the interface, such as chip pin numbers or are used to specify an external simulator using the extlib directive. Interface specifications apply to all ports in the interface. You can also assign specifications to individual ports.

Port definitions

If the interface has been previously declared, the port definitions must be prototyped in their interface declaration, and must have the same types as those in the prototype. The declaration must have at least one port into Handel-C or from Handel-C. Port definitions are delimited by commas. Each port definition consists of:

- the data type that uses it (either defined or inferred from its first use). Only `signed` and `unsigned` types may be passed over interfaces.
- a port name
- port specifications (optional). The port specifications are enclosed in a set of braces {...} and delimited by commas.

Example

```
interface Sort_A (int 4 inPort1, int 4 inPort2)
   interfaceName (unsigned outPort = x)
```

4.6.3 Example interface to external code

This example shows an interface declaration used to connect to a piece of foreign code, and the definition that uses this declaration.

```
set clock = external "D17";
set family = XilinxVirtex;
set part = "V1000BG560-4";

// Interface declaration
interface ttl7446(unsigned 7 segments, unsigned 1 rbon)
    (unsigned 1 ltn, unsigned 1 rbin, unsigned 4 digit,
    unsigned 1 bin);

unsigned 1 ltnVal;
unsigned 1 rbinVal;
unsigned 1 binVal;
unsigned 4 digitVal;

// Interface definition
interface ttl7446(unsigned 7 segments, unsigned 1 rbon)
    decode(unsigned 1 ltn=ltnVal, unsigned 1 rbin=rbinVal,
        unsigned 4 digit=digitVal, unsigned 1 bin=binVal)
        with {extlib="PluginModelSim.dll",
        extinst="decode; model=ttl7446_wrapper; delay=1"};
```

This declares an interface of sort `ttl7446`. The inputs from the interface to the Handel-C design are `segments` and `rbon`. The interface would therefore connect to a black box named `ttl7746` with ports `segments`, `rbon`, `ltn`, `rbin`, `digit`, and `bin`.

The instance of the interface is `decode`. The instance specifies the data going into the ports `ltn`, `rbin`, `digit`, and `bin` and connects to a plugin, `PluginModelSim.dll`, for simulation.

If you did not want to use forward references to the interface, and only wanted to define a single instance of the interface sort `ttl7446`, you would not need to declare the interface. (The interface definition would be exactly the same as that shown above.)

4.6.4 Interface specifications

Predefined bus interface specs:		**Default:**
`data`	list the pins used for transferring data, MSB to LSB	None

`speed`	set buffer speed (output)	2: Actel ProASIC/ ProASIC+ 1: others
`pull`	set pull-up or pull-down for bus pins	None
`infile`	set file source for input bus data	None
`outfile`	set file destination for output bus data	None

All interface specs:		**Default:**
`base`	specify display base for variables in debugger	10
`bind`	bind component to work library	0
`busformat`	text format of exported wires in EDIF netlist	"B_I"
`data`	list the pins used for transferring data, MSB to LSB	None
`dci`	apply Digital Controlled Impedance to buses (Xilinx only)	0 (No)
`extlib`	specify external plugin for simulator	None
`extfunc`	specify external simulator function for this port	`PlugInSet` or `PlugInGet`
`extpath`	specify any direct logic (combinational logic) connections to another port	None
`extinst`	specify connection to external code	None
`intime`	maximum allowable time between a port and the sequential elements it drives (in ns)	None
`outtime`	maximum allowable time between a port and the sequential elements it is driven from (in ns)	None
`properties`	parameterize instantiations of external black boxes	None
`sc_type`	specify type of port in `port_in`, `port_out` or generic interface for SystemC	`bool` for 1 bit ports, `sc_uint` otherwise
`standard`	specify I/O standard (electrical characteristics) to use on port(s) in question	`LVCMOS33` for Actel ProASIC/ProASIC+, `LVTTL` for others
`strength`	specify drive strength (in mA) for output buses	Standard dependent

`vhdl_type`	specify type of port in `port_in`, `port_out` or generic interface in VHDL	`std_logic` for 1 bit ports, `unsigned` otherwise
`warn`	disable some compiler warnings	1 (No)

4.7 RAMs and ROMs

RAMs and ROMs may be built from the logic provided in the FPGA/PLD using the `ram` and `rom` keywords.

For example:

```
ram int 6 a[43];
static rom int 16 b[4] = { 23, 46, 69, 92 };
```

This example constructs a RAM consisting of 43 entries each of which is 6 bits wide and a ROM consisting of 4 entries each of which is 16 bits wide.

ROMs must be declared as static or global. If you declare a static ROM in a macro procedure, each call to the macro creates a separate version of the ROM. RAMs can be declared as static, global or auto (i.e. non-static).

All RAMs and ROMs must be declared as arrays, so to declare a RAM that holds one 4 bit integer, you must declare it as an array with a dimension of 1.

```
ram int 4 ramname[1];
```

RAMs and ROMs may only have one entry accessed in any clock cycle.

4.7.1 Initialization

You can only initialize ROMs or RAMs if they are static, or have global scope. For example, a global ROM could be initialized as shown below:

```
rom int 16 b[4] = { 23, 46, 69, 92 } with {block = 1};
```

The ROM is initialized with the constants given in the following list in the same way as an array would be initialized in C. In this example, the ROM entries are given the following values:

ROM entry	Value
b[0]	23
b[1]	46
b[2]	69
b[3]	92

4.7.2 Inferring size from use

The Handel-C compiler can also infer the widths, types and the number of entries in RAMs and ROMs from their usage. Thus, it is not always necessary to explicitly declare these attributes. For example:

```
ram int undefined a[123];
ram int 6 b[];
ram c[43];
ram d[];
```

4.7.3 Accessing RAMs and ROMs

RAMs and ROMs are accessed in the same way as arrays. For example:

```
ram int 6 b[56];

b[7] = 4;
```

This sets the eighth entry of the RAM to the value 4. Note that as in conventional C, the first entry in the memory has an index of 0 and the last has an index of n-1 where n is the total number of entries in the memory.

4.7.4 Differences between RAMs and arrays

RAMs differ from arrays in that an array is equivalent to declaring a number of variables. Each entry in an array may be used exactly like an individual variable, with as many reads, and as many writes to a different element in the array as required within a clock cycle. RAMs, however, are normally more efficient to implement in terms of hardware resources than arrays, but they only allow one location to be accessed in any one clock cycle. Therefore, you should use an array when you wish to access the elements more than once in parallel and you should use a RAM when you need efficiency.

4.7.5 RAM and ROM support on different devices

Creating internal RAMs can only be done if the target device supports on-chip RAMs. Most devices currently targeted by Handel-C do so (e.g. Altera Flex 10K, APEX, APEXII, Mercury, Stratix and Cyclone, Xilinx Spartan series devices and Virtex series devices).

No Actel families support ROMs. ProASIC and ProASIC+ devices support RAMs, but these may not be initialized.

4.7.6 Multidimensional memory arrays

You can create simple multi-dimensional arrays of memory using the `ram`, `rom` and `wom` keywords. The definitions can be made clearer by using the optional disambiguator <>.

Syntax

```
ram | rom | wom logicType entry_width
        Name[[const_expression]] {[[const_expression]]}
        [= {initialization}];
```

Possible logic types are `ints`, `structs`, pointers and arrays.

The last constant expression is the index for the RAM. The other indices give the number of copies of that type of RAM.

Example

```
ram <int 6> a[15][43];
static rom <int 16> b[4][2][2] =
    {{{1, 2},
       {3, 4}
      },
       {{5, 6},
        {7, 8}
       },
       {{9, 10},
        {11, 12}
       },
       {{13, 14},
        {15, 16}
       }
    };
```

This example constructs 15 RAMs, each consisting of 43 entries of 6 bits wide and 4 * 2 ROMs, each consisting of 2 entries of 16 bits wide. The ROM is initialized with the constants in the following list in the same way as a multidimensional array would be

initialized in C. The last index (that of the RAM entry) changes fastest. In this example, the ROM entries are given the following values:

ROM entry	Value	ROM entry	Value
b[0][0][0]	1	b[0][0][1]	2
b[0][1][0]	3	b[0][1][1]	4
b[1][0][0]	5	b[1][0][1]	6
b[1][1][0]	7	b[1][1][1]	8
b[2][0][0]	9	b[2][0][1]	10
b[2][1][0]	11	b[2][1][1]	12
b[3][0][0]	13	b[3][0][1]	14
b[3][1][0]	15	b[3][1][1]	16

Because of their architecture, RAMs and ROMs are restricted to performing operations sequentially. Only one element of a RAM or ROM may be addressed in any given clock cycle and, as a result, familiar looking statements are often disallowed. For example:

```
ram <unsigned int 8> x[4];
x[1] = x[3] + 1;
```

This code is inadvisable because the assignment attempts to read from the third element of `x` in the same cycle as it writes to the first element.

In a multi-dimensional array, you can access separate elements of the arrays, so long as you are not accessing the same RAM. For example:

`x[2][1]=x[3][0]` is valid

`x[2][1]=x[2][0]` is invalid

Note that arrays of variables do not have these restrictions but may require substantially more hardware to implement than RAMs depending on the target architecture.

4.8 mpram (multi-ported RAMs)

You can create multiple-ported RAMs (MPRAMs) by constructing something similar to an ANSI-C union. You must use the `mpram` keyword.

`mprams` can be used to connect two independent code blocks. The clock of the `mpram` port is taken from the function in which it is used.

The normal declaration of a MPRAM would be to create a dual-ported RAM by declaring two ports of equal width:

- for Actel devices, one port must be read-only, and one write-only.

- for Altera ApexII and Mercury devices, both ports can be bi-directional. For Cyclone and Stratix devices this depends on the type of memory used. For other Altera families, one port would be read-only and one write-only
- Altera Mercury devices can have up to four ports. You can have (one or two write ports AND one or two read ports) OR two read/write ports. Depending on how you have configured the port, you can have up to four simultaneous accesses of the same block of memory.
- for Virtex and SpartanII devices, both ports would be read/write for block RAM, and for LUT RAM, one port would be read/write and one read-only. Spartan and SpartanXL devices only have distributed (LUT) RAM.

You can use `mpram` ports of different widths for certain devices.

The `mpram` construct allows the declaration of any number of ports. Your only restriction is the target hardware.

You can apply clock specifications to the whole MPRAM, or to individual ports. MPRAM write ports will be synchronous and read ports will be asynchronous by default, if the target hardware allows it. For example, Stratix memories are fully synchronous and do not allow an asynchronous read port.

You can create synchronous read ports explicitly by using clock position specifications (`rclkpos` and `clkpulselen`), and asynchronous write ports by using write-enable specifications (`westart`, `welength` or `wegate`). However, you cannot have an asynchronous write port and a synchronous read port, since this would violate Handel-C's timing semantics.

Syntax

```
mpram MPRAM_name
{
    ram_Type variable_Type RAM_Name[size]
        [with {ClockPosition/WriteEnableSpecs = value}];
    ram_Type variable_Type RAM_Name[size]
        [with {ClockPosition/WriteEnableSpecs = value}];
};
```

Examples

In the example below, the first MPRAM is a bi-directional dual-port RAM, with clock specifications applied to the whole MPRAM. The second is a simple dual-port RAM, with different clock specifications applied to each port.

```
set clock = external_divide "C1" 4;

mpram
{
    ram unsigned 4 Port1[4];
    ram unsigned 4 Port2[4];
} TMax with {wclkpos = {2}, rclkpos = {2.5}, clkpulselen = 1};

mpram
{
    wom unsigned 4 WritePort[4] with {wclkpos = {2}, clkpulselen = 1};
    rom unsigned 4 ReadPort[4] with {rclkpos = {2.5}, clkpulselen = 1};
} SMax;
```

4.8.1 Initialization of mprams

The first member of the `mpram` can be initialized.

```
static mpram Fred
{
    ram <unsigned 8> ReadWrite[256];  // Read/write port
    rom <unsigned 8> Read[256];       // Read only port
} Mary ={10,11,12,13};
```

This would have the same effect as

```
Mary.ReadWrite[0]=10;
Mary.ReadWrite[1]=11;
Mary.ReadWrite[2]=12;
Mary.ReadWrite[3]=13;
```

The other elements of `Fred.ReadWrite` will be initialized as zero (since `Mary` is `static`). In this case, since `Fred.Read` is the same size as `Fred.ReadWrite`, elements 0 – 3 of `Fred.Read` would be initialized with the same values.

4.8.2 Mapping of different width mpram ports

If the ports of the `mpram` are of different widths, they will be mapped onto each other according to the specifications of the chip you are using. If the ports used are of different widths, the widths should have values of 2^n.

Different width ports are available for Xilinx Virtex and Spartan-II, Spartan-IIE and Spartan-3 devices and Altera Apex II, Stratix and Cyclone devices. They are not available with other Altera devices or Actel devices.

Xilinx bit mapping

To find the bits that an array element occupies in a Xilinx Virtex or Spartan RAM, you can use the formula
RAM array `ram y Name[a]` will have a start bit of (*y* * (*a*+1)) - 1 and an end bit of *y* * *a*.

Xilinx mapping is little-endian. This means that the address points to the LSB.

The bits between the declarations of RAM are mapped directly across, so that bit 27 in one declaration will have the same value as bit 27 in another declaration, even though the bits may be in different array elements in the different declarations.

```
mpram Joan
{
    ram <unsigned 4> ReadWrite[256]; // Read/write port
    rom <unsigned 8> Read[256];      // Read only port
};
```

`Joan.ReadWrite[100]` will run from 400 to 403.

`Joan.Read[100]` will run from 800 to 807.

`Joan.Read[50]` will run from 400 to 407.

`Joan.ReadWrite[100]` is equivalent to `Joan.Read[50][0:3]`.

ApexII bit mapping

To find the bits that an array element occupies in an ApexII RAM, you can use the formula
RAM array `ram y Name[a]` will have a start bit of (*y* * (*a*+1)) - 1 and an end bit of *y* * *a*.

ApexII mapping is little-endian. This means that the address points to the LSB.

The bits between the declarations of RAM are mapped directly across, so that bit 27 in one declaration will have the same value as bit 27 in another declaration, even though the bits may be in different array elements in the different declarations.

```
mpram Joan
{
    ram <unsigned 4> ReadWrite[256];  // Read/write port
    rom <unsigned 8> Read[256];       // Read only port
};
```

`Joan.ReadWrite[100]` will run from 400 to 403.

`Joan.Read[100]` will run from 800 to 807.

`Joan.Read[50]` will run from 400 to 407.

`Joan.ReadWrite[100]` is equivalent to `Joan.Read[50][0:3]`.

4.8.3 mprams example

Using an `mpram` to communicate between two independent logic blocks:

File 1:

```
mpram Fred
{
    ram <unsigned 8> ReadWrite[256];  // Read/write port
    rom <unsigned 8> Read[256];       // Read only port
};

mpram Fred Joan ;  // Declare Joan as an mpram like Fred

set clock = internal "F8M";

void main(void)
{
    unsigned 8 data;

    Joan.ReadWrite[7] = data;
}
```

File 2:

```
mpram Fred
{
    ram <unsigned 8> ReadWrite[256];  // Read/write port
    rom <unsigned 8> Read[256];       // Read only port
};

extern mpram Fred Joan;
set clock = external "P2";

void main(void)
{
    unsigned 8 data;
    data= Joan.Read[7];
}
```

4.9 WOM (write-only memory)

You can declare a write-only memory using the keyword `wom`. The only use of a write-only memory would be to declare an element within a multi-ported RAM. Since `wom`s only exist inside multi-port rams, it is illegal to declare one outside an `mpram` declaration.

Syntax

```
wom  variable_Type variable_Size WOM_Name[dimension] =
   initialize_Values [ with {specs}]
```

Example

```
mpram connect
{
    wom <unsigned 8> Writeonly[256]; // Write only port
    rom <unsigned 8> Read[256];      // Read only port
}
```

4.10 sema

Handel-C provides semaphores for protecting critical areas of code. Semaphores are declared with the `sema` keyword. For example:

```
sema RAMguard;
```

Semaphores have no type or width associated with them. They cannot be assigned to or have their value assigned to anything else. You can only access semaphores through the `trysema(`***`semaphore`***`)` expression and `releasesema(`***`semaphore`***`)` statement. `trysema` tests to see if the semaphore is currently taken. If it is not, it takes the semaphore and returns one. If it is taken, it returns zero. `releasesema` releases the semaphore. After you have taken a semaphore, you should ensure that you release it cleanly once you have left the critical area.

Semaphores may be included in structures. They cannot be passed to directly to functions, over channels or interfaces. They may be passed to functions or channels by reference.

Syntax

```
sema Name
```

Example

```
inline void critRAMaccess(sema *RAMsema, ram int 8
    (*danger)[4], unsigned count)
{
    int 8 x;
    // wait till you've got the // RAM
    while(trysema(*RAMsema)==0) delay;
    x= (*danger)[count];
    releasesema(*RAMsema);
}
```

4.11 signal

A signal is an object that takes on the value assigned to it but only for that clock cycle. The value is assigned at the start of the clock cycle and can be read back during the same clock cycle. At all other times the signal takes on its initialization value. The optional disambiguator <> can be used to clarify complex signal definitions.

If a signal is assigned to when you are debugging code, values shown in the **Watch** and **Variables** windows are updated immediately, rather than at the end of the clock cycle (step).

Signals represent wires in hardware.

Syntax

```
signal [<type data-width>] signal_Name;
```

Example

```
int 15 a, b;
signal <int> sig;

a = 7;
par
{
    sig = a;
    b = sig;
}
```

`sig` is assigned to and read from in the same clock cycle, so `b` is assigned the value of `a`.

Since the signal only holds the value assigned to it for a single clock cycle, if it is read from just before or just after it is assigned to, you get its initial value. For example:

```
int 15 a, b;
static signal <int> sig = 690;

a = 7;
par
{
    sig = a;
    b = sig;
}
a = sig;
```

Here, `b` is assigned the value of `a` through the signal, as before. Since there is a clock cycle before the last line, `a` is finally assigned the signal's initial value of 690.

4.12 Storage class specifiers

Storage class specifiers define how variables are accessed.

`extern` and `static` are used within functions to allocate storage. `static` gives the declared objects static storage class, and `extern` specifies that the variable is defined elsewhere. For compatibility with ANSI-C, the specifiers `auto` and `register` can be used but have no effect.

The expansion of a function is defined by the specifier `inline`.

The `typedef` specifier does not reserve storage, but allows you to declare new names for existing types.

4.12.1 auto

`auto` defines a local automatic variable. In Handel-C, all local variables default to `auto`. You cannot initialize an `auto` variable, but must assign it a value. The initialization status of auto variables is undefined.

Example

```
set clock = external "P1";

void main (void)
{
    auto 8 pig;
    pig = 15;
}
```

4.12.2 extern (external variables)

`extern` declares a variable that is external to all functions; the variable may be accessed by name from any function.

External variables must be defined exactly once outside any function, and declared in each function that wants to access them. The declaration may be an explicit extern , or else be implicit from the context (if the variable has been defined outside a function without static).

If the variable is used in multiple source files, it is good practice to collect all the extern declarations in a header file, included at the top of each source file using the #include *headerFileName* directive.

You may use `extern "`*`language`*`"` to access variables in C or C++ files.

You cannot access the same variable from different clock domains.

Example

```
extern int 16 global_fish;
int global_frog = 1234;

main()
{
  global_fish =  global_frog;
  …
}
```

Syntax

extern *variable declaration*;

4.13 extern language construct

The `extern "`*`language`*`"` construct allows you to declare that names used in Handel-C code have ANSI-C or C++ linkage.

- For ANSI-C functions, use `extern "C"`
- For C++ functions, use `extern "C++"`

These functions can only be compiled for simulations targeting the simulator. They may not be used in targeting devices.

`extern "C"` and `extern "C++"` functions have the same timing as Handel-C functions. For example, a `printf()` function would take at least one clock cycle, even if the return value is ignored, and a C function with a body that takes 0 clock cycles and a `void` return type would not take any clock cycles.

Examples

```
extern "C" int printf(const char *format, ...);
```

declares `printf()` with C linkage.

```
extern "C++"
{
    int 14 x;
}
```

declares a variable, x, with C++ linkage.

```
extern "C"
{
    //remove Microsoft-specific extensions from the header file
    #define __cdecl
    #include <stdio.h>
}
```

causes everything in `stdio.h` to have C linkage.

Mapping of types to C/C++

Handel-C types will be mapped to C/C++ types in the following way when inside an `extern "`*`language`*`"` construct:

Handel-C type	C/C++ type
`char`	`char`
`short`	`short`
`long`	`long`
`int`	`int` (only valid within an `extern "`*`language`*`"` construct)
`int width`	`Int<`*`width`*`>` (C++ only)
`unsigned int width`	`UInt<`*`width`*`>` (C++ only)
`struct`	`struct`
`type` `ram[`*`n`*`]`	*`convertedType`*`[`*`n`*`]`
`type` `rom[`*`n`*`]`	*`convertedType`*`[`*`n`*`]`
Others	Generate an error

Mapping of types outside extern

Mapping of types outside the `extern "`*`language`*`"` construct is the same, except signed and unsigned `int`s must have a specified width.

When outside an `extern "`*`language`*`"` construct, an `int` without a specified width will generate an error.

For example, the following Handel-C:

```
extern "C" int printf(const char *format, ...);
extern "C++"
{
    int 14 x;
    long y;
}
char f(long y);  //outside extern construct
```

will map to this C++:

```
int printf(const char *format, ...);
Int<14> x;
long y;
char f(long y);
```

4.14 register

`register` has been implemented for reasons of compatibility with ANSI-C. `register` defines a variable that has local scope. Its initial value is undefined.

Example

```
register int 16 fish;
fish = f(plop);
```

4.15 inline functions

`inline` causes a function to be expanded where it is called. The logic will be generated every time it is invoked. This ensures that the function is not accessed at the same time by parallel branches of code.

If you have a local static variable in an inline function there is one copy of the variable per function instantiation.

By default, functions are assumed to be shared (not inline).

Example

```
inline int 4 knit(int needle, int stitch)
{
    needle = needle + stitch;
    return(needle);
}

int 4 jumper[100];
par(needle = 1; needle < 100; needle = needle+2)
{
    jumper[needle] = knit(needle, 1);
}
```

Syntax

`inline` *`function_Declaration`*

4.16 static

`static` gives a variable static storage (its values are kept at all times). This ensures that the value of a variable is preserved across function calls. It also affects the scope of a variable or a function. static functions and static variables declared outside functions can only be used in the file in which they appear. static variables declared within an inline function or an array of functions can only be used in the copy of the function in which they appear. Handel-C uses static in a different way to C++. In C++, if you have an `inline` function and a local static variable, one copy of the variable is shared across each function instantiation. In Handel-C, there is one copy of the variable per function instantiation.

`static` variables are the only local variables (excluding `const`s) that can be initialized. To get a default value, initialize the variable.

Example

```
static int 16 local_function (int water, int weed);
static int 16 local_fish = 1234;

void main(void)
{
    int fresh, pondweed;
    local_fish = local_function(fresh, pondweed);
    ...
}
```

Syntax

```
static variable_declaration;
static functionName(parameter-type-list);
```

Static variables in arrays of functions

If a static variable is declared in an arrayed function, each instance of the function will have its own independent copy of the variable.

4.17 typedef

`typedef` defines another name for a variable type. This allows you to clarify your code. The new name is a synonym for the variable type.

```
typedef int 4 SMALL_FISH;
```

If the `typedef` is used in multiple source files, it is good practice to collect all the type definitions in a header file, included at the top of each source file using the `#include` ***headerFileName*** directive. It is conventional to differentiate typedef names from standard variable names, so that they are easily recognizable.

Example

```
typedef int 4 SMALL_FISH;

extern SMALL_FISH stickleback;
```

4.18 typeof

The `typeof` type operator allows the type of an object to be determined at compile time. The argument to `typeof` must be an expression. Using `typeof` ensures that related variables maintain their relationship. It makes it easy to modify code by simplifying the process of sorting out type and width conflicts.

A `typeof` construct can be used anywhere a type name could be used. For example, you can use it in a declaration, in casts.

Syntax

```
typeof ( expression )
```

Example

```
unsigned 9 ch;
typeof(ch @ ch) q;
struct
{
    typeof(ch) cha, chb;
} s1;
typeof(s1) s2;

ch = s1.cha + s2.chb;
q = s1.chb @ s2.cha;
```

If the width of variable `ch` were changed in this example, there would be no need to modify any other code.

This is also useful for passing parameters to macro procedures. The code below shows how to use a `typeof` definition to deal with multiple parameter types.

```
macro proc swap (a, b)
{
    typeof(a) t;
    t=a;
    a=b;
    b=t;
}
```

4.19 const

`const` defines a variable or pointer or an array of variables or pointers that cannot be assigned to. This means that they keep the initialization value throughout. They may be initialized in the declaration statement. The `const` keyword can be used instead of `#define` to declare constant values. It can also be used to define function parameters which are never modified. The compiler will perform type-checking on const variables and prevent the programmer from modifying it.

Example 1

```
const int i = 5;

i = 10;  // Error
i++;     // Error
```

Example 2

```
const int *const p;

p = p + 1;  // Error
*p = 3;     // Error
```

4.20 volatile

In ANSI-C, `volatile` is used to declare a variable that can be modified by something other than the program.

It is mostly used for hard-wired registers. `volatile` controls optimization by forcing a re-read of the variable. It is only a guide, and may be ignored. The initial value of `volatile` variables is undefined.

Handel-C does nothing with `volatile`. It is accepted for compatibility purposes.

4.21 Complex declarations

It is possible to have extremely complex declarations in Handel-C. You can combine arrays of functions, structs, arrays, and pointers with architectural types. To clarify such expressions, it is wise to use `typedef`.

4.21.1 Macro expressions in widths

If you use a macro expression to provide the width in a type declaration, you must enclose it in parentheses. This ensures that it will be correctly parsed as a macro.

```
int (mac(x)) y;
```

To declare a pointer to a function returning that type, you get

```
int (mac(x)) (*f)();
```

4.21.2 <> (type clarifier)

< > is a Handel-C extension used to disambiguate complex declarations of architectural types. You cannot use it on logic types. It is good practice to use it whenever you declare channels, memories or signals, to clarify the format of data passed or stored in these variables.

It is required to disambiguate a declaration such as:

```
chan int *x; //pointer to channel or
             //channel of pointers?
```

This should be declared as

```
chan <int *> x; //channel of pointers
```

or

```
chan <int> *x; //pointer to channel
```

Example

```
struct fishtank
{
    int 4 koi;
    int 8 carp;
    int 2 guppy;
} bowl;

signal <struct fishtank> drip;
chan <int 8 (*runwater)()> tap;
```

4.21.3 Using signals to split up complex expressions

You can use signals to split up complex expressions. E.g.,

```
b = (((a * 2) - 55) << 2) + 100;
```

could also be written

```
int 17 a, b;
signal s1, s2, s3, s4;

par
{
    s1 = a;
    s2 = s1 * 2;
    s3 = s2 - 55;
    s4 = s3 << 2;
    b = s4 + 100;
}
```

Breaking up expressions also enables you to re-use sub-expressions:

```
unsigned 15 a, b;
signal sig1;

par
{
    sig1 = x + 2;
    a = sig1 * 3;
    b = sig1 / 2;
}
```

4.22 Variable initialization

Global, static and const variables

Global variables (i.e. those declared outside all code blocks) may be initialized with their declaration. For example:

```
static int 15 x = 1234;
```

```
static int 7 y = 45 with {outfile = "out.dat"};
```

Variables declared within functions or macros can only be initialized if they have `static` storage or are `const`s.

Global and static variables may only be initialized with constants. If you do not initialize them, they will have a default value of zero.

If you use the `set reset` construct, variables will be reset to their initial values. If you use the `try...reset` construct, variables will not be re-initialized.

All other variables

Local non-static variables have no default initial value. You cannot initialize them. Instead, you must use an explicit sequential or parallel list of assignments following your declarations to achieve the same effect. For example:

```
{
    int 4 x;
    unsigned 5 y;

    x = 5;
    y = 4;
}
```

Simulation

In simulation, variables (including static variables inside functions) are initialized before the simulation run begins (i.e. before the first clock cycle is simulated).

5 Statements

5.1 Sequential and parallel execution

Handel-C implicitly executes instructions sequentially. When targeting hardware it is extremely important to use parallelism. For this reason, Handel-C has a parallel composition keyword `par` to allow statements in a block to be executed in parallel.

Three assignments that execute in parallel and in the same clock cycle:

```
par
{
    x = 1;
    y = 2;
    z = 3;
}
```

Three assignments that execute sequentially, requiring three clock cycles:

```
x = 1;
y = 2;
z = 3;
```

The `par` example executes all assignments literally in parallel. Three specific pieces of hardware are built to perform these three assignments. This is about the same amount as is needed to execute the assignments sequentially.

Sequential branches

Within parallel blocks of code, sequential branches can be added by using a code block denoted with the `{...}` brackets instead of a single statement. For example:

```
par
{
    x = 1;
    {
        y = 2;
        z = 3;
    }
}
```

In this example, the first branch of the parallel statement executes the assignment to `x` while the second branch sequentially executes the assignments to `y` and `z`. The assignments to `x` and `y` occur in the same clock cycle, the assignment to `z` occurs in the next clock cycle.

The instruction following the `par {...}` will not be executed until all branches of the parallel block complete.

5.2 seq

To allow replication, the `seq` keyword exists. Sequential statements can be written with or without the keyword.

The following example executes three assignments sequentially:

```
x = 1;
y = 2;
z = 3;
```

as does this:

```
seq
{
    x = 1;
    y = 2;
    z = 3;
}
```

5.3 Replicated par and seq

You can replicate `par` and `seq` blocks by using a counted loop (a similar construct to a `for` loop). The count is defined with a start point (index_Base *below),* an end point (*index_Limit*) and a step size (*index_Count*). The body of the loop is replicated as many times as there are steps between the start and end points. If it is a `par` loop, the replicated processes will run in parallel, if a `seq`, they will run sequentially.

Syntax

```
par | seq (index_Base; index_Limit; index_Count)
{
    Body
}
```

The apparent variables used in *index_Base*, *index_Limit* and *index_Count* are `macro exprs` that are implicitly declared. *index_Base*, *index_Limit* and *index_Count* do not need to be single expressions, for example, you could declare `par (i=0, j=23; i != 76; i++, j--)`. In this case `i` and `j` are implicit `macro exprs`

Example

```
par (i=0; i<3; i++)
    {
        a[i] = b[i];
    }
```

expands to:

```
par
{
    a[0] = b[0];
    a[1] = b[1];
    a[2] = b[2];
}
```

Replicated pipeline example

```
unsigned init;
unsigned q[149];
unsigned 31 out;

init = 57;
par (r = 0; r < 16; r++)
{
    ifselect(r == 0)
        q[r] = init;
    else ifselect(r == 15)
        out = q[r-1];
    else
        q[r] = q[r-1];
}
```

`ifselect` checks for the start of the pipeline, the replicator rules create the middle sections and `ifselect` checks the end. The replicated code expands to:

```
par
{
    q[0] = init;
    q[1] = q[0];
    q[2] = q[1];
    etc...

    q[14] = q[13];
    out = q[14];
}
```

5.4 prialt

The `prialt` statement selects the first channel ready to communicate from a list of channel cases. The syntax is similar to a conventional C `switch` statement.

```
prialt
{
    case CommsStatement:
        Statement
        break;
    ......
    case CommsStatement:
        Statement
        break;
    ......
    [default:
        Statement
        break;]
}
```

`prialt` selects between the communications on several channels depending on the readiness of the other end of the channel. *CommsStatement* must be one of the following:

Channel ? *Variable*

Channel ! *Expression*

The case whose communication statement is the first to be ready to transfer data will execute and data will be transferred over the channel. The statements up to the next `break` statement will then be executed. If no channel is ready within a given clock tick, the default clause will be executed (if one is present)

Priority

If two channels are ready simultaneously, then the first one listed in the code takes priority.

Default

`prialt` with no `default` case:
execution halts until one of the channels becomes ready to communicate.

`prialt` statement with `default` case:
if none of the channels is ready to communicate immediately then the `default` branch statements executes and the `prialt` statement terminates.

Restrictions

Fall through of cases in a `prialt` construct is prohibited. This means that each `case` must have its own `break` statement. If the same channel is listed twice in its cases, only the first occurrence will ever be accessed. You would only wish to do this if the channel within the `prialt` is the result of an expression (e.g., a pointer to a channel or a reference to an array of channels). The compiler cannot reliably check this condition, so it will not cause a warning.

If a channel between clock domains has `fifolength=0` (default) and has a `prialt` on both sides, the compiler will convert it to have a `fifolength=1`. This is also true if a channel within a `prialt` has the other side within a `try reset` in a different clock domain.

5.5 Using prialt: examples

The `prialt` statement selects the first channel ready to communicate from a list of channel cases.

```
int 4 x, y, z;
chan <int 4> first, second;

par
{
    prialt
    {
        case first ! x:
            break;
        case second ! y:
            break;
}

seq
{
    delay;
    second ? z;
}
```

Send and receive statements can be mixed within a `prialt`. For example:

```
int 4 num, even, odd;
chan <int 4> ch1, ch2;

par
{
    if (num[0] != 0)
        ch1 ? odd;
    else
        ch2 ! num;

    prialt
    {
        case ch1 ! num:
            break;
        case ch2 ? even:
            break;
    }
}
```

Restrictions on using prialt

```
int 4 x, y;
chan <int 4> ch;

prialt
{
    case ch ! x:
        break;
    case ch ! y: //illegal: ch already used
        break;
}
```

```
int 4 x, y;
chan <int 4> ch;

prialt
{
    case ch ! x:
        break;
    case ch ? y: //illegal: ch already used
        break;
}
```

5.6 Assignments

Handel-C assignments are of the form:

Variable = *Expression*;

For example:

```
x = 3;
y = a + b;
```

The expression on the right hand side must be of the same width and type (signed or unsigned) as the variable on the left hand side. The compiler generates an error if this is not the case.

The left hand side of the assignment may be any variable, array element or RAM element. The right hand side of the assignment may be any expression.

Short cuts

The following short cut assignment statements cannot be used in expressions as they can in conventional C but only in stand-alone statements. See Introduction: Expressions for more information.

Shortcuts cannot be used with RAM variables, as they contravene the RAM access restrictions

Statement	Expansion
Variable ++;	*Variable* = *Variable* + 1;
Variable --;	*Variable* = *Variable* - 1;
++ *Variable*;	*Variable* = *Variable* + 1;
-- *Variable*;	*Variable* = *Variable* - 1;
Variable += *Expression*;	*Variable* = *Variable* + *Expression*;
Variable -= *Expression*;	*Variable* = *Variable* – *Expression*;
Variable *= *Expression*;	*Variable* = *Variable* * *Expression*;
Variable /= *Expression*;	*Variable* = *Variable* / *Expression*;
Variable %= *Expression*;	*Variable* = *Variable* % *Expression*;
Variable <<= *Expression*;	*Variable* = *Variable* << *Expression*;
Variable >>= *Expression*;	*Variable* = *Variable* >> *Expression*;
Variable &= *Expression*;	*Variable* = *Variable* & *Expression*;
Variable |= *Expression*;	*Variable* = *Variable* | *Expression*;
Variable ^= *Expression*;	*Variable* = *Variable* ^ *Expression*;

5.6.1 continue

`continue` moves straight to the next iteration of a `for`, `while` or `do` loop. For do or while, this means that the test is executed immediately. In a for statement, the increment step is executed. This allows you to avoid deeply nested `if ... else` statements within loops.

Example

```
for (i = 100; i > 0; i--)
{
    x = f( i );
    if ( x == 1 )
        continue;
    y += x * x;
}
```

You cannot use `continue` to jump out of or into `par` blocks.

5.6.2 goto

`goto` *`label`* moves straight to the statement specified by *`label`*. *`label`* has the same format as a variable name, and must be in the same function as the `goto`. Labels are local to the whole function, even if placed within an inner block. Formally, `goto` is never necessary. It may be useful for extracting yourself from deeply nested levels of code in case of error.

Example

```
for(… )
{
    for(… )
    {
        if(disaster)
            goto Error;
    }
}

Error:
    output ! error_code;
```

You cannot use `goto` to jump out of or into `par` blocks.

5.6.3 return [expression]

The return statement is used to return from a function to its caller. return terminates the function and returns control to the calling function. Execution resumes at the line immediately following the function call. return can return a value to the calling function. The value returned is of the type declared in the function declaration. Functions that do not return a value should be declared to be of type void.

Example

```
int power(int base, int n)
{
    int i, p;

    p = 1;
    for (i = 1; i <= n; ++i)
        p = p * base;
    return(p);
}
```

You cannot use return to jump out of par blocks.

5.6.4 Conditional execution (if ... else)

Handel-C provides the standard C conditional execution construct as follows:

```
if (Expression)
    Statement
else
    Statement
```

As in conventional C, the else portion may be omitted if not required. For example:

```
if (x == 1)
    x = x + 1;
```

Statement may be replaced with a block of statements by enclosing the block in {...} brackets. For example:

```
if (x>y)
{
    a = b;
    c = d;
}
else
{
    a = d;
    c = b;
}
```

The first branch of the conditional is executed if the expression is true and the second branch is executed if the expression is false. Handel-C treats zero values as false and non-zero values as true. Relational and logical operators return values to match this meaning but it is also possible to use variables as conditions. For example:

```
if (x)
    a = b;
else
    c = d;
```

This is expanded by the compiler to:

```
if (x!=0)
    a = b;
else
    c = d;
```

When executed, if `x` is not equal to 0 then `b` is assigned to `a`. If `x` is 0 then `d` is assigned to `c`.

5.6.5 while loops

Handel-C provides `while` loops exactly as in conventional C:

```
while (Expression)
    Statement
```

The contents of the `while` loop may be executed zero or more times depending on the value of *`Expression`*. While *`Expression`* is true then *`Statement`* is executed repeatedly. *`Statement`* may be replaced with a block of statements. For example:

```
x = 0;
while (x != 45)
{
    y = y + 5;
    x = x + 1;
}
```

This code adds 5 to `y` 45 times (equivalent to adding 225 to `y`).

5.6.6 do ... while loops

Handel-C provides `do ... while` loops exactly as in conventional C:

```
do
    Statement
while (Expression);
```

The contents of the `do ... while` loop is executed at least once because the conditional expression is evaluated at the end of the loop rather than at the beginning as is the case with `while` loops. ***Statement*** may be replaced with a block of statements. For example:

```
do
{
    a = a + b;
    x = x - 1;
} while (x>y);
```

5.6.7 for loops

Handel-C provides `for` loops similar to those in conventional C.

```
for (Initialization ; Test ; Iteration)
    Statement
```

The body of the `for` loop may be executed zero or more times according to the results of the condition test. There is a direct correspondence between `for` loops and `while` loops. Because of the benefits of parallelism, it is nearly always preferable to implement a `while` loop instead.

```
for (Init; Test; Inc)
    Body;
```

is directly equivalent to:

```
{
    Init;
    while (Test)
    {
        Body;
        Inc;
    }
}
```

unless the *Body* includes a `continue` statement. In a `for` loop `continue` jumps to before the increment, in a `while` loop `continue` jumps to after the increment.

Unless a specific `continue` statement is needed, it is always faster to implement the *for* loop as a *while* loop with the *Body* and *Inc* steps in parallel rather than in sequence when this is possible.

Each of the initialization, test and iteration statements is optional and may be omitted if not required. Note that `for` loops with no iteration step can cause combinational loops. As with all other Handel-C constructs, *Statement* may be replaced with a block of statements. For example:

```
for ( ; x>y ; x++ )
{
    a = b;
    c = d;
}
```

The difference between a conventional C `for` loop and the Handel-C version is in the initialization and iteration phases. In conventional C, these two fields contain expressions and by using expression side effects (such as ++ and --) and the sequential operator ',' conventional C allows complex operations to be performed. Since Handel-C does not allow side effects in expressions the initialization and iteration expressions have been replaced with statements. For example:

```
for (x = 0; x < 20; x = x+1)
{
    y = y + 2;
}
```

Here, the assignment of 0 to `x` and adding one to `x` are both statements and not expressions. These initialization and iteration statements can be replaced with blocks of statements by enclosing the block in {...} brackets. For example:

```
for ({ x=0; y=23;} ; x < 20; {x+=1; x*=2;} )
{
    y = y + 2;
}
```

5.6.8 switch

Handel-C provides `switch` statements similar to those in conventional C.

```
switch (Expression)
{
    case Constant:
        Statement
        break;
    ......
    default:
        Statement
        break;
}
```

The `switch` expression is evaluated and checked against each of the `case` compile time constants. The statement(s) guarded by the matching constant is executed until a `break` statement is encountered.

If no matches are found, the `default` statement is executed. If no default option is provided, no statements are executed.

Each of the *`Statement`* lines above may be replaced with a block of statements by enclosing the block in `{...}` brackets.

As with conventional C, it is possible to make execution drop through case branches by omitting a `break` statement. For example:

```
switch (x)
{
case 10:
    a = b;
case 11:
    c = d;
    break;

case 12:
    e = f;
    break;
}
```

Here, if `x` is 10, `b` is assigned to `a` and `d` is assigned to `c`, if `x` is 11, `d` is assigned to `c` and if `x` is 12, `f` is assigned to `e`.

The values following each case branch must be compile time constants.

5.6.9 break

Handel-C provides the normal C `break` statement for:

- terminating loops
- separation of `case` branches in `switch` and `prialt` statements.

`break` cannot be used to jump into or out of `par` blocks.

Loops

When used within a `while`, `do...while` or `for` loop, the loop is terminated and execution continues from the statement following the loop. For example:

```
for (x=0; x<32; x++)
{
    if (a[x]==0)
        break;
    b[x]=a[x];
}
// Execution continues here
```

switch

When used within a `switch` statement, execution of the `case` branch terminates and the statement following the `switch` is executed. For example:

```
switch (x)
{
    case 1:
    case 2:
        y++;
        break;
    case 3:
        z++;
        break;
}
// Execution continues here
```

prialt

When used within a `prialt` statement, execution of the case branch terminates and the statement following the `prialt` is executed. For example:

```
prialt
{
    case a ? x:
        x++;
        break;
    case b ! y:
        y++;
        break;
}
// Execution continues here
```

5.6.10 delay

Handel-C provides a `delay` statement, not found in conventional C, which does nothing but takes one clock cycle to do it. This may be useful to avoid resource conflicts (for example to prevent two accesses to one RAM in a single clock cycle) or to adjust execution timing.

`delay` can also be used to break combinational logic cycles.

5.6.11 try... reset

`try...reset` allows you to perform actions on receipt of a reset signal within a specified section of code. You can form the same kind of construct with other control statements, but this requires more complex code and therefore more hardware.

Syntax

```
try
{
    statements
}
reset(condition)
{
    statements
}
```

During the execution of statements within the try block, if condition is true, the reset statement block will be executed immediately, else it will not. The condition expression is continually checked. If it occurs in the middle of a function, execution will immediately go to the reset thread. Static variables within the function will remain in the state they were in when the reset condition occurred. Variables and RAMs will not be re-initialized.

Examples

```
void main(void)
{
    interface bus_in(int 1 input) resetbus();
    try
    {
        someFunction();
    }
    reset(resetbus.input == 1)
    {
        cleanUpSomeFunction();
    }
}
```

If you have nested `try…reset` statements, and more than one try condition is true, only the outermost reset statement is executed. For example:

```
unsigned 4 a, s, t, x, y;
static unsigned 1 condition = 0;

par
{
     while(1)
     {
           condition = (a == 1);
     }

     try
     {
          try
          {
                a = 1;
                a = 2;
                a = 3;
          }
          reset(condition)
          {
                s = 1;
                t = 1;
          }
     }
     reset (condition)
     {
          x = 1;
          y = 1;
     }
}
```

will execute the second reset statement only.

5.6.12 trysema()

`trysema(`***`semaphore`***`)` tests to see if the semaphore is owned. If not, it returns one and takes ownership of the semaphore. If it is, it returns zero. A semaphore may be freed by using the statement `releasesema(`***`semaphore`***`)`.

Example

```
inline void critRAMaccess(sema *RAMsema, ram int 8
    (*danger)[4], unsigned count)
{
    int 8 x;
    // wait till you've got the RAM
    while(trysema(*RAMsema)==0) delay;
    x= (*danger)[count];
    releasesema(*RAMsema);
}
```

Note that you can no longer take the semaphore twice without releasing it.

```
while(1)
{
    // always succeeds because its the same 'trysema' expression
    if (trysema(s)) {...}
}
```

In DK version 1, this worked. In DK version 1.1 and subsequent versions, the second and subsequent `trysema()` will always fail. Instead, use

```
while(1)
{
    if (trysema(s))
    {
        ...
        releasesema(s)
    }
}
```

5.6.13 releasesema()

`releasesema(`***`semaphore`***`)` releases a `semaphore` that was previously taken by `trysema(`***`semaphore`***`)`.

Example

```
inline void critRAMaccess(sema *RAMsema, ram int 8
   (*danger)[4], unsigned count)
{
    int 8 x;
    while(trysema(*RAMsema)==0) delay; // wait till you've got the RAM
    x= (*danger)[count];
    releasesema(*RAMsema);
}
```

6 Expressions

6.1 Introduction to expressions

6.1.1 Clock cycles required

Expressions in Handel-C take no clock cycles to be evaluated, and so have no bearing on the number of clock cycles a given program takes to execute.

They affect the maximum possible clock rate for a program: the more complex an expression, the more hardware is involved in its evaluation and the longer it is likely to take because of combinational delays in the hardware. The clock period for the entire hardware program is limited by the longest such evaluation in the whole program.

Because expressions are not allowed to take any clock cycles, expressions with side effects are not permitted in Handel-C. For example;

```
if (a<b++)   /* NOT PERMITTED */
```

This is not permitted because the ++ operator has the side effect of assigning `b+1` to `b` which requires one clock cycle.

6.1.2 Breaking down complex expressions

The longest and most complex C statement with many side effects can be written in terms of a larger number of simpler expressions and assignments. The resulting code is normally easier to read. For example:

```
a = (b++) + (((c-- ? d++ : e--)) , f);
```

can be rewritten as:

```
a = b + f;
b = b + 1;
if (c)
    d = d + 1;
else
    e = e - 1;
c = c - 1;
```

6.1.3 Prefix and postfix operators

Handel-C provides the prefix and postfix ++ and -- operations as statements rather than expressions. For example:

```
a++;
b--;
++c;
--d;
```

is directly equivalent to:

```
a = a + 1;
b = b - 1;
c = c + 1;
d = d - 1;
```

6.2 Casting of expression types

Automatic conversions between signed and unsigned values are not allowed. Values must be cast between types to ensure that the programmer is aware that a conversion is occurring that may alter the meaning of a value.

You can cast to a type of undefined width. For example:

```
int 4 x;
unsigned int undefined y;

x = (int undefined)y;
```

The compiler will infer that y must be 4 bits wide.

Explanation of signed/unsigned casting

The following piece of Handel-C is invalid:

```
int 4 x;           // Range of x: -8...7
unsigned int 4 y; // Range of y: 0...15

x = y;             // Not allowed
```

This is because x is a signed integer while y is an unsigned integer. When generating hardware, it is not clear what the compiler should do here. It could simply assign the 4 bits of y to the 4 bits of x or it could extend y with an extra zero as its most significant bit to preserve its value and then assign these 5 bits to x assuming x was declared to be 5 bits wide.

To see the difference, consider the case when y is 10. By simply assigning these 4 bits to a signed integer, a result of -6 would be placed in x. A better solution might be to extend y to a five bit value by adding a 0 bit as its MSB to preserve the value of 10.

A programmer must explicitly cast the variables to the same type. Assuming that they wish to use the 4-bit value as a signed integer, the above example then becomes:

```
int 4 x;
unsigned int 4 y;

x = (int 4)y;
```

It is now clear that the value of x is the result of treating the 4 bits extracted from y as a signed integer.

6.2.1 Restrictions on casting

Casting cannot be used to change the width of values. For example, this is not allowed:

```
unsigned int 7 x;
int 12 y;

y = (int 12)x;  // Not allowed
```

The conversion should be done explicitly:

```
y = (int 12)(0 @ x);
```

Here, the concatenation operation produces a 12-bit unsigned value. The casting then changes this to a 12-bit signed integer for assignment to y.

This is to ensure that the programmer is aware of such conversions.

Explanation

```
int 7 x;
unsigned int 12 y;

x = -5;
y = (unsigned int 12)x;
```

The Handel-C compiler could take two routes. One would be to sign extend the value of x and produce the result 4091. The second would be to zero pad the value of x and produce the value of 123. Since neither method can preserve the value of x in y Handel-C performs neither automatically. Rather, it is left up to the programmer to decide which approach is correct in a particular situation and to write the expression accordingly. You may sign extend using the adjs macro and zero-pad using the adju macro. These macros are provided in the standard macro library within the Celoxica Platform Developer's Kit.

6.3 Restrictions on RAMs and ROMs

Because of their architecture, RAMs and ROMs are restricted to performing operations sequentially. Only one element of a RAM or ROM may be addressed in a single clock

cycle. In hardware, this means you can only write one value to the address port of a memory, allowing one read access or one write access. You can detect simultaneous memory accesses when you are debugging your code by using the **Detection of simultaneous memory accesses** option on the **Debug** tab in Project Settings, or the `-S+parmem` option in the command line compiler.

If you want to make more than one access to a memory at a time, use an MPRAM (multi-ported RAM). You can access more than one port at a time, but you can only make a single access to any one `mpram` port in a single clock cycle.

Example of disallowed assignment

Only one element of a RAM or ROM may be addressed in any given clock cycle and, as a result, familiar looking statements will often produce unexpected results. For example:

```
ram <unsigned int 8> x[4];
x[1] = x[3] + 1;
```

This code should not be used because the assignment attempts to read from the third element of `x` in the same cycle as it writes to the first element, and the memory may produce undefined results.

Example of disallowed condition evaluation

```
ram unsigned int 8 x[4];

if (x[0]==0)
    x[1] = 1; //double access, disallowed
```

This code is illegal because the condition evaluation must read from element 0 of the RAM in the same clock cycle as the assignment writes to element 1. Similar restrictions apply to `while` loops, `do ... while` loops, `for` loops and `switch` statements.

Incorrect execution with conditional operator

This code will not execute correctly because of the double access.

```
x = y>z ? RamA[1] : RamA[2];
```

The solution is to re-write the code as follows:

```
x = RamA[y>z ? 1 : 2];
```

Here, there is only a single access to the RAM so the problem does not occur.

Arrays of variables do not have these restrictions but may require substantially more hardware to implement than RAMs (depending on the target architecture).

6.4 assert

assert allows you to generate messages at compile-time if a condition is met. The messages can be used to check compile-time constants and help guard against possible problematic code alterations. The user uses an expression to check the value of a compile-time constant, and if the expression evaluates to false, an error message is sent to the standard error channel in the format

filename:*line number*, *start column* - *end column*::Assertion failed: *user-defined error string*

The default error message is:

"Error : User assertion failed"

If the expression evaluates to true, the whole assert expression is replaced by a constant expression.

assert can be used as a statement by passing 0 as the *trueValue*. If the condition is true, the whole assert statement is replaced by 0 (a null statement). This is shown in the example below. If the width of x is 3 (the condition is true), the whole statement is replaced by the *trueValue* of 0, so nothing happens.

```
assert (width(x)==3, 0, "Width of x is not 3 (it is %d)", width(x));
```

A more detailed example is given below. assert can also be used as an expression, where its return value is assigned to something. This is illustrated in the second example below, where the return value is assigned to *ReturnVal*.

Syntax

assert(*condition*,*trueValue* [*string with format specification(s)*
{,*argument(s)*}]);

If *condition* is true, the whole expression reduces to ***trueValue.*** If *condition* is false, *string* will be sent to the standard error channel, with each *format specification* replaced by an *argument*. When assert encounters the first format specification (if any), it converts the value of the first argument into that format and outputs it. The second argument is formatted according to the second format specification and so on. If there are more expressions than format specifications, the extra expressions are ignored. The results are undefined if there are not enough arguments for all the format specifications.

The format specification is one of:

%c	Display as a character	%s	Display as a string
%d	Display as a decimal	%f	Display as a floating-point
%o	Display as an octal	%x	Display as a hexadecimal

An assert evaluates to an empty statement and can only appear after all declarations in a macro or function

Using assert as a statement

In the example below assert is used as a statement.

```
set clock = external "C1";
int f(int x)
{
    assert(width(x)==3, 0, "Width of x is not 3 (it is %d)", width(x));
    return x+1;
}

void main(void)
{
    int 4 y;
    y = f(y);
}
```

x will be inferred to have a width of 4, so the following message will be displayed.

```
F:\proj\test.hcc(4)(2) : Assertion failed : Width of x is not 3 (it is 4)
```

Using assert as an expression

In the example below, assert is used as an expression.

```
set clock = external "C1";
unsigned func(unsigned p, unsigned q)
{
    macro expr WidthSum(a, b) = width(a) + width(b);
    macro expr CheckWidths(a, b) = assert((WidthSum(a, b)==32
        || WidthSum(a, b)==16), WidthSum(a, b),
        "Sum of widths of function parameters is not 16 or 32 (it is %d)",
        WidthSum(a, b));
    unsigned 16 ReturnVal;

    ReturnVal = CheckWidths(p, q);

    return ReturnVal;
}

void main(void)
{
    static unsigned 9 x;
    static unsigned 7 y;
    unsigned result;

    result = func(x, y);
}
```

6.5 Bit manipulation operators

The following bit manipulation operators are provided in Handel-C:

`<<`	Shift left
`>>`	Shift right
`<-`	Take least significant bits
`\\`	Drop least significant bits
`@`	Concatenate bits
`[]`	Bit selection
`width(`*`Expression`*`)`	Width of expression

6.5.1 Shift operators

The shift operators shift a value left or right by a variable number of bits resulting in a value of the same width as the value being shifted. Any bits shifted outside this width are lost.

When shifting unsigned values, the right shift pads the upper bits with zeros. When right shifting signed values, the upper bits are copies of the top bit of the original value. Thus, a shift right by 1 divides the value by 2 and preserves the sign. For example:

```
static unsigned 4 a = 0b1101;
static unsigned (log2ceil(width(a)+1)) b = 2;

a = a >> b; //a becomes 0b0011
b--;
a = a >> b; //a becomes 0b0001
```

The width of `b` needs to have a width equal to `log2(width(a)+1)` rounded up to the nearest whole number. This can be calculated using the `log2ceil` macro which is provided as part of the standard library in the Platform Developer's Kit.

6.5.2 Take / drop operators

The take operator, <-, returns the n least significant bits of a value. The drop operator, \\, returns all but the n least significant bits of a value. *n* must be a compile-time constant. For example:

```
macro expr four = 8 / 2;
unsigned int 8 x;
unsigned int 4 y;
unsigned int 4 z;

x = 0xC7;
y = x <- four;
z = x \\ 4;
```

This results in `y` being set to 7 and `z` being set to 12 (or 0xC in hexadecimal).

6.5.3 Concatenation operator

The concatenation operator, `@`, joins two sets of bits together into a result whose width is the sum of the widths of the two operands. For example:

```
unsigned int 8 x;
unsigned int 4 y;
unsigned int 4 z;

y = 0xC;
z = 0x7;
x = y @ z;
```

This results in x being set to 0xC7. The left operand of the concatenation operator forms the most significant bits of the result.

You may also use the concatenation operator to zero pad a variable to a given width.

```
unsigned int 8 x;
unsigned int 8 y;
unsigned int 16 z;

//width of zero constant inferred to be 8 bits
z = (0 @ x) * (0 @ y);
```

If you want to use sign extension, you need to copy the 1 or the 0 from the most significant bit into the new bits. For example:

```
signed int 8 i;
signed int 12 j;
j = i[7] @ i[7] @ i[7] @ i[7] @ i;
```

6.5.4 Bit selection

Individual bits or a range of bits may be selected from a value by using the [] operator. Bit 0 is the least significant bit and bit *n*-1 is the most significant bit where *n* is the width of the value. For example:

```
unsigned int 8 x;
unsigned int 1 y;
unsigned int 5 z;

x = 0b01001001;
y = x[4];
z = x[7:3];
```

This results in y being set to 0 and z being set to 9. Note that the range of bits is of the form *MSB*:*LSB* and is inclusive. Thus, the range 7:3 is 5 bits wide.

The bit selection values must be fixed at compile time.

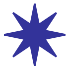
The value before or after ':' can be omitted. If you omit the value after the semi-colon, then zero is assumed, so the LSBs are taken. If you omit the value before the semi-colon, then n–1 is assumed, so the MSBs are taken.

Bit selection is allowed in RAM, ROM and array elements. For example:

```
ram int 7 w[23];
int 5 x[4];
int 3 y;
unsigned int 1 z;

y = w[10][4:2];
z = (unsigned 1)x[2][0];
```

The 10 specifies the RAM entry and the `4:2` selects three bits from the middle of the value in the RAM *w* is set to the value of the selected bits.

Similarly, `z` is set to the least significant bit in the `x[2]` variable.



You cannot assign to bit ranges, only read from them.

6.5.5 Width operator

The `width()` operator returns the width of an expression. It is a compile time constant. For example:

```
x = y <- width(x);
```

This takes the least significant bits of `y` and assigns them to `x`. The `width()` operator ensures that the correct number of bits is taken from `y` to match the width of `x`.

6.6 Arithmetic operators

The following arithmetic operators are provided in Handel-C:

Operator	Meaning
+	Addition
-	Subtraction
*	Multiplication
/	Division
%	Modulo arithmetic

Any attempt to perform one of these operations on two expressions of differing widths or types results in a compiler error. For example:

```
int 4 w;
int 3 x;
int 4 y;
unsigned 4 z;

y = w + x; // ILLEGAL
z = w + y; // ILLEGAL
```

The first statement is illegal because `w` and `x` have different widths. The second statement is illegal because `w` and `y` are signed integers and `z` is an unsigned integer.

Width of results

All operators return results of the same width as their operands. Thus, all overflow bits are lost. For example:

```
unsigned int 8 x;
unsigned int 8 y;
unsigned int 8 z;

x = 128;
y = 192;
z = 2;

x = x + y;
z = z * y;
```

This example results in `x` being set to 64 and `z` being set to 128.

By using the bit manipulation operators to expand the operands, it is possible to obtain extra information from the arithmetic operations. For instance, the carry bit of an addition or the overflow bits of a multiplication may be obtained by first expanding the operands to the maximum width required to contain this extra information. For example:

```
unsigned int 8 u;
unsigned int 8 v;
unsigned int 9 w;
unsigned int 8 x;
unsigned int 8 y;
unsigned int 16 z;

w = (0 @ u) + (0 @ v);
z = (0 @ x) * (0 @ y);
```

In this example, `w` and `z` contain all the information obtainable from the addition and multiplication operations. Note that the constant zeros do not require a width specification because the compiler can infer their widths from the usage. The zeros in the first assignment must be 1 bit wide because the destination is 9 bits wide while the source operands are only 8 bits wide. In the second assignment, the zero constants must be 8 bits wide because the destination is 16 bits wide while the source operands are only 8 bits wide.

6.7 Relational operators

Operator	**Meaning**
==	Equal to
!=	Not equal to
<	Less than
>	Greater than
<=	Less than or equal
>=	Greater than or equal

These operators compare values of the same width and return a single bit wide `unsigned int` value of 0 for false or 1 for true. This means that this conventional C code is invalid:

```
unsigned 8 w, x, y, z;

w = x + (y >z); // NOT ALLOWED
```

Instead, you should write:

```
w = x + (0 @ (y > z));
```

6.7.1 Signed/unsigned compares

Signed/signed compares and unsigned/unsigned compares are handled automatically. Mixed signed and unsigned compares are not handled automatically. For example:

```
unsigned 8 x;
int 8 y;

if (x>y) // Not allowed
    ...
```

To compare signed and unsigned values you must sign extend each of the parameters. The above code can be rewritten as:

```
unsigned 8 x;
int 8 y;

if ((int)(0@x) > (y[7]@y))
    ...
```

6.7.2 Implicit compares

The Handel-C compiler inserts implicit compares with zero if a value is used as a condition on its own. For example:

```
while (1)
{
    ...
}
```

Is directly expanded to:

```
while (1 != 0)
{
    ...
}
```

6.8 Logical operators

Operator	Meaning
`&&`	Logical and
`||`	Logical or
`!`	Logical not

These operators are provided to combine conditions as in conventional C. Each operator takes 1-bit unsigned operands and returns a 1-bit unsigned result.

Note that the operands of these operators need not be the results of relational operators. This feature allows some familiar looking conventional C constructs.

Example

```
if (x || y > z)
    w = 0;
```

In this example, the variable `x` need not be 1 bit wide. If it is wider, the Handel-C compiler inserts a compare with 0.

```
if (x != 0 || y > z)
    w = 0;
```

The condition of the `if` statement is true if `x` is not equal to 0 or `y` is greater than `z`.

C-like example

```
while (x || y)
{
    ...
}
```

Again, if the variables are wider than 1-bit, the Handel-C compiler inserts compares with 0.

6.8.1 Bitwise logical operators

Operator	Meaning
`&`	Bitwise and
`|`	Bitwise or
`^`	Bitwise exclusive or
`~`	Bitwise not

These operators perform bitwise logical operations on values. Both operands must be of the same type and width: the resulting value will also be this type and width. For example:

```
unsigned int 6 w;
unsigned int 6 x;
unsigned int 6 y;
unsigned int 6 z;

w = 0b101010;
x = 0b011100;
y = w & x;
z = w | x;
w = w ^ ~x;
```

This example results in y having the value 0b001000, z having the value 0b111110 and w having the value 0b001001.

6.9 Conditional operator

Handel-C provides the conditional expression construct familiar from conventional C. Its format is:

Expression ? *Expression* : *Expression*

The first expression is evaluated and if true, the whole expression evaluates to the result of the second expression. If the first expression is false, the whole expression evaluates to the result of the third expression. For example:

```
x = (y > z) ? y : z;
```

This sets x to the maximum of y and z. This code is directly equivalent to:

```
if (y > z)
    x = y;
else
    x = z;
```

The advantage of using this construct is that the result is an expression so it can be embedded in a more complex expression. For example:

```
x = ((w==0) ? y : z) + 4;
```

In this case, the signedness and widths of x, y and z must match (as the value of y or z may be assigned to x), but those of w need not.

6.10 Member operators (. / ->)

The structure member operator (.) is used to access members of a structure or mpram, or to access a port within an interface.

The structure pointer operator (->) can be used, as in ANSI-C. It is used to access the members of a structure or mpram, when the structure/mpram is referenced through a pointer.

```
mpram Fred
{
    ram <unsigned 8> ReadWrite[256]; // Read/write port
    rom <unsigned 8> Read[256];   // Read only port
} Joan;

mpram Fred *mpramPtr;
mpramPtr = &Joan;
x = mpramPtr->Read[56];
```

If a memory is made up of structures, the structure member operator can be used to reference structure members within the memory.

```
ram struct S compRAM[100];
ram struct S (*ramStructPtr)[];
ramStructPtr = &compRAM;
x = (*ramStructPtr)[10].a;
```

7 Functions and macros

7.1 Functions and macros: overview

Handel-C includes and extends the range of functions and macros offered by ANSI-C.

	Return value?	**Typed return values and parameters ?**	**Called by reference?**	**Shared hardware?**
Functions	Can have	Yes	No	Yes
Arrays of functions	Can have	Yes	No	Yes
Inline functions	Can have	Yes	No	No
Preprocessor macros	Can have	No	Yes	No
Macro expressions	Must have	No	Yes	No
Shared expressions	Must have	No	Yes	Yes
Macro procedures	None	No	Yes	No

7.1.1 Functions and macros: language issues

Called by reference or value

Functions employ call-by-value on their parameters, whereas macros effectively employ call-by-reference. Consider the code:

```
void inline f_pseudoswap (int 12 x, int 12 y)
{
    par
    {
        x = y;
        y = x;
    }
}
```

```
macro proc mp_swap (x, y)
{
    par
    {
        x = y;
        y = x;
    }
}
```

If you call `mp_swap(a,b)` the values of `a` and `b` will be swapped.

If you call call `f_pseudoswap(a,b)` the values `a` and `b` are copied to the formal parameters `x` and `y` of `f_pseudoswap`. `x` and `y` are swapped, but `a` and `b` are unaffected. The swap function with the same behaviour as the macro procedure is therefore

```
void inline f_swap (int 12 * x, int 12 * y)
{
    par
    {
        * x = * y;
        * y = * x;
    }
}
```

with a call of the form `f_swap(&a,&b)`.

Typed or untyped parameters

Function parameters must have a type, although the width can sometimes be inferred by the compiler.

Macro expressions and procedures are un-typed in the sense that their formal parameters can't be given types. The type of macro parameters is inferred from the type in the call statement.

This means that it is better to use macros for parameterizable code. For example, macro procedures can be used in libraries if you want to create multiple instances of hardware, but leave them untyped to make the code more generic.

Recursion

In Handel-C, functions may not be recursive. Macro procedure and macro expressions can be used to capture compile-time recursion.

If you use recursive macro procedures you need to use `ifselect` to guard the base case (the condition where the recursion terminates). If you use recursive macro expressions, you need to use `select` to guard the base case.

Macro procedure example:

```
unsigned 4 g;
macro proc p(x)
{
    ifselect(width(x) != 0)
    {
        g = 0@x;
        p(x\\1);
    }
    else
    delay;
}

set clock = external;
void main()
{
    unsigned 4 i;
    p(i);
}
```

Macro expression example:

```
macro expr copycat (copies, bits) =
    select (copies <= 0, (unsigned 0) 0,
    bits @ copycat (copies - 1, bits));
```

7.1.2 Functions and macros: sharing hardware

Calls to functions and shared expressions result in a single shared piece of hardware. This is equivalent to an ANSI-C function resulting in a single shared section of machine code.

Shared hardware will reduce the size of your design, but care is needed if you have parallel code where multiple branches access the shared hardware. Shared hardware may also compromise the speed of your design as it tends to lead to an increase in logic depth.

Each call to an inline function, macro procedure or macro expression results in a separate piece of hardware.

Arrays of functions allow a specified number of copies to be created.

7.1.3 Functions and macros: clock cycles

Macro expressions and shared expressions are evaluated in a single clock cycle, where the expression is assigned to a variable. Functions and macro procedures may involve control logic, and may take many cycles.

7.1.4 Functions and macros: examples

There are many ways in which a much-used code fragment can be expressed. The examples below all multiply a value by 1.5. For hints on when to use the different types of macros and functions, see:

- Functions and macros: overview
- Comparison of macros and functions

Preprocessor macro

```
#define de_sesqui(s) ((s) + ((s) >> 1))
#define dp_sesqui(d,s) ((d) = (s) + ((s) >> 1))
```

Macro expression

```
macro expr me_sesqui (s) = s + (s >> 1);
```

Shared expression

```
shared expr se_sesqui (s) = s + (s >> 1);
```

Macro procedure

```
macro proc mp_sesqui (d, s)
{
    d = s;
    d += (d >> 1);
}
```

Function

```
void f_sesqui (int * d, int s) //"shared" function without return
{
    * d = s;
    * d += ((* d) >> 1);
}
```

```
int rf_sesqui (int s) //"shared" function with return
{
    int ret;
    ret = s;
    ret += (ret >> 1);
    return ret;
}
```

Array of functions

```
void af_sesqui [5] (int * d, int s) //function array without return
{
   * d = s;
   * d += ((* d) >> 1);
}

int arf_sesqui [5] (int s) // function array with return
{
    int ret;
    ret = s;
    ret += (ret >> 1);
    return ret;
}
```

Inline function

```
void inline if_sesqui (int * d, int s) // inline function without return
{
    * d = s;
    * d += ((* d) >> 1);
}


// inline function with return
int inline irf_sesqui (int s)
{
    int ret;
    ret = s;
    ret += (ret >> 1);
    return ret;
}
```

How to call the example macros and functions

The example macros and functions above can be called using code such as:

```
{
    int 5 x, y;
    x = 10;

    y = de_sesqui (x);
    dp_sesqui (y, x);

    y = me_sesqui (x);

    y = se_sesqui (x);

    mp_sesqui (y, x);

    f_sesqui (& y, x);
    y = rf_sesqui (x);

    af_sesqui[2] (& y, x);
    y = arf_sesqui[2] (x);

    if_sesqui (& y, x);
    y = irf_sesqui (x);
}
```

7.1.5 Accessing external names

You can refer to functions, macros and shared expressions that have been defined in another file by prototyping them. You prototype by declaring an object at the top of the file in which it is used.

Function prototypes are in the following format:

returnType functionName(*parameterTypeList*);

Macro prototypes are of the form:

macro expr ***Name***(*parameterList*);

macro proc ***Name***(*parameterList*);

Functions and macros may be `static` or `extern`. `static` functions and macros may only be used in the file where they are defined.

You can collect all the prototypes into a single header file and then `#include` it within your code files.

You can access variables declared in other files by using the `extern` keyword.

You cannot use variables to communicate between clock domains. Variables are restricted to a single clock domain. The only items that can connect across separate clock domains are channels and `mprams`.

7.1.6 Recursion in macros and functions

Macros can be recursive in Handel-C, but due to the absence of a stack in Handel-C, functions cannot be recursive.

The depth of recursion, though unbounded, must be determinable at compile-time.

7.2 Introduction to functions

Functions are similar to functions in ANSI-C. A function is compiled to be a single shared piece of hardware, much as a C compiler generates a single shared block of machine code.

Handel-C has been extended to provide arrays of functions and inline functions.

Arrays of functions provide multiple copies of a function. You can select which copy is used at any time.

Inline functions are similar to macros in that they are expanded wherever they are used.

You may also use a macro proc (a parameterized macro procedure).

Functions take arguments and return values. A function that does not return a value is of type `void`. Valid return types are integers and structs. The default return type is `int undefined`. Functions that do not take arguments have `void` as their parameter list, for example:

```
void main(void)
```

As in ANSI-C, function arguments are passed by value. This means that a local copy is created that is only in scope within the function. Changes take place on this copy.

To access a variable outside the function, you must pass the function a pointer to that variable. A local copy will be made of the pointer, but it will still point to the same variable. This is known as passing by reference.

Architectural types (hardware constructs) must be passed by reference (a pointer to or address of the construct). The only architectural type that can be passed to or returned by a function by value is a signal. All others (and structures containing them) must be passed by reference. Arrays and functions can also only be passed by reference.

7.2.1 Function definitions and declarations

Function definitions and declarations are defined as in ANSI-C. Functions must be declared in every file that they are used in, though they should only be defined once. It is common to put function declarations into a header file and `#include` that in every file where they are used.

Function definition

The definition of a function consists of its name and parameters followed by the function body (the block of code that it performs when it is called).

The syntax is:

```
returnType Name(parameterList)
{
    declarations
    statements
}
```

For example:

```
int 4 add (int 4 left, int 4 right)
{
    int 4 sum;
    sum = left + right;
    return sum;
}
```

If there is nothing returned from the function, a void return type must be specified.

Old-style ANSI-C function definitions, where the types of the parameters are specified between the parameter list and the function body, are not supported. For example:

```
int 4 add (left, right) //old-style not supported
int 4 left, right;
{
    return left + right;
}
```

Function declaration

A function declaration lists the function name, return type and the types of the parameters. The syntax is:

```
returnType Name(parameterType_1 parameter_1, parameterType_n parameter_n);
```

Note the semicolon following the parameter list.

You may omit the parameter names in a declaration. The parameter types are used by the compiler to check that the correct types are used for the function arguments within the rest of the file.

Old-style ANSI-C declarations, where the names but not the type of the parameters are given, are not supported.

7.2.2 Functions: scope

Functions cannot be defined within other functions. By default, functions are `extern` (they can be used anywhere). Functions can also be defined as `static` (they can only be used in the file in which they are defined).

7.2.3 Arrays of functions

An array of functions is a collection of identical functions. It is not the same as an array of function pointers (each of whose elements can point to a different function). A function array allows you to run different copies of the same function in parallel. Without this construct, the only safe way to run a function in parallel with itself would be to explicitly declare two functions with different names.

Function arrays allow functions to be copied and shared neatly. For example:

```
unsigned func[2](unsigned x, unsigned y)
{
    return (x + y);
}
```

Syntax

The syntax is a normal function declaration, with square brackets added to specify that this is an array declaration as well as a function declaration. The general form of a function array declaration is:

```
returnType Name[Size](parameterList);
```

7.2.4 Function arrays: example

```
set clock = external "P1";

// Function array prototype
unsigned func[2](unsigned x, unsigned y);

// Main program
void main(void)
{
    unsigned a, b, c, d, e, f;
    unsigned short r1, r2, r3, r4;
    unsigned result;

    par
    {
        a = 12;
        b = 22;
        c = 32;
        d = 42;
        e = 52;
        f = 62;
    }

    par
    {
        r1 = func[0](a, b);
        r2 = func[1](c, d);
    }

    par
    {
        r3 = func[0](e, f);
        r4 = func[1](r1, r2);
    }

    result = func[0](r3, r4);
}
```

```
// Function array definition
unsigned func[2](unsigned x, unsigned y)
{
    return (x + y);
}
```

7.2.5 Function arrays example with static variables

In the example below each function in the array has its own copy of the `static` variable 't'. Thus, if func[0]'s copy of 't' is modified, func[1]'s copy remains unaffected.

```
set clock = external "C1";

unsigned func[2](unsigned a, unsigned b)
{
    static unsigned t = 0;
    t++;
    return a + b + t;
}

void main(void)
{
    unsigned 7 p, q, r, s, t, u, v, w, x, y, z;

    par
    {
        p = 1;
        q = 1;
        r = 1;
        s = 1;
        t = 1;
        u = 1;
    }

    par
    {
        v = func[0](p, q);    // v = 3   (t in func[0] is 1)
        w = func[1](r, s);    // w = 3   (t in func[1] is 1)
    }

    x = func[0](t, u);      // x = 4   (t in func[0] is 2)
    y = func[0](v, w);      // y = 9   (t in func[0] is 3)

    z = func[1](x, y);      // z = 15  (t in func[1] is 2)
}
```

7.2.6 Function pointers

These are a very powerful, yet potentially confusing feature. In situations where any one of a number of functions can be called at a particular point, it is neater and more concise to use a function pointer, where the alternative might be a long if-else chain, or a long switch statement (see example).

Function pointers can be assigned with or without the address operator & (similar to assigning array addresses). Functions pointed to can be called with or without the indirection operator.

A function name can be assigned to a pointer without the &

```
p = addeven;
```

although the & format is clearer:

```
p = &addeven;
```

A function pointed to can be called by writing

```
(*chk)(a, b);
```

This can also be written in the shorthand form:

```
chk(a, b);
```

The first form is preferable, as it tips off anyone reading the code that a function pointer is being used.

7.2.7 Function pointers example

Consider the following program:

```
set clock = external "P1";

unsigned 1 check(short int *a, short int *b,
    unsigned 1 (*chk)(const short int *,
    const short int *));

unsigned 1 addeven(const short int *x, const short int *y);
unsigned 1 minuseven(const short int *x, const short int *y);
unsigned 1 diveven(const short int *x, const short int *y);
unsigned 1 modeven(const short int *x, const short int *y);

void main(void)
{
    short int m, n;
    unsigned 2 choice;
    unsigned 1 result;
    unsigned 1 (*p)(const short *, const short *);

    par
    {
        m = 19;
        n = 47;
    }

    do
    {
    switch (choice)
        {
        case 0:
            p = addeven;
            break;
        case 1:
            p = minuseven;
            break;
        case 2:
            p = diveven;
            break;
        case 3:
            p = modeven;
            break;
        default:
            delay;
```

```
            break;
        }

        par
        {
            result = check(&m, &n, p);
            choice++;
        }
    }
    while(choice);
}

unsigned 1 check(short int *a, short int *b,
    unsigned 1 (*chk)(const short int *,
    const short int *))
{
    return (*chk)(a, b);
}

unsigned 1 addeven(const short int *x, const short int *y)
{
    return (unsigned)(*x + *y)[0];
}

unsigned 1 minuseven(const short int *x, const short int *y)
{
    return (unsigned) (*x - *y)[0];
}

unsigned 1 diveven(const short int *x, const short int *y)
{
    return (unsigned) (*x / *y)[0];
}

unsigned 1 modeven(const short int *x, const short int *y)
{
    return (unsigned) (*x % *y)[0];
}
```

The function `addeven` checks whether the sum of two numbers is even. Similar checks are carried out by `minuseven` (difference of two numbers), `diveven` (division) and `modeven` (modulus). The function `check` simply calls the function whose pointer it receives, with the arguments it receives. This gives a consistent interface to the *xxx*`even`

functions. Pay close attention to the declaration of `check`, and of function pointer `p`. The parentheses around `*p` (and `*chk` in the declaration of `check`) are necessary for the compiler to make the correct interpretation.

Possible code optimization

Inside the main program body, `check` was called like this:

```
check(&m, &n, p);
```

It could have been written like this:

```
check(&m, &n, xxxeven);
```

eliminating the need for an additional pointer variable.

Here is the `main` section written using this form of expression:

```
void main(void)
{
    short int m, n;
    unsigned 2 choice;
    unsigned 1 result;

    par
    {
        m = 19;
        n = 47;
    }

    do
    {
        switch (choice)
        case 0:
            result = check(&m, &n, &addeven);
            break;
        case 1:
            result = check(&m, &n, &multeven);
            break;
        case 2:
            result = check(&m, &n, &diveven);
            break;
        case 3:
            result = check(&m, &n, &modeven);
            break;
        default:
            break;
            choice++;
     }
    while(choice);
}
```

7.2.8 Simultaneous function calls

In Handel-C, a function corresponds to a shared piece of hardware, which may only be used by one thread at a time. Simultaneous calls to a function, or even overlapping execution of a function, will cause problems.

You can check for simultaneous accesses to a function when you are debugging your code by using the **Detection of simultaneous function calls** option on the **Debug** tab in Project Settings, or the `-S+parfunc` option in the command line compiler.

You can ensure that the function usage does not overlap by declaring functions to be inline (so they are expanded whenever they are used) or by declaring an array of functions, one to be used in each parallel branch. This is illustrated in the example below.

Example

```
int func(int x, int y);

void main(void)
{
    int a, b, c, d, e, f, foo;
    // etc ...

    par
    {
        a = func(b, c);
        {
            b = foo;
            d = func(e, f); // NOT ALLOWED
        }
    }
    // etc ...
}


int func(int x, int y)
{
    if (x == y)
        delay;
    else
    {
        x = x % y;
    }
    x *= 10;

    return(x);
}
```

This is not allowed because part of the single function is used twice in the same clock cycle.

The code can be re-written to use inline functions, or an array of functions:

```
inline int func(x, y);

par
{
    a = func(b, c);
    {
        b = foo;
        d = func(e, f);
    }
}
```

or

```
int func[2](x, y);

par
{
    a = func[0](b, c);
    {
        b = foo;
        d = func[1](e, f);
    }
}
```

7.2.9 Multiple functions in a statement

Because each statement in Handel-C must take a single clock cycle, you cannot have multiple functions in a single statement.

Instead of

```
y = f(g(x));// illegal
```

you can write

```
z=g(x);
y=f(z);
```

Instead of

```
y = f(x) + g(z); // illegal
```

you can write:

```
par
{
    a = f(x);
    b = g(z);
}
y = a+b;
```

7.3 Introduction to macros

The Handel-C compiler passes source code through a standard C preprocessor before compilation allowing the use of `#define` to define constants and macros in the usual manner. Since the preprocessor can only perform textual substitution, some useful macro constructs cannot be expressed. For example, there is no way to create recursive macros using the preprocessor.

Handel-C provides additional macro support to allow more powerful macros to be defined (for example, recursive macro expressions). In addition, Handel-C supports shared macro expressions to generate one piece of hardware which is shared by a number of parts of the overall program similar to the way that procedures allow conventional C to share one piece of code between many parts of a conventional program.

7.3.1 Non-parameterized macro expressions

Non-parameterized macro expressions are of two types:

- simple constant equivalent to `#define`
- a constant expression

Constant

This first form of the macro is a simple expression. For example:

```
macro expr DATA_WIDTH = 15;

int DATA_WIDTH x;
```

This form of the macro is similar to the `#define` macro. Whenever `DATA_WIDTH` appears in the program, the constant 15 is inserted in its place.

Constant expression

To provide a more general solution, you can use a real expression. For example:

```
macro expr sum = (x + y) @ (y + z);

v = sum;
w = sum;
```

7.3.2 Parameterized macro expressions

Handel-C allows macros with parameters. For example:

```
macro expr add3(x) = x+3;

y = add3(z);
```

This is equivalent to the following code:

```
y = z + 3;
```

This form of the macro is similar to the `#define` macro in that every time the `add3()` macro is referenced, it is expanded in the manner shown above. In this example, an adder is generated in hardware every time the `add3()` macro is used.

7.3.3 select operator

The `select(...)` operator is used to mean 'select at compile time'. Its general usage is:

`select(`*`Expression1`*`, `*`Expression2`*`, `*`Expression3`*`)`

Expression1 must be a compile time constant. If *Expression1* evaluates to true then the Handel-C compiler replaces the whole expression with *Expression2*. If *Expression1* evaluates to false then the Handel-C compiler replaces the whole expression with *Expression3*.

Comparison with conditional operator

The difference between `select` and the conditional operator is seen in this example:

```
w = (width(x)==4 ? y : z);
```

The example generates hardware to compare the width of the variable `x` with 4 and set `w` to the value of `y` or `z` depending on whether this value is equal to 4 or not.

This is probably not what was intended because both `width(x)` and 4 are constants. What was probably intended was for the compiler to check whether the width of `x` was 4 and then simply replace the whole expression above with `y` or `z` according to the value. This can be written as follows:

```
w = select(width(x)==4 , y , z);
```

In this example, the compiler evaluates the first expression and replaces the whole line with either `w=y;` or `w=z;`. No hardware for the conditional is generated.

Combining with macros

This is more useful when macros are combined with this feature.

```
macro expr adjust(x, n) =
    select(width(x) < n, (0 @ x), (x <- n));

unsigned 4 a;
unsigned 5 b;
unsigned 6 c;

b = adjust(a, width(b));
b = adjust(c, width(b));
```

This example is for a macro that equalizes widths of variables in an assignment. If the right hand side of an assignment is narrower than the left hand side then the right hand side must be padded with zeros in its most significant bits. If the right hand side is wider than the left hand side, the least significant bits of the right hand side must be taken and assigned to the left hand side.

The `select(...)` operator is used here to tell the compiler to generate different expressions depending on the width of one of the parameters to the macro. The last two lines of the example could have been written by hand as follows:

```
b = 0 @ a;
b = c <- 5;
```

The macro comes into its own if the width of one of the variables changes. Suppose that during debugging, it is discovered that the variable `a` is not wide enough and needs to be 8 bits wide to hold some values used during the calculation. Using the macro, the only change required would be to alter the declaration of the variable `a`. The compiler would then replace the statement `b = 0 @ a;` with `b = a <- 5;` automatically.

This form of macro also comes in useful when variables of undefined width are used. If the compiler is used to infer widths of variables, it may be tedious to work out by hand which form of the assignment is required. By using the `select(...)` operator in this way, the correct expression is generated without you having to know the widths of variables at any stage.

7.3.4 ifselect

`ifselect` checks the result of a compile-time constant expression at compile time. If the condition is true, the following statement or code block is compiled. If false, it is dropped and an else condition can be compiled if it exists. Thus, whole statements can be selected or discarded at compile time, depending on the evaluation of the expression.

The `ifselect` construct allows you to build recursive macros, in a similar way to select. It is also useful inside replicated blocks of code as the replicator index is a compile-time constant. Hence, you can use ifselect to detect the first and last items in a replicated block of code and build pipelines.

Syntax

```
ifselect (condition)
    statement 1
[else
    statement 2]
```

Example

```
int 12 a;
int 13 b;
int undefined c;

ifselect(width(a) >= width(b))
    c = a;
else
    c = b;
```

c is assigned to by either a or b, depending on their width relationship.

Pipeline example

```
unsigned init;
unsigned q[15];
unsigned 31 out;

init = 57;
par (r = 0; r < 16; r++)
{
    ifselect(r == 0)
        q[r] = init;
    else ifselect(r == 15)
        out = q[r-1];
    else
        q[r] = q[r-1];
}
```

7.3.5 Recursive macro expressions

Preprocessor macros (those defined with #define) cannot generate recursive expressions. By combining Handel-C macros (those defined with macro expr) and the select(...) operator, recursive macros can express complex hardware simply. This type of macro is particularly important in Handel-C where the exact form of the macro may depend on the width of a parameter to the macro.

Variable sign extension example

When assigning a narrow signed variable to a wider variable, the most significant bits of the wide variable should be padded with the sign bit (MSB) of the narrow variable.

Value	4-bit representation	Conversion to 8-bit representation
-2	0b1110	0b11111110
6	0b0110	0b00000110

The following code suffices for a 4-bit to 8-bit conversion

```
int 8 x;
int 4 y;

x = y[3] @ y[3] @ y[3] @ y[3] @ y;
```

but it is tedious for variables that differ by a significant number of bits. It also does not deal with the case when the exact widths of the variables are not known. What is needed is a macro to sign extend a variable. For example:

```
macro expr copy(x, n) =
    select(n==1, x, (x @ copy(x, n-1)));

macro expr extend(y, m) =
    copy(y[width(y)-1], m-width(y)) @ y;

int a;
int b;  // Where b is known to be wider than a

b = extend(a, width(b));
```

The `copy` macro generates `n` copies of the expression `x` concatenated together. The macro is recursive and uses the `select(...)` operator to evaluate whether it is on its last iteration (in which case it just evaluates to the expression) or whether it should continue to recurse by a further level.

The `extend` macro concatenates the sign bit of its parameter `m-k` times onto the most significant bits of the parameter. Here, `m` is the required width of the expression `y` and `k` is the actual width of the expression `y`.

The final assignment correctly sign extends `a` to the width of `b` for any variable widths where `width(b)` is greater than `width(a)`.

7.3.6 Recursive macro expressions example

This example illustrates the generation of large quantities of hardware from simple macros. The example is a multiplier whose width depends on the parameters of the macro. Although Handel-C includes a multiplication operator as part of the language, this example serves as a starting point for generating large regular hardware structures using macros.

The multiplier generates the hardware for a single cycle long multiplication operation from a single macro. The source code is:

```
macro expr multiply(x, y) =  select(width(x) == 0, 0,
    multiply(x \\ 1, y << 1) +
    (x[0] == 1 ? y : 0));
a = multiply (b , c);
```

At each stage of recursion, the multiplier tests whether the bottom bit of the `x` parameter is 1. If it is then `y` is added to the 'running total'. The multiplier then recurses by dropping the LSB of `x` and multiplying `y` by 2 until there are no bits left in `x`. The overall result is an expression that is the sum of each bit in `x` multiplied by `y`. This is the familiar long multiplication structure. For example, if both parameters are 4 bits wide, the macro expands to:

```
a = ((b \\ 3)[0]==1 ? c<<3 : 0) +
    ((b \\ 2)[0]==1 ? c<<2 : 0) +
    ((b \\ 1)[0]==1 ? c<<1 : 0) +
    (b[0]==1 ? c : 0);
```

This code is equivalent to:

```
a = ((b & 8)==8 ? c*8 : 0) +
    ((b & 4)==4 ? c*4 : 0) +
    ((b & 2)==2 ? c*2 : 0) +
    ((b & 1)==1 ? c : 0);
```

which is a standard long multiplication calculation.

7.3.7 Shared expressions

By default, Handel-C generates all the hardware required for every expression in the whole program. This can mean that large parts of the hardware are idle for long periods. Shared expressions allow hardware to be shared between different parts of the program to decrease hardware usage.

The shared expression has the same format as a macro expression but does not allow recursion. You can use recursive macro expressions or `let...in` to generate recursive shared expressions.

Example

```
a = b * c;
d = e * f;
g = h * i;
```

This code generates three multipliers. Each one will only be used once and none of them simultaneously. This is a massive waste of hardware. You can improve the hardware efficiency with a shared expression:

```
shared expr mult(x, y) = x * y;

a = mult(b, c);
d = mult(e, f);
g = mult(h, i);
```

In this example, only one multiplier is built and it is used on every clock cycle. If speed is required, you can build three multipliers executing in parallel.

Warning

It is not always the case that less hardware is generated by using shared expressions because multiplexors may need to be built to route the data paths. Some expressions use less hardware than the multiplexors associated with the shared expression.

7.3.8 Using recursion to generate shared expressions

Although shared expressions cannot use recursion directly, macro expressions can be used to generate hardware which can then be shared using a shared expression. For example, to share a recursive multiplier you could write:

```
macro expr multiply(x, y) =  select(width(x) == 0, 0,
    multiply(x \\ 1, y << 1) +
    (x[0] == 1 ? y : 0));

shared expr mult(x, y) = multiply(x, y);

a = mult(b, c);
d = mult(e, f);
```

The macro expression builds a multiplier and the shared expression allows that hardware to be shared between the two assignments.

7.3.9 Restrictions on shared expressions

Shared expressions must not be shared by two different parts of the program on the same clock cycle. For example:

```
shared expr mult(x, y) = x * y;

par
{
    a = mult(b, c);
    d = mult(e, f); // NOT ALLOWED
}
```

This is not allowed because the single multiplier is used twice in the same clock cycle.

You need to ensure that shared expressions in parallel branches are not shared on the same clock cycle.

7.3.10 let ... in

`let` and `in` allow you to declare macro expressions within macro expressions. In this way, complex macros may be broken down into simple ones, whilst still being grouped together in a single block of code. They also provide easy sharing of recursive macros.

The `let` keyword starts the declaration of a local macro; the `in` keyword ends the declaration and defines its scope.

Example

```
macro expr Fred(x) =
  let macro expr y = x*2; in
   y+3;  // Returns x*2+3
```

The top line defines the macro name and parameters. The second line defines `y` within the macro definition. The last line expresses the value of the macro in full.

Independent `let …in` definitions

```
macro expr op(a, b) =
    let macro expr t2(x) = x * 2; in
    let macro expr d3(x) = x / 3; in
    let macro expr t4(x) = x * 4; in
        t2(a) + d3(b) + t4(a - b) + t2(b - a);
```

is equivalent to writing

```
macro expr op(a, b) = (a * 2) + (b / 3) + ((a-b) * 4) + ((b-a) * 2);
```

Related `let …in` definitions

```
macro expr op(a, b) =
    let macro expr sum(x, y) = x + y; in
    let macro expr mult(x, y) = x * sum(x, y); in
        mult(a, b) - (b * b);
```

sum is defined within the macro definition, then `mult` is defined using `sum`. This example is equivalent to:

```
macro expr op(a, b) = (a * (a + b)) - (b * b);
```

Shared recursive macro

A recursive multiplier illustrating the way in which `let…in` can be used to share recursive macros.

```
shared expr mult(p, q) =
    let macro expr multiply(x, y) =
        select(width(x) == 0, 0, multiply(x \\ 1, y << 1)
        + (x[0] == 1 ? y : 0)); in
        multiply(p, q)
```

Scope of definitions

The inner macros are not accessible outside the outer macro

```
{
    chanout <unsigned 16> och;
    int 16 i, j, k;
    {
        macro expr Cube(x) =
            let macro expr Sqr(x) = x*x; in
                x * Sqr(x)
        i = Cube(3) // Correct use
        j = Sqr(3)  // Error - out of scope
    }
    k = Cube(2);    //Error - out of scope
}
```

7.3.11 Macro procedures

Macro procedures may be used to replace complete statements to avoid tedious repetition while coding. A single macro procedure can be expanded into a complex block of code. It generates the hardware for the statement each time it is referenced.

The general syntax of macro procedures is:

```
macro proc Name(Params) Statement
```

Macros may be prototyped (like functions). This allows you to declare them in one file and use them in another. A macro prototype consists of the name of the macro plus a list of the names of its parameters. E.g.

```
macro proc work(x, y);
```

If you have local or static declarations within the macro procedure, a copy of the variable will be created for each copy of the macro.

Macro procedures that don't take any parameters require an empty parameter list. For example:

```
macro proc MyMacro ();
```

Example

```
macro proc output(x, y)
    {
        out ! x;
        out ! y;
    }

output(a + b, c * d);
output(a + b, c * d);
```

This example writes the two expressions `a+b` and `c*d` twice to the channel `out`. This example also illustrates that the statement may be a code block - in this case two instructions executed sequentially.

It expands to 4 channel output statements.

7.3.12 Macro procedures compared to pre-processor macros

Macro procedures differ from preprocessor macros in that they are not simple text replacements. The statement section of the definition must be a valid Handel-C statement.

The following code is valid as a `#define` pre-processor macro but not as a macro procedure:

```
#define test(x,y) if (x!=(y<<2)) // not valid as a macro procedure as not a
complete statement

test(a,b)
{
    a++;
}
else
{
    b++;
}
```

Incomplete statements will not compile as macro procedures:

```
macro proc test(x,y) if (x!=(y<<2)) // Incomplete statement, will not
compile
```

A complete statement will not successfully replace an incomplete one:

```
macro proc test(x,y) if (x!=(y<<2)); // Complete statement will compile

test(a,b)  // will expand to if (x!=(y<<2));
{
    a++;
}
else      // this else has no associated if
{
    b++;
}
```

Here, the macro procedure is not defined to be a complete statement so the Handel-C compiler generates an error. This restriction provides protection against writing code which is generally unreadable and difficult to maintain.

8 Introduction to timing

A Handel-C program executes with one clock source for each `main` statement. It is important to be aware exactly which parts of the code execute on which clock cycles. This is not only important for writing code that executes in fewer clock cycles but may mean the difference between correct and incorrect code when using Handel-C's parallelism. Experienced programmers can immediately tell which instructions execute on which clock cycles. This information becomes very important when your program contains multiple interacting parallel processes.

Knowing about clock cycles also becomes important when considering interfaces to external hardware. It is important to understand timing issues before moving on to implementing such interfaces because it is likely that the external device will place constraints on when signals should change.

Avoiding certain constructs has a dramatic influence on the maximum clock rate that your Handel-C program can run at.

8.1 Statement timing

The basic rule for working out the number of cycles used in a Handel-C program is:

Assignment and delay take 1 clock cycle. Everything else is free.

- One clock cycle is used every time you write an assignment statement or a `delay` statement. `releasesema` also uses one clock cycle.
  A special case statement is supported of the form:
  `a = f(x);`
  to allow function calls which take multiple clock cycles.
- Channel communications use one clock cycle in the same clock domain if both ends are ready to communicate. If one of the branches is not ready for the data transfer then execution of the other branch waits until both branches become ready.
- You can write any other piece of code and not use any clock cycles to execute it.

8.1.1 Example timings

Statements

```
x = y;
x = (((y * z) + (w * v))<<2)<-7;
```

Each of these statements takes one clock cycle.

Notice that even the most complex expression can be evaluated in a single clock cycle. Handel-C builds the combinational hardware to evaluate such expressions; they do not need to be broken down into simpler assembly instructions as would be the case for conventional C.

Parallel statements

```
par
{
    x = y;
    a = b * c;
}
```

This code executes in a single cycle because each branch of the parallel statement takes only one clock cycle. This example illustrates the benefits of parallelism. You can have as many non-interdependent instructions as you wish in the branches of a parallel statement. The total time for execution is the length of time that the longest branch takes to execute. For example:

```
par
{
    x = y;
    {
        a = b;
        c = d;
    }
}
```

This code takes two clock cycles to execute. On the first cycle, `x = y` and `a = b` take place. On the second clock cycle, `c = d` takes place. Since both branches of the `par` statement must complete before the `par` block can complete, the first branch delays for one clock cycle while the second instruction in the second branch is executed.

While loop

```
x = 5;
while (x>0)
{
    x--;
}
```

This code takes a total of 6 clock cycles to execute. One cycle is taken by the assignment of 5 to `x`. Each iteration of the `while` loop takes 1 clock cycle for the assignment of `x-1` to `x` and the loop body is executed 5 times. The condition of the `while` loop takes no clock cycles as no assignment is involved.

For loop

```
for (x = 0; x < 5; x ++)
{
    a += b;
    b *= 2;
}
```

This code has an almost direct equivalent:

```
{
    x = 0;
    while (x<5)
    {
        a += b;
        b *= 2;
        x ++;
    }
}
```

This code takes 16 clock cycles to execute. One is required for the initialization of `x` and three for each execution of the body. Since the body is executed 5 times, this gives a total of 16 clock cycles.

Decision

```
if (a>b)
{
    x = a;
}
else
{
    x = b;
}
```

This code takes exactly one clock cycle to execute. Only one of the branches of the `if` statement is executed, either `x = a` or `x = b`. Each of these assignments takes one clock cycle. Notice again that no time is taken for the test because no assignment is made. A slightly different example is:

```
if (a>b)
{
    x = a;
}
```

Here, if `a` is not greater than `b`, there is no `else` branch. This code therefore takes either 1 clock cycle if `a` is greater than `b` or no clock cycles if `a` is not greater than `b`.

Channels

Channel timings can be complex. The simplest example is with a channel `link` of `fifolength` 0 (default):

```
chan unsigned 8 link;
par
{
    link ! x; // Transmit
    link ? y; // Receive
}
```

This code takes a single clock cycle to execute because both the transmitting and receiving branches are ready to transfer at the same time. All that is required is the assignment of `x` to `y` which, like all assignments, takes 1 clock cycle. A more complex example is:

```
chan unsigned 8 link;
par
{
    {               // Parallel branch 1
        a = b;
        c = d;
        link ! x;
    }

    link ? y;       // Parallel branch 2
}
```

Here, the first branch of the `par` statement takes three clock cycles to execute. However, the second branch of the `par` statement also takes three clock cycles to execute because it must wait for two cycles before the transmitting branch is ready. The usage of clock cycles is as follows:

Cycle	Branch 1	Branch 2
1	`a = b;`	`delay`
2	`c = d;`	`delay`
3	Channel output	Channel input

FIFOs

FIFOs add another layer of complexity.

```
chan unsigned link_FIFO with {fifolength=4};
int i = 0;

par
{
    while(1)
    {
        i++;           //Cannot be in parallel to channel write
                       //Do not change a variable in parallel with sending
it
        link_FIFO ! i; // Parallel branch 1
    }
    // Parallel branch 2
    a = b; //Parallel code: used here instead of delay
    c = d;
    link_FIFO ? y;
    }
}
```

Here, the write branch of the `par` statement takes two clock cycles to execute and the read branch takes three clock cycles to execute. If it were a simple channel, the write branch would have to wait until the channel had been read, before it could write the next value of `i`. However, because it is a FIFO, the write branch can keep writing until the FIFO is full. On the third clock cycle, the read branch reads the first value from the FIFO.

When the FIFO is full the first branch must wait until the FIFO is read from before it can write to it again.

The precise timing of FIFOs depends on many different factors. The throughput will be close to one word per cycle for sufficiently large FIFOs.

FIFO: channel and FIFO comparison code

This example shows a loop using a channel to communicate between two processes.

```
Process A:
static unsigned 4 Val = 1;
while(1)
{
    Val = Val[2:0]@Val[3];
    MyChan ! Val; // Send
    delay;
}
```

```
Process B:
static unsigned 4 Count;
while(1)
{   // wait 0 or more cycles
    while (Count != 0)
    {
        Count--;
    }
    MyChan ? Count; // Receive
    delay;
}
```

The delay statements in each process always take place on the same clock cycle in the same clock domain.

Example with FIFO

This shows the same process, but using a FIFO with `fifolength` 4. The loop in process A would execute 4 times without pausing and then run after each time process B reads from the FIFO.

```
chan myFIFO with {fifolength = 4};


Process B:
static unsigned 4 Count;
while(1)
{   // wait 0 or more cycles
    while (Count != 1)
    {
        Count--;
    }
    MyFIFO ? Count; // Receive
    delay;
}

Process A:
static unsigned 4 Val = 1;
while(1)
{
    Val = Val[2:0]@Val[3];
    MyFIFO ! Val; // Send
    delay;
}
```

See the summary of statement timings for more detail.

8.1.2 Statement timing summary

Statement	**Timing**
`{...}`	Sum of all statements in sequential block
`par {...}`	Length of longest branch in block
`Function``()`, `break`, `goto`, `continue`	No clock cycles
`return(`*`Expression`*`);`	1 clock cycle if *`Expression`* is assigned on return, otherwise none.
`Variable` `=` *`Expression`*`;`	1 clock cycle
`Variable` `++;`	1 clock cycle
`Variable` `--;`	1 clock cycle
`++` *`Variable;`*	1 clock cycle
`--` *`Variable`*`;`	1 clock cycle
`Variable` `+=` *`Expression`*`;`	1 clock cycle
`Variable` `-=` *`Expression`*`;`	1 clock cycle
`Variable` `*=` *`Expression`*`;`	1 clock cycle
`Variable` `/=` *`Expression`*`;`	1 clock cycle
`Variable` `%=` *`Expression`*`;`	1 clock cycle
`Variable` `<<=` *`Constant`*`;`	1 clock cycle
`Variable` `>>=` *`Constant`*`;`	1 clock cycle
`Variable` `&=` *`Expression`*`;`	1 clock cycle
`Variable` `|=` *`Expression`*`;`	1 clock cycle
`Variable` `^=` *`Expression`*`;`	1 clock cycle
`Channel` `?` *`Variable`*`;`	1 clock cycle when transmitter is ready (in same clock domain)
`Channel` `!` *`Expression`*`;`	1 clock cycle when receiver is ready (in same clock domain)
`if (`*`Expression`*`) {...} else {...}`	Length of executed branch
`while (`*`Expression`*`) {...}`	Length of loop body * number of iterations
`do {...} while (`*`Expression`*`);`	Length of loop body * number of iterations
`for (`*`Init`*`;` *`Test`*`;` *`Iter`*`) {...}`	Length of Init + (Length of body + length of *`Iter`*) * number of iterations
`switch (`*`Expression`*`) {...}`	Length of executed case branch

`prialt {...}`	1 clock cycle for case communication when other party is ready plus length of executed case branch *or* length of default branch if present and no communication case is ready *or* infinite if no default branch and no communication case is ready
`releasesema();`	1 clock cycle
`delay;`	1 clock cycle

The Handel-C compiler may insert `delay` statements to break combinational loops.

8.2 Avoiding combinational loops

If you wish to wait for a variable to be modified in a parallel process before continuing, you might write:

```
while (x!=3); // WARNING!!
```

This is bad Handel-C code because it generates a combinational loop in the logic (This is because of the way that Handel-C expressions are built to evaluate in zero clock cycles.)

This is easier to see if it is written as

```
while (x!=3)
{
    // wait until x == 3
}
```

This empty loop must be broken by changing the code to:

```
while (x!=3)
{
    delay;
}
```

This code takes no longer to execute but does not contain a combinational loop because of the clock cycle delay in the loop body.

The Handel-C compiler spots this form of error, inserts the `delay` statement, and generates a warning. It is considered better practice to include the `delay` statement in the code to make it explicit

Similar problems occur with `do ... while` loops and `switch` statements in similar circumstances. `for` loops with no iteration step can also cause combinational loops.

Further combinational loop code example

Code may look correct but still include an empty loop. For example:

```
while (x!=3)
{
    if (y>z)
    {
        a++;
    }
}
```

This `if` statement may take zero clock cycles to execute if `y` is not greater than `z` so even though this loop body does not look empty a combinational loop is still generated. This is more obvious written as

```
while (x!=3)
{
    if (y>z)
    {
        a++;
    }
    else
    {
            // do nothing
    }
}
```

The solution is to add the `else` part of the if construct as follows:

```
while (x!=3)
{
    if (y>z)
    {
        a++;
    }
    else
    {
        delay;
    }
}
```

8.3 Parallel access to variables

The rules of parallelism state that the same variable must not be accessed from two separate parallel branches. This avoids resource conflicts on the variables.

The rule may be relaxed to state that the same variable must not be assigned to more than once on the same clock cycle but may be read as many times as required. This gives powerful programming techniques. For example:

```
par
{
    a = b;
    b = a;
}
```

This code swaps the values of a and b in a single clock cycle.

Since exact execution time may be run-time dependent, the Handel-C compiler cannot determine when two assignments are made to the same variable on the same clock cycle. You should therefore check your code to ensure that the relaxed rule of parallelism is still obeyed.

Example

Using this technique, a four-place queue can be written:

```
while(1)
{
    par
    {
        int x[3];

        x[0] = in;
        x[1] = x[0];
        x[2] = x[1];
        out = x[2];
    }
}
```

The value of out is the value of in delayed by 4 clock cycles. On each clock cycle, values will move one place through the x array. For example:

Clock	in	x[0]	x[1]	x[2]	out
1	5	0	0	0	0
2	6	5	0	0	0
3	7	6	5	0	0
4	8	7	6	5	0
5	9	8	7	6	5
6	10	9	8	7	6
7	11	10	9	8	7
8	12	11	10	9	8
9	13	12	11	10	9

8.4 Detailed timing example

This is an analyzed example that generates signals tied to real-world constraints. It shows the generation of signals for a real time clock. The signals required are for microseconds, seconds, minutes and hours.

The hardware generated will eventually be driven from an external clock. In order to write the program, the rate of this clock must be known. It has been assumed to be 5 MHz on pin P1.

The loop body takes one clock cycle to execute. The `Count` variable is used to divide the clock by 5 to generate microsecond increments. As each variable wraps round to zero, the next time step up is incremented.

```
set clock = external "P1";
void main(void)
{
    unsigned 20 MicroSeconds;
    unsigned 6 Seconds;
    unsigned 6 Minutes;
    unsigned 16 Hours;
    unsigned 3 Count;

    par
    {
        Count = 0;
        MicroSeconds = 0;
        Seconds = 0;
        Minutes = 0;
        Hours = 0;
    }
    while (1)
    {
        if (Count!=4)
            Count++;
        else
        par
       {
            Count = 0;
            if (MicroSeconds!=999999)
                MicroSeconds++;
            else
            par
            {
                MicroSeconds = 0;
                if (Seconds!=59)
                    Seconds++;
                else
                par
                {
                    Seconds = 0;
                    if (Minutes!=59)
                        Minutes++;
                    else
                    par
                   {
```

```
                        Minutes = 0;
                        Hours++;
                    }
                }
            }
        }
    }
}
```

8.5 Time efficiency of Handel-C hardware

Handel-C requires that the clock period for a program is longer than the longest path through combinational logic in the whole program. This means that, for example, once FPGA or PLD place and route has been completed, the maximum clock rate for the system can be calculated from the reciprocal of the longest path delay in the circuit.

For example, suppose the FPGA place and route tools calculate that the longest path delay between flip-flops in a design is 70ns. The maximum clock rate that that circuit should be run at is then 1/70ns = 14.3MHz.

If this calculated rate is not fast enough for the system performance or real time constraints you can optimize your program to reduce the longest path delay and increase the maximum possible clock rate. You can also use the retiming option to try and match your target clock rate.

One standard technique for optimizing efficiency is to use pipelining.

8.5.1 Reducing logic depth

Certain operations in Handel-C combine to produce deep logic. Deep logic results in long path delays in the final circuit so reducing logic depth should increase clock speed.

Guidelines for reducing logic depth

- Division and modulo operators produce the deepest logic. Multiplication also produces deep logic. A single cycle divide, mod or multiplier produces a large amount of hardware and long delays through deep logic so you should avoid using them wherever possible.
- Most common division and multiplications can be done with the shift operators. Also consider using a long multiplication with a loop, shift and add routine or a pipelined multiplier.
- Most common modulo operations can be done with the AND operator.
- Wide adders require deep logic for the carry ripple. Consider using more clock cycles with shorter adders.
- Avoid greater than and less than comparisons - they produce deep logic.

- Reduce complex expressions into a number of stages.
- Avoid long strings of empty statements. Empty statements result from, for example, missing `else` conditions from `if` statements.

Adder example

To reduce a single, 8-bit wide adder to 3, narrower adders:

```
unsigned 8 x;
unsigned 8 y;
unsigned 5 temp1;
unsigned 4 temp2;

par
{
    temp1 = (0@(x<-4)) + (0@(y<-4));
    temp2 = (x \\ 4) + (y \\ 4);
}
x = (temp2+(0@temp1[4])) @ temp1[3:0];
```

Comparison example

```
while (x<y)
{
    ......
    x++;
}
```

can be replaced with:

```
while (x!=y)
{
    ......
    x++;
}
```

The == and != comparisons produce much shallower logic although in some cases it is possible to remove the comparison altogether. Consider the following code:

```
unsigned 8 x;

x = 0;
do
{
    ......
    x = x + 1;
} while (x != 0);
```

This code iterates the loop body 256 times but it can be re-written as follows:

```
unsigned 9 x;

x = 0;
do
{
    ......
    x = x + 1;
} while (!x[8]);
```

By widening x by a single bit and just checking the top bit, we have removed an 8-bit comparison.

Complex expression example

```
x = a + b + c + d + e + f + g + h;
```

reduces to:

```
par
{
    temp1 = a + b;
    temp2 = c + d;
    temp3 = e + f;
    temp4 = g + h;
}
par
{
    temp1 = temp1 + temp2;
    temp3 = temp3 + temp4;
}
x = temp1 + temp3;
```

This code takes three clocks cycles as opposed to one but each clock cycle is much shorter and so the rest of the circuit should be speeded up by the faster clock rate permitted.

Empty statement example

```
if (a>b)
    x++;
if (b>c)
    x++;
if (c>d)
    x++;
if (d>e)
    x++;
if (e>f)
    x++;
```

If none of these conditions is met then all the comparisons must be made in one clock cycle. By filling in the `else` statements with `delays`, the long path through all these `if` statements can be split at the expense of having each `if` statement take one clock cycle whether the condition is true or not.

8.5.2 Pipelining

A classic way to increase clock rates in hardware is to pipeline. A pipelined circuit takes more than one clock cycle to calculate any result but can produce one result every clock cycle. The trade off is an increased latency for a higher throughput so pipelining is only effective if there is a large quantity of data to be processed: it is not practical for single calculations.

Pipelined multiplier example

```
unsigned 8 sum[8];
unsigned 8 a[8];
unsigned 8 b[8];
//ina.dat is a data file. You must provide your own
chanin inputa with {infile = "ina.dat"};
chanin inputb with {infile = "ina.dat"};
chanout output with {outfile = "out.dat"};

par
{
    while(1)
        inputa ? a[0];

    while(1)
        inputb ? b[0];

    while(1)
        output ! sum[7];

    while(1)
    {
        par
        {
            macro proc level(x)
                par
                {
                    sum[x] = sum[x - 1] +
                        ((a[x][0] == 0) ? 0 : b[x]);
                    a[x] = a[x - 1] >> 1;
                    b[x] = b[x - 1] << 1;
                }

            sum[0] = ((a[0][0] == 0) ? 0 : b[0]);
                par ( i=1; i <=7; i++)
                {
                    level (i);
                }
        }
    }
}
```

This multiplier calculates the 8 LSBs of the result of an 8-bit by 8-bit multiply using long multiplication. The multiplier produces one result per clock cycle with a latency of 8 clock cycles. This means that although any one result takes 8 clock cycles, you get a throughput of 1 multiply per clock cycle. Since each pipeline stage is very simple, combinational logic is shallow and a much higher clock rate is achieved than would be possible with a complete single cycle multiplier.

At each clock cycle, partial results pass through each stage of the multiplier in the `sum` array. Each stage adds on 2^n multiplied by the `b` operand if required. The LSB of the `a` operand at each stage tells the multiply stage whether to add this value or not. Stages are generated with a macro procedure instantiated several times using a replicator

Operands are fed in on every clock cycle through `a[0]` and `b[0]`. Results appear 8 clock cycles later on every clock cycle through `sum[7]`.

9 Clocks overview

You can have multiple clocks interfacing with your design. Each `main()` function must be associated with a single clock. If you have more than one `main` function in the same source file, they must all use the same clock.

Clocks may be fed from expressions (internal clocks) or fed from a pin (external clocks).

The current clock may be referred to using the keyword `__clock`

You can specify the maximum delay in MHz allowed between components fed from a clock by using the rate specification.

The general syntax of the clock specification is:

```
set clock = Location with {Rate_spec, periodSpec};
```

If you are communicating between clock domains, you also need to set timing specifications (`resolutiontime` or `minperiod`). These control the synchronization hardware generated.

You must specify a clock. When generating simulation output, a dummy clock such as '`set clock = external "P1";`' is valid.

9.1 Locating the clock

Since each Handel-C `main()` code block generates synchronous hardware, a single clock source is required for each one.

The general syntax of the clock specification is:

```
set clock = Location;
```

`Location` may be any of the following:

Location	Meaning
`internal` *`Expression`*	Clock from expression
`internal_divide` *`Expression Factor`*	Clock from expression with integer division
`external [`*`Pin`*`]`	Clock from device pin
`external_divide [`*`Pin`*`]` *`Factor`*	Clock from device pin with integer division

9.1.1 External clocks

External clocks may be accessed by associating the clock with a specific pin using `set clock external = "pin_Name"` or `set clock external_divide = "pin_Name" factor`, where the `external_divide` keyword is a constant integer. For example:

```
set clock = external "P35";
set clock = external_divide "P35" 3;
set clock = external_divide 3;
```

The first of these examples specifies a clock taken from pin P35. The second specifies a clock taken from pin P35 which is divided on the FPGA/PLD by a factor of 3. The third option shows a clock divided by 3 with no pin number specified.

When the pin number is omitted, the place and route tools will choose an appropriate pin. Omitting pin specifications can speed up the clock rate of the design.

You can also define an `interface` that reads an external clock. If the clock is associated with a specific pin, you can use the `interface` sort `bus_in`. You would only need to do this if the external clock has been divided, otherwise you can use the intrinsic `__clock`.

Example

```
interface bus_in(unsigned 1 in with {clockport=1})
    InputBus() with {data={"Pin1"}};
set clock = external_divide "Pin1" 4;
```

You can now use `InputBus.in` to get an undivided external clock.

9.1.2 Internal clocks fed from expressions

You can set the clock to be any expression or any expression divided by a given factor.

```
set clock = internal <Expression>;

set clock = internal_divide <Expression> factor;
```

The clock division factor specified with the `internal_divide` keyword must be a constant integer.

Example

This allows you to set the clock to a value read from an `interface`.

```
interface port_in(unsigned 1 clk with {clockport = 1}) ClockPort();
set clock = internal ClockPort.clk;
```

9.2 Current clock

The current clock used by a function can be referenced using the keyword `__clock`. This allows the function to pass the current clock to an external interface. The value of the system variable `__clock` will be the value after any divide. The clock may be an internal or an external clock.

Example

The code below shows the assignment of the current clock to a port in an interface.

```
interface reg32x1k() registers(unsigned addr=address,
    unsigned data=data_in, unsigned 1 clk = __clock,
    unsigned out = write);
```

9.3 Multiple clock domains

You can have multiple clock domains in your Handel-C design by declaring more than one `main()` function. If you have more than one `main()` function in the same source file, they must all use the same clock. The clock is defined in each file using the `set clock` construct.

You can communicate between clock domains by:

- using channels with time constraints set on the clock
- using a defined custom interface. You cannot use multiple clock domains within the pre-defined Handel-C interface sorts.

Variables, signals and functions cannot be written to by one clock domain and read in another.

Communicating between clock domains means that you need to consider metastability issues.

If you reset one clock domain without synchronously resetting any clock domains that it communicates with, the communicating channels will go to an undefined state.

9.3.1 Channels communicating between clock domains

Channels that connect between clock domains can only be written to in a single domain and read from in a single domain. Their first use defines their direction and the domains

in which they transmit and receive. If you attempt to re-use the channel in a different direction or to or from a different clock domain, the compiler generates an error.

Channels used between clock domains must be defined in one file and then declared as `extern` in another.

The timing between domains is unspecified, but the transmission is guaranteed to occur provided metastability is resolved. If `fifolength` is 0, both sides will wait until the transmission is certain to complete. Otherwise, the channel will write as soon as the FIFO is ready (and has space) and read as soon as the FIFO is ready (and isn't empty).

If you use channels to communicate between clock domains you must specify the `rate` and the `resolutiontime` for both clocks.

Most cases will be dealt with by setting the resolution time to three-quarters of the clock period.

Example

For a 10ns clock

```
set clock = external "A22" with {rate=100, resolutiontime=7.5};
```

Note that the `rate` is in MHz and the `resolutiontime` is in nanoseconds.

If you need to adjust the channel timing due to latency issues, you may do so by adjusting the `resolutiontime` and the number of flip-flops used to prevent metastability being propagated through the circuit.

If the resolution time is set incorrectly then intermittent failures due to metastability may cause the generated hardware to be unreliable. You must test channels communicating between clock domains extremely thoroughly (unless you know that `resolutiontime` is sufficiently long to guarantee an acceptable probability of failure).

Timing issues for channels communicating between clock domains

The timing of channels across clock domains is unspecified. The values read into `i` and `j` may differ in the example below, as the reads may not complete on the same clock cycle.

Domain 1:

```
set clock = external with { paranoia=1, minperiod=2.0,
unconstrainedperiod=9, rate=101 };

chan <unsigned 8> ch1, ch2;

unsigned 8 i = 0;
while (1) par
{
  while ( 1 ) i ++;
  ch1 ! i;
  ch2 ! i;
}
```

Domain 2:

```
set clock = external with { minperiod=2.0, unconstrainedperiod=10, rate=100
}

extern chan ch1, ch2;

unsigned 8 i, j;

while ( 1 )
{
  par {
    ch1 ? j;
    ch2 ? i;
  }
}
```

Channel communication example

This example uses a channel to communicate between two clock domains. One clock domain runs at half the speed of the other.

```
/*
 * File: receive.hcc: primary clock domain
 */

set clock = external "A22" with { rate=100, resolutiontime = 7.5 };
unsigned 4 result;
interface bus_out() O(unsigned o = result) with {warn = 0};

//channel defined in other file
extern chan unsigned 4 ReturnData;

void main(void)
{
    while(1)
    {
        delay;
        //program will wait until data received
        ReturnData ? result;
    }
}
/*
 * File: transmit.hcc:secondary clock domain,
 * running at half the speed of the primary one
 */

set clock = external_divide "R25" 2;

chan unsigned 4 ReturnData; //channel must have global scope

void main(void)
{
    static unsigned 4 x;

    while(1)
    {
        x++;
        ReturnData ! x;
    }
}
```

Example: channels between clock domains

```
//File: transmit.hcc
chan 8 c ; // channel must have global scope

set clock = external "P1" with (paranoia =2);
void main(void)
{
    int 8 x, y;
    c ! x; //program will wait until data successfully transmitted
    c ! y;
}

//File: receive.hcc
extern chan c;

set clock = external "P2";
void main(void)
{
    int 8 p, q;

    c ? p;
    c ? q;
}
```

Managing channel timing

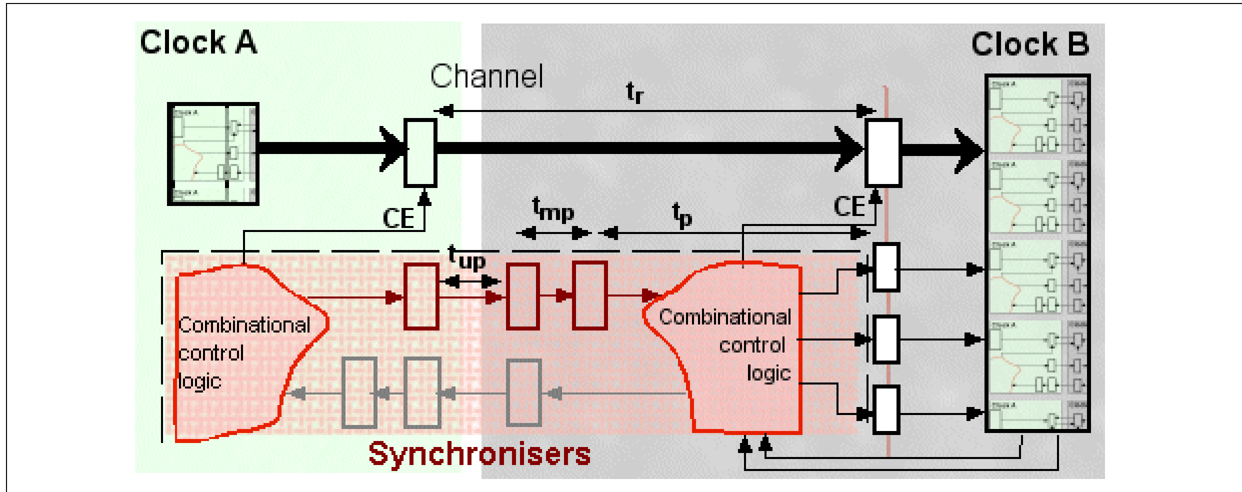
The timing of channels between clock domains is controlled by:

- the number of flip-flops used to resolve metastability (this is set by the `paranoia` specification, defaulting to 1)
- the value of `resolutiontime` (how long it is before you sample a signal

  OR

  the value of `minperiod` (how long is available before the signal is moved on)

- the value of `unconstrainedperiod`. This is the timing constraint on the compiler generated synchronizing control paths.



t_r time to transfer between domains (`paranoia` in domain B + 1) X t_p

t_p clock period in domain B

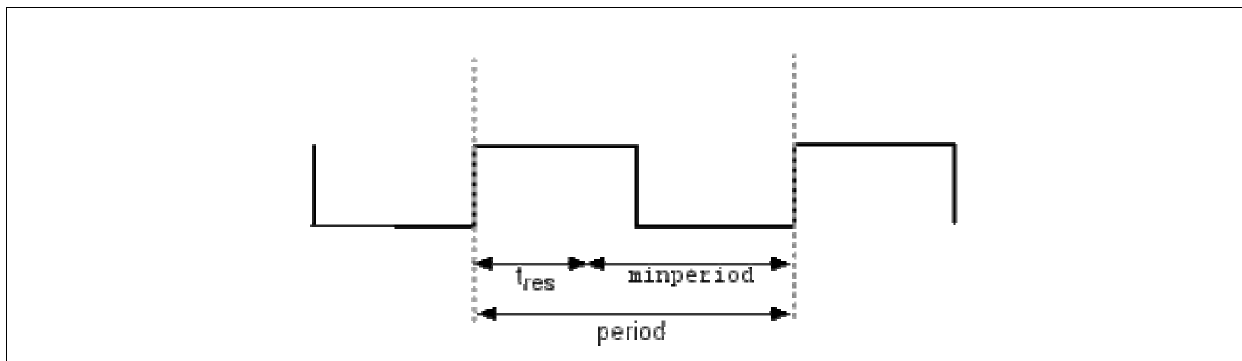
t_{up} `unconstrainedperiod`

t_{mp} `minperiod`

Using clock specifications to manage timing between clock domains
You can use clock specifications to specify the timing of the synchronization hardware. Set the values on a clock to affect the timing of ALL channels to and from that domain.

```
set clock = external "C43" with {rate = 40,  resolutiontime  = 20};
```

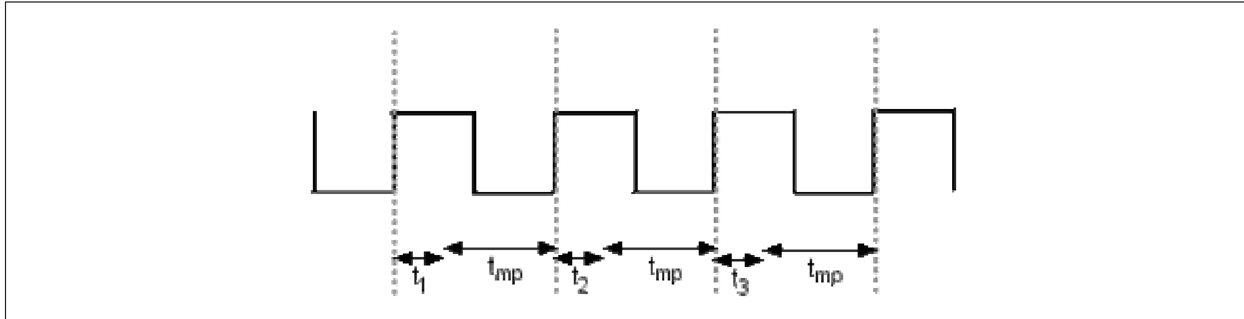
This gives a clock period of 25ns. It assumes that 20ns is required for the control signal to stabilize, leaving 5ns for it to be routed onwards.



RESOLUTION TIME AND PROPAGATION TIME IN ONE CLOCK TICK

If it takes longer for the data to stabilize, you can increase the number of flip-flops used to stabilize the data by setting the `paranoia` specification. In this case, the stabilization time in each clock period is (`resolutiontime` / `paranoia`)

If paranoia is set to 3, then the resolution time required in each clock tick is 1/3 the value of `resolutiontime`, giving a larger possible value for data to be routed on.

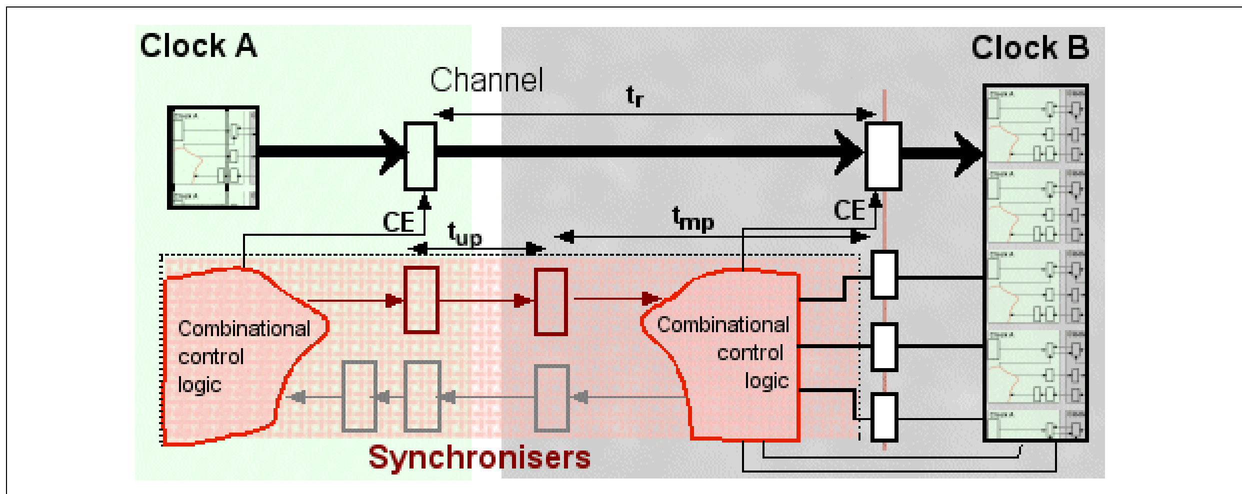


Speed versus metastability

When you increase the `paranoia` specification on the clock domains, you increase the latency of channels into the domain. The value of the `paranoia` specification sets the number of flip-flops used to give the synchronization data time to stabilize. The higher the value of `paranoia`, the more stable the data and the greater the latency.

Setting paranoia to 0 to decrease latency

If you know there is not going to be a metastability problem (e.g, the clock domains you are working with have a low clock rate or are synchronous with each other), you may choose to set `paranoia` to 0. In this case, you must use the `minperiod` constraint rather than the `resolutiontime` constraint.



t_r time to transfer between domains (clock period in domain B when `paranoia` = 0)

t_{up} `unconstrainedperiod`

t_{mp} `minperiod`

Latency between clock domains

The latency of channels between clock domains is unpredictable. It is dependent on:

- the value of `resolutiontime`
- the value of `paranoia`
- the value of `unconstrainedperiod`

For FIFOs whose size is an exact power of two, latency is higher.

In addition, it may be affected by

- the way a channel has been implemented in hardware the sequence of communications sent across the channel
- the number of clock cycles elapsed since power-up
- environmental factors such as temperature and supply voltage
- the individual FPGA

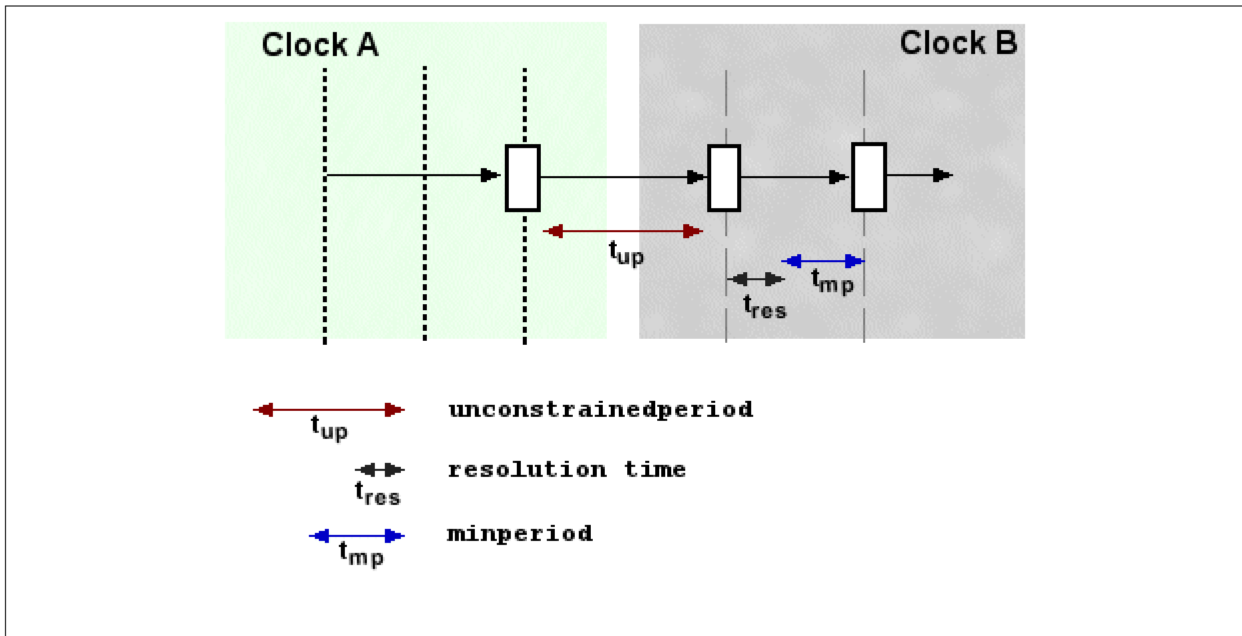
Program your code so it can deal with variable latency. It is unsafe to rely on a particular observed latency, either in hardware or in simulation.

Do not make assumptions about the latency of cross clock-domain channels

The effect of constraining unconstrained paths
Unconstrained paths are created in the synchronization hardware used to connect channels across clock domains.

```
set clock = external "C43" with {rate = 40, unconstrainedperiod = 100,
minperiod=5};
```

`unconstrainedperiod` can be given a value to stop place and route tools giving an unconstrained period warning. If latency is critical in your design, note that as the value of `unconstrainedperiod` increases, latency may increase.



Throughput between clock domains using channels

To ensure a throughput of one word per cycle between clock domains, you need to have a sufficiently large FIFO.

The table below shows the average clock cycles needed to send 1000 words from one clock domain to another of a similar frequency and back again. It was measured in the originating domain using timing accurate simulation of a back-annotated netlist.

fifolength	paranoia		
	0	**1**	**2**
0	5008	7010	9011
1	5008	7010	9011
2	3009	5011	7012
3	1341	2010	5013
4	1808	2513	4013
5	1510	2155	3215
6	1343	1888	2731
7	1008	1010	2389
8	1102	1443	2269
9	1012	1348	1926
10	1012	1265	1771
11	1012	1192	1634
12	1012	1128	1521
13	1012	1070	1422
14	1012	1010	1354
15	1008	1010	1304
16	1012	1018	1239
17	1012	1018	1188
18	1012	1018	1142
19	1012	1018	1060
20	1012	1018	1060

This demonstrates that timing between different clock domains cannot be predicted accurately. FIFOs with a length of a power of 2 are slower. Other differences may be accounted for by layout.

It will not vary widely over different devices and different clock rates.

Synchronization between clock domains

You may use a zero-width channel to convey synchronization information between clock domains. Sending 0 along a 0 width channel will create the synchronization hardware, such that subsequent statements will be synchronized.

For example, if you have two statements in different clock domains and you want one to execute if and only if the other one does then you can do something like:

Domain 1:

```
chan unsigned 0 ch;
ch!0;
<statement 1>
```

Domain 2:

```
extern chan ch;
unsigned 0 junk;
ch?junk;
<statement 2>
```

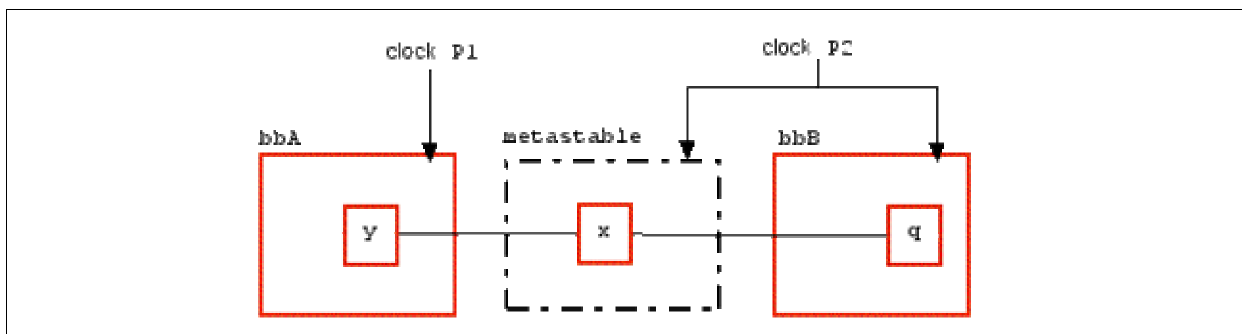
In this example, each domain will wait for the other before statement 1 and statement 2 are executed

Using interfaces to communicate between clock domains

If you are using interfaces rather than channels to communicate between hardware components in separate clock domains, you will need to insert resynchronizing hardware if it is not included in the components. For example, if data is sent from `port_out A` in domain bbA and received from `port_in B` in domain bbB, the data must be resynchronized to the clock in domain bbB.

Using interfaces: External resynchronizing example

This example shows the three files required to connect two EDIF blocks (`bbA` and `bbB`) which use different clocks. The small files `bbA.hcc` and `bbB.hcc` compile to the EDIF code using the `port_out from` and `port_in to` interfaces. The `metastable.hcc` file connects the two together and generates one flip –flop that resynchronizes the data by reading the value from `bbA` into a variable.



File: metastable.hcc

```
/*
* Black box code to resynchronize
* Needs to be clocked from the reading clock
* (i.e. bbB.hcc's clock)
*/

int 1 x;
interface bbA(int 1 from) A();
interface bbB() B(int 1 to=x, unsigned 1 clk = __clock);

set clock = external "P1";
void main(void)
{
    while(1)
    {
        /*
         * stabilize the data by adding resynchronization FF
         */
        x = A.from;
    }
}
```

File: bbA.hcc

```
/*
* Domain bbA
* Compiles to bbA.edf
*/
interface port_in(unsigned 1 clk with { clockport = 1 }) clk();
set clock = internal clk.clk;
void main(void)
{
    int 1 y;
    interface port_out() from (int 1 from = y);
}
```

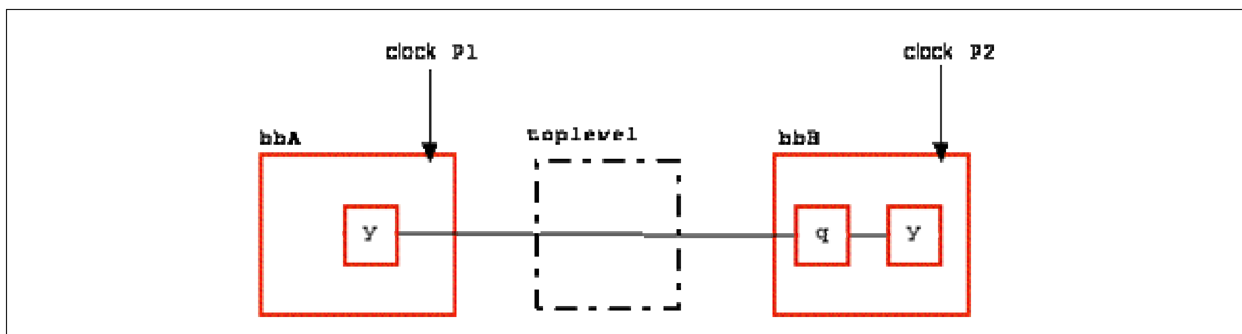
File: bbB.hcc

```
/*
*Domain bbB
* Compiles to bbB.edf
*/

set clock = external "P2";
void main(void)
{
    int 1 q;

    interface port_in(int 1 to) to();
    par
    {
        while(1)
        {
            q = to.to; // Read data
        }
    }
}
```

Internal resynchronizing example

The resynchronizing flip-flop can be placed in the file that reads the data from the foreign code block.

This example shows the three files required to connect two EDIF blocks (bbA and bbB) which use different clocks. The small files bbA.hcc and bbB.hcc compile to the EDIF code using the port_out from and port_in to interfaces. The toplevel.hcc file connects them together. The data is resynchronized in the bbB.hcc file.



File: toplevel.hcc

```
/*
* Code to connect data between two cores
*/

interface bbA(int 1 from) A();
interface bbB() B(int 1 to=A.from);
```

File: bbA.hcc

```
/*
* Domain bbA
* Compiles to bbA.edf
*/
set clock = external "P1";
void main(void)
{
    int 1 y;
    interface port_out() from (int 1 from = y);
}
```

File: bbB.hcc

```
/*
*Domain bbB
* Complies to bbB.edf
*/

set clock = external "P2";
void main(void)
{
    int 1 q, y;

    interface port_in(int 1 to) to();
    while(1)
    {
        par
        {
            q = to.to;  // Resynchronize data
            y = q;
        }
    }
}
```

9.3.2 Simulating multiple clock domains

You may simulate your design by

- using the DK simulator
- untimed simulation of generated VHDL code
- simulation of back-annotated netlist

Note that as the timing accuracy of the simulation increases, it is harder to relate errors to the original Handel-C code.

Using the DK simulator

The DK simulator may not simulate the timing of channels between clock domains identically to that in the generated hardware. You must not rely on observed latency or timing behaviour in either simulation or hardware.

When you simulate designs with multiple clocks, you will get a Select Clock dialog in the GUI asking you which clock you want to follow. If you want to synchronize the clocks in a simulation, use the `DKSync.dll` plugin.

10 Targeting hardware and simulation

10.1 Interfacing with the simulator

Communication with the simulator takes place over channels. They are declared using the keywords `chanin` and `chanout`. Standard channel communication statements can then be used to transfer data. It is assumed that channels to and from the simulator never block and will always complete a transfer in one clock cycle.

Channels to and from the simulator are declared using `chanin` and `chanout` instead of `chan`.

The special channels `chanin` and `chanout` are normally connected to files. Only integer values can be used as input data, and files connected to `chanin` must be correctly formatted. An unconnected channel that outputs data to the simulator will be displayed in the debug window. You can declare multiple channels for input and output and connect more than one channel to the same file, but you cannot read from the same channel more than once in a clock cycle. If the simulation is still running when the end of the file has been reached, the simulator will read in zeroes.

You cannot use `chanin` or `chanout` in a `struct`. Use pointers to `chanin` or `chanout` instead.

Simple example

```
chanin unsigned Input with {infile = "../Data/source.dat"};
chanout unsigned Output;

input ? x;
output ! y;
```

This example declares two channels: one for input from the simulator and one for output to the simulator. The input channel connects to a file managed by the simulator; the output channel connects to the simulator's standard output (the debug window in the DK GUI).

Multiple channel example

```
chanin int 8 input_1 with
        {infile = "../Data/source_1.dat"};
chanin int 16 input_2 with
        {infile = "../Data/source_2.dat"};
chanout unsigned 3 output_1;
chanout char output_2;

int 8 a;
int 16 b;

input_1 ? a;
input_2 ? b;
output_1 ! (unsigned 3)(((0 @ a) + b) <- 3);
output_2 ! a;
```

When simulated, such a program displays the name of channels before outputting their value on the simulating computer screen.

10.1.1 Simulator input file format

The data input file should have one number per line separated by newline characters (either DOS or UNIX format text files may be used). Each number may be in any format normally used for constants by Handel-C. You can only use integer values. Blank lines are ignored as are lines prefixed by // (comments). For example:

```
56
0x34
0654
0b001001

//is a comment, blank lines ignored
27
```

If EOF file is reached while reading an input file, zeroes will be read in until the simulation completes.

10.1.2 Block data transfers

The Handel-C simulator has the ability to read data from a file and write results to another file. For example:

```
chanin int 16 input with {infile = "in.dat"};
chanout int 16 output with {outfile = "out.dat"};

void main (void)
{
    while (1)
    {
        int value;

        input ? value;
        output ! value+1;
    }
}
```

This program reads data from the file `in.dat` and writes its results to the file `out.dat`. The simulator will open and close the specified files for reading or writing as appropriate. If EOF file is reached while reading an `infile` file, zeroes will be read in until the simulation completes.

If the `in.dat` file consists of:

```
56
0x34
0654
0b001001
```

the `out.dat` will contain the decimal results as follows:

```
57
53
429
10
```

The `base` specification can be used to write to the outfile in different formats.

Block data transfers allow algorithms to be debugged and tested without needing to build actual hardware. For example, an image processing application may store a source image in a file and place its results in a second file. All that need be done outside the Handel-C compiler is a conversion from the image (e.g. JPEG file) into the text file (which can then be used by the simulator) and a conversion back from the output file to the image format. The results can then be viewed and the correct operation of the Handel-C program confirmed.

10.2 Targeting FPGA and PLD devices

The Handel-C language is designed to target real hardware devices. To do this, you must supply this information to the compiler:

- the FPGA/PLD family and part that the design will be implemented in
  These are supplied on the **Chip** tab of the **Project>Settings** dialog. They can also be specified in the source code using the `set family` and `set part` statements or they can be supplied to the command line using the -f *`family`* and `-p` *`part`* switches. They will be passed to the FPGA/PLD place and route tool to inform it of the device it should target.
- in some cases, the location of a reset source (required for Actel devices)
  The reset source is specified using the set reset command.

Your license may restrict the devices you can target. The devices available to you are listed in the **Family** box on the **Chip** tab in Project Settings.

10.2.1 Summary of supported devices

In order to target a specific FPGA or PLD, the compiler must be supplied with the part number. Ultimately, this information is passed to the place and route tool to inform it of the device it should target.

You can specify your target device using the **Chip** tab on the **Project Settings** dialog, or within your source code.Your license may restrict the devices you can target. The devices available to you are visible in the **Family** list on the **Chip** tab.

Recognized families are:

Description	On-chip asynchronous RAMs	On-chip synchronous RAMs
Actel ProASIC series FPGAs	Block RAM, dual-port	Block RAM, dual-port

Actel ProASIC+ series FPGAs	Block RAM, dual-port	Block RAM, dual-port
Altera Apex 20K series PLDs	Block RAM (in ESBs), dual-port	Block RAM (in ESBs), dual-port
Altera Apex 20KE series PLDs	Block RAM (in ESBs), dual port	Block RAM (in ESBs), dual port
Altera Apex 20KC series PLDs	Block RAM (in ESBs), dual port	Block RAM (in ESBs), dual port
Altera ApexII series PLDs	Block RAM (in ESBs), dual-port	Block RAM (in ESBs), dual-port
Altera Cyclone PLDs	-	M4K dual port RAM
Altera Cyclone II PLDs	-	M4K dual port RAM
Altera Excalibur ARM series PLDs	Block RAM (in ESBs), dual-port	Block RAM (in ESBs), dual-port
Altera Flex10K series PLDs	Block RAM (in EABs), dual-port	Block RAM (in EABs), dual-port
Altera Flex10KA series PLDs	Block RAM (in EABs), dual-port	Block RAM (in EABs), dual-port
Altera Flex10KB series PLDs	Block RAM (in EABs), dual-port	Block RAM (in EABs), dual-port
Altera Flex10KE series PLDs	Block RAM (in EABs), dual-port	Block RAM (in EABs), dual-port
Altera Mercury series ASSPs	Block RAM (in ESBs), dual-port, quad-port	Block RAM (in ESBs), dual-port, quad-port
Altera Stratix PLDs	-	3 types of dual-port RAM in TriMatrix blocks
Altera Stratix GX PLDs	-	3 types of dual-port RAM in TriMatrix blocks
Altera Stratix II PLDs	-	3 types of dual-port RAM in TriMatrix blocks
Xilinx Spartan series FPGAs	SelectRAM, dual-port	-
Xilinx Spartan-XL series FPGAs	SelectRAM, dual-port	-
Xilinx Spartan-II series FPGAs	SelectRAM, dual-port	Block RAM
Xilinx Spartan-IIE series FPGAs	SelectRAM, dual-port	Block RAM, dual-port

Xilinx Spartan-3 series FPGAs	SelectRAM, dual-port	Block RAM, dual-port
Xilinx Spartan-3E series FPGAs	SelectRAM, dual-port	Block RAM, dual-port
Xilinx Spartan-3L series FPGAs	SelectRAM, dual-port	Block RAM, dual-port
Xilinx Virtex series FPGAs	SelectRAM, dual-port	Block RAM, dual-port
Xilinx VirtexE series FPGAs	SelectRAM, dual-port	Block RAM, dual-port
Xilinx Virtex-II series FPGAs	SelectRAM, dual-port	Block RAM, dual-port
Xilinx Virtex-II Pro series FPGAs	SelectRAM, dual-port	Block RAM, dual-port
Xilinx Virtex-II Pro X series FPGAs	SelectRAM, dual-port	Block RAM, dual-port
Xilinx Virtex-4 series FPGAs	SelectRAM, dual-port	Block RAM, dual-port
"Generic" (VHDL or Verilog projects only. Results in HDL without target-specific constructs.)	-	-

10.2.2 Detecting the current device family

The `__isfamily` construct allows you to detect what the current device family is. If you are writing platform-independent libraries, you can use this to conditionally select pieces of the source code to exploit the resources available on different FPGAs.

The construct takes a device string and returns true or false. The possible device names are the same as those used to specify devices with the `set family` construct. An error is returned if the string specified inside the construct is not a recognized family string.

Example

```
set family = XilinxVirtex;

macro expr DoThis() =
    select (__isfamily(XilinxVirtex) : DoThing1() :
        select (__isfamily(AlteraApex20K) : DoThing2() :
            select (__isfamily(MadeUpDevice) : DoThing3() : DoThing4())
        )
        );
```

The first use of `__isfamily()` would return true, the second would return false, and the third would result in a compiler error. The source code specified in the `DoThing1()` function would be selected.

10.2.3 Targeting specific devices via source code

If you are not using the GUI or the command line to specify the target device, you must insert lines in the code to specify it. In order to target a specific FPGA or PLD, the compiler must be supplied with the FPGA part number. Ultimately, this information is passed to the FPGA/PLD place and route tool to inform it of the device it should target.

Targeting devices is in two parts: specifying the target family and the target device. The general syntax is:

```
set family = Family;
set part = Chip Number;
```

Recognized families are:

Family name	Description
`Actel500K`	Actel ProASIC series FPGAs
`ActelPA`	Actel ProASIC+ series FPGAs
`AlteraFlex10K`	Flex10K series Altera PLDs
`AlteraFlex10KA`	Flex10KA series Altera PLDs
`AlteraFlex10KB`	Flex10KB series Altera PLDs
`AlteraFlex10KE`	Flex10KE series Altera PLDs
`AlteraApex20K`	Apex 20K series Altera PLDs
`AlteraApex20KE`	Apex 20KE series Altera PLDs
`AlteraApex20KC`	Apex 20KC series Altera PLDs
`AlteraApexII`	Apex II series PLDs
`AlteraMercury`	Altera Mercury series PLDs
`AlteraStratix`	Altera Stratix PLDs
`AlteraStratixII`	Altera Stratix II PLDs
`AlteraStratixGX`	Altera Stratix GX PLDs
`AlteraCyclone`	Altera Cyclone PLDs
`AlteraCycloneII`	Altera Cyclone II PLDs
`AlteraExcaliburARM`	Altera Excalibur ARM series PLDs

`XilinxVirtex`	Virtex Xilinx FPGAs
`XilinxVirtexE`	VirtexE Xilinx FPGAs
`XilinxVirtexII`	Virtex-II Xilinx FPGAs
`XilinxVirtexIIPro`	Virtex-II Pro Xilinx FPGAs
`XilinxVirtexIIProX`	Virtex-II Pro X Xilinx FPGAs
`XilinxVirtex4`	Virtex-4 Xilinx FPGAs
`XilinxSpartan`	Spartan Xilinx FPGAs
`XilinxSpartanXL`	Spartan-XL Xilinx FPGAs
`XilinxSpartanII`	Spartan-II Xilinx FPGAs
`XilinxSpartanIIE`	Spartan-IIE Xilinx FPGAs
`XilinxSpartan3`	Spartan-3 Xilinx FPGAs
`XilinxSpartan3E`	Spartan-3E Xilinx FPGAs
`XilinxSpartan3L`	Spartan-3L Xilinx FPGAs

Your license may restrict the devices you can target. The devices available to you are visible in the **Family** list on the **Chip** tab in Project Settings.

The part string is the complete Actel, Altera or Xilinx device string. For example:

```
set family = XilinxVirtex;
set part = "V1000BG560-4";
```

This instructs the compiler to target a v1000 device in a BG560 package. It also specifies that the device is a -4 speed grade. This last piece of information is required for the timing analysis of your design by the Xilinx tools.

The family is used to inform the compiler of which special blocks it may generate.

To target Altera Flex 10K devices:

```
set family = AlteraFlex10K;
set part = "EPF10K20RC240-3";
```

This instructs the compiler to target an Altera Flex 10K20 device in a RC240 package. It also specifies that the device is a -3 speed grade. This last piece of information is required for the timing analysis of your design by the Altera Max Plus II or Quartus tools. Note that when performing place and route on the resulting design, the device and package must also be selected via the menus in the Max Plus II or Quartus software.

To target Actel ProASIC devices:

```
set family = Actel500K;
set part = "A500K270-BG456I";
```

This instructs the compiler to target an Actel ProASIC device with 270,000 gates in a BG456 package. It also specifies that the device is a standard speed grade, and that the device is to be used for an industrial application: the "I" at the end of the part string specifies that the device is to conform to industrial temperature range standards. The speed information is required for the timing analysis of your design by the Actel Designer tools. The application information ("industrial" in this example) is required for place and route of your design by the Actel Designer tools. Note that when performing place and route on the resulting design, the device and package must also be selected via the menus in the Designer software.

10.2.4 Specifying a global reset

`set reset` allows you to reset your device into a known state at any time. It is particularly useful for setting up devices which are not in a known state at start up.

`set reset` causes the program to return to its initial state and resets global and static variables to their initial values. However, it does not reset any RAMs (distributed or block). By default, the reset is asynchronous and thus occurs immediately (it does not wait for the next clock tick.) To make the global reset synchronous, use the synchronous specification.

Examples

```
signal unsigned 1 x;
set reset = internal !x;    // resets when x is zero

set reset = external "P1"; // resets when signal sent to named pin

set reset = external;       // connects to pin, but doesn't specify which
```

10.3 Use of RAMs and ROMs with Handel-C

Handel-C provides support for:

- interfacing to on-chip and off-chip RAMs and ROMs using the `ram` and `rom` keywords.
- specifying RAMs and ROMs external to the Handel-C code by using the `ports` specification keyword.
- controlling the timing for read/write cycles by using specification keywords that define the relationship between the RAM strobe and the Handel-C clock.

The usual technique for specifying timing in synchronous and asynchronous RAM is to have a fast external clock which is divided down to provide the Handel-C clock and used directly to provide the pulses to the RAM.

10.3.1 Asynchronous RAMs

There are three techniques for timing asynchronous RAMs, depending on the clock available

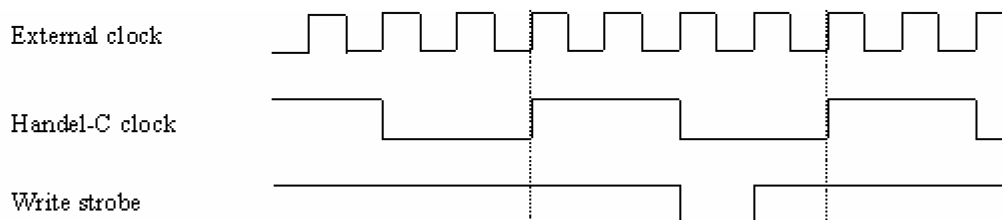
- Fast external clock. Use the Handel-C `westart` and `welength` specifications to position the write strobe.
- External clock at the same speed as the Handel-C clock. Use multiple reads to give the RAM enough time to respond.
- Use the `wegate` specification to position the write enable signal within the Handel-C clock.

Fast external clock

This method of timing asynchronous RAMs depends on having an external clock that is faster than the internal clock (i.e. the location of the clock is `internal_divide` or `external_divide` with a division factor greater than 1). If so, Handel-C can generate a write strobe for the RAM which is positioned within the Handel-C clock cycle. This is done with the `westart` and `welength` specifications. For example:

```
set clock = external_divide "P78" 4;
ram unsigned 6 x[34] with { westart = 2,
                            welength = 1 };
```

The write strobe can be positioned relative to the Handel-C clock cycle by half cycle lengths of the external (undivided) clock. The above example starts the pulse 2 whole external clock cycles into the Handel-C clock cycle and gives it a duration of 1 external clock cycle. Since the external clock is divided by a factor of 4, this is equivalent to a strobe that starts half way through the internal clock cycle and has a duration of one quarter of the internal clock cycle. This signal is shown below:



TIMING DIAGRAM: POSITIONED WRITE STROBE

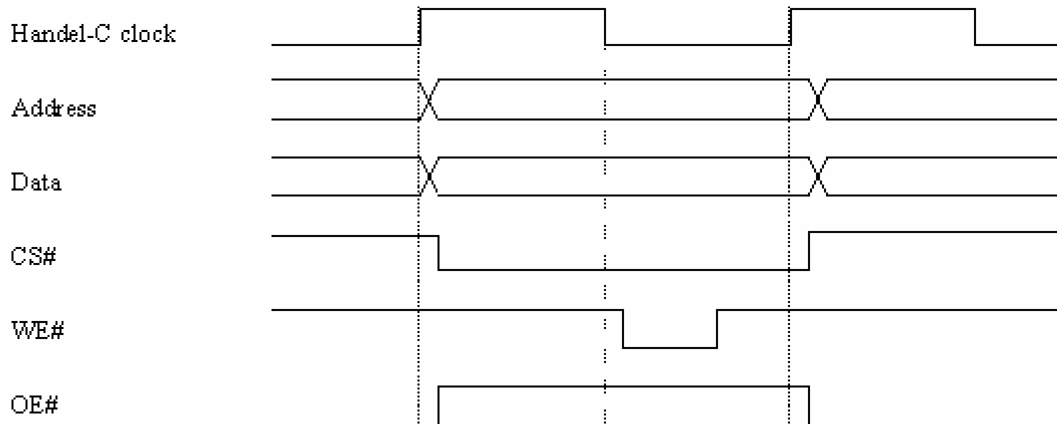
This timing allows half a clock cycle for the RAM set-up time on the address and data lines and one quarter of a clock cycle for the RAM hold times. This is the recommended way to access asynchronous RAMs.

Fast external clock example

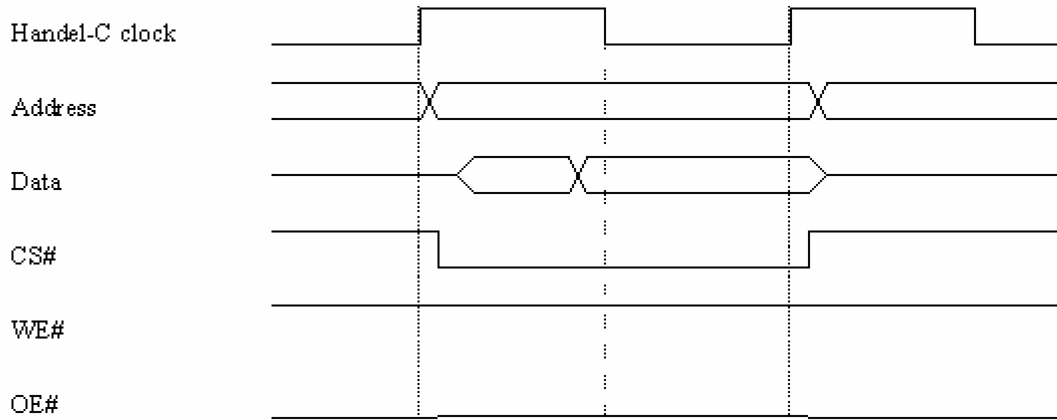
To declare a 16Kbyte by 8-bit RAM:

```
set clock = external_divide "P99" 4;

ram unsigned 8 ExtRAM[16384] with {
       offchip = 1,
       westart = 2,
       welength = 1,
       data = {"P1", "P2", "P3", "P4",
               "P5", "P6", "P7", "P8"},
       addr = {"P9", "P10", "P11", "P12",
               "P13", "P14", "P15", "P16",
               "P17", "P18", "P19", "P20",
               "P21", "P22"},
       we = {"P23"},
       oe = {"P24"},
       cs = {"P25"}};
```

The compiled hardware generates the following cycle for a write to external RAM:



The compiled hardware generates the following cycle for a read from external RAM:



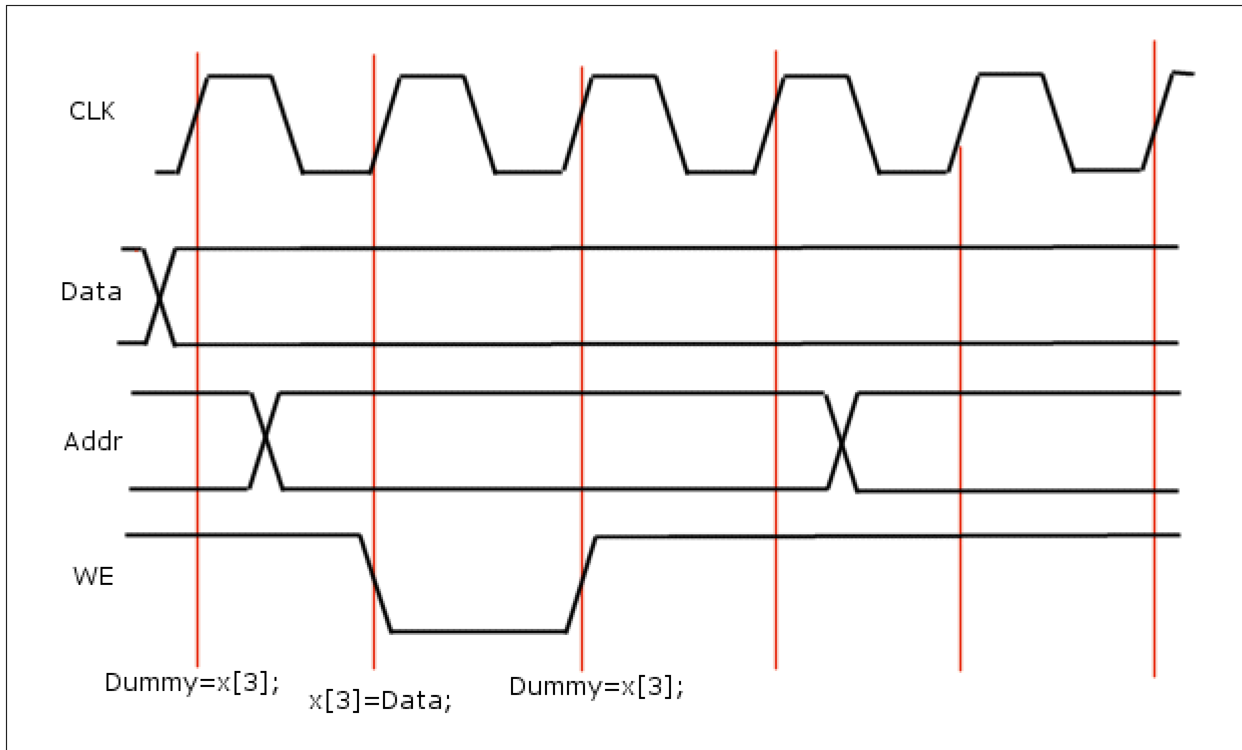
Same rate external clock

This method of timing asynchronous RAMs uses multiple Handel-C RAM accesses to meet the setup and hold times of the RAM.

```
ram unsigned 6 x[34];

Dummy = x[3];
x[3] = Data;
Dummy = x[3];
```

This code holds the address constant around the RAM write cycle, enabling a write to an asynchronous RAM.

The timing diagram below shows the address being held constant during the write strobe. It is held constant by the two assignments to `Dummy`.



Undivided external clock

This method of accessing asynchronous RAMs is a compromise between the other two methods (fast external clock and multiple RAM accesses). `wegate` is used with an undivided external clock and keeps the write strobe in the first or second half of the clock cycle. It is still necessary to hold the address constant either in the clock cycle before or in the clock cycle after the access. For example:

```
ram unsigned 6 x[34] with { wegate = 1 };

x[3] = Data;
Dummy = x[3];
```

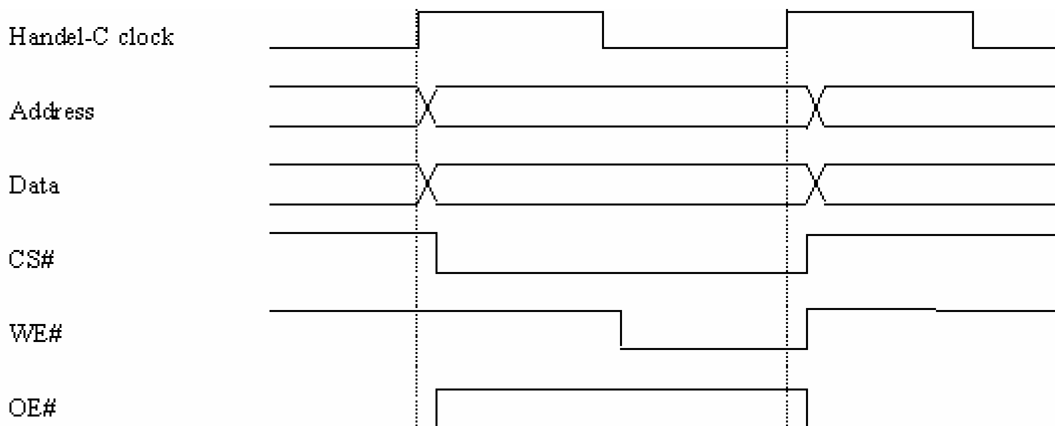
This places the write strobe in the second half of the clock cycle (use a value of -1 to put it in the first half) and holds the address for the clock cycle after the write. The RAM therefore has half a clock cycle of set-up time and one clock cycle of hold time on its address lines.

wegate example

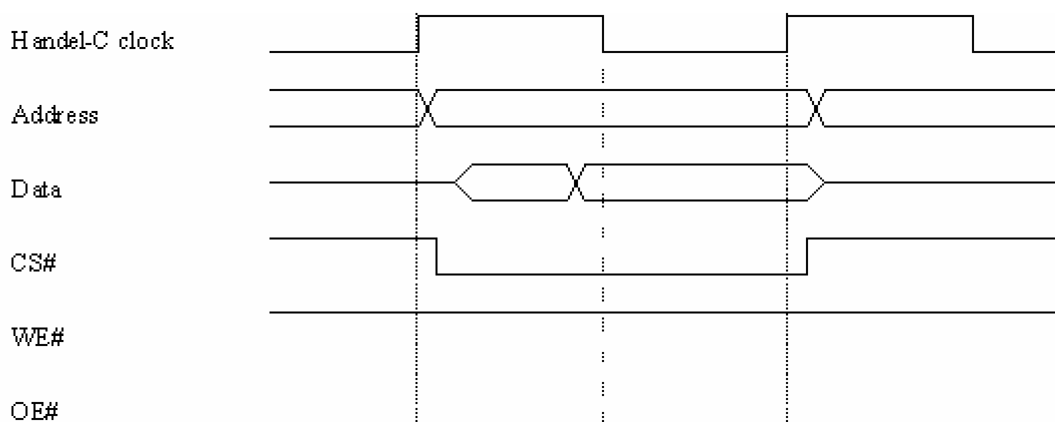
The `wegate` specification may be used when a divided clock is not available. For example, to declare a 16Kbyte by 8-bit RAM:

```
ram unsigned 8 ExtRAM[16384] with {
       offchip = 1,
       wegate = 1,
       data = {"P1", "P2", "P3", "P4",
               "P5", "P6", "P7", "P8"},
       addr = {"P9", "P10", "P11", "P12",
               "P13", "P14", "P15", "P16",
               "P17", "P18", "P19", "P20",
               "P21", "P22"},
       we = {"P23"},
       oe = {"P24"},
       cs = {"P25"}};
```

The compiled hardware generates the following cycle for a write to external RAM:



The compiled hardware generates the following cycle for a read from external RAM:



Note that the timing diagram above may violate the hold time for some asynchronous RAM devices. If the delay between rising clock edge and rising write enable is longer than the delay between rising clock edge and the change in data or address then corruption in

the write may occur in these devices. The two cycle access does not solve the problem since it is not possible to hold the data lines constant beyond the end of the clock cycle. If this causes a problem then a multiplied external clock must be used as described above.

Using the `wegate` specification may violate the hold time for some asynchronous RAM devices.

Targeting external asynchronous RAMs

Handel-C provides support for accessing off-chip static RAMs in the same way as you access internal RAMs. The syntax for an external RAM declaration is:

```
ram Type Name[Size] with {
        offchip = 1,
        data = Pins,
        addr = Pins,
        we = Pins,
        oe = Pins,
        cs = Pins};
```

To declare a 16Kbyte by 8-bit RAM:

```
ram unsigned 8 ExtRAM[16384] with {
       offchip = 1,
       data = {"P1", "P2", "P3", "P4",
               "P5", "P6", "P7", "P8"},
       addr = {"P9", "P10", "P11", "P12",
               "P13", "P14", "P15", "P16",
               "P17", "P18", "P19", "P20",
               "P21", "P22"},
       we = {"P23"},
       oe = {"P24"},
       cs = {"P25"}};
```

Note that the lists of address and data pins are in the order of most significant to least significant. It is possible for the compiler to infer the width of the RAM (8 bits in this example) and the number of address lines used (14 in this example) from the RAM's usage. This is not recommended since this declaration deals with real external hardware which has a fixed definition.

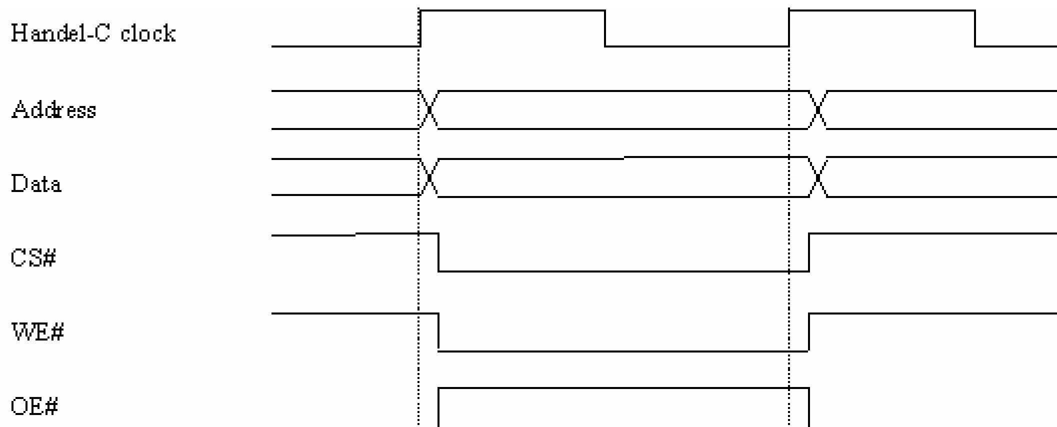
Accessing RAM

Accessing the RAM is the same as for accessing internal RAM. For example:

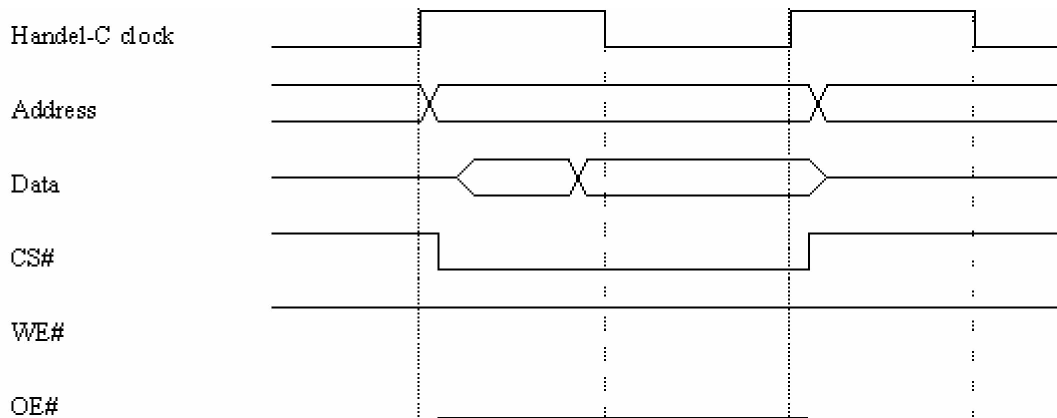
```
ExtRAM[1234] = 23;
y = ExtRAM[5678];
```

Similar restrictions apply as with internal RAM - only one access may be made to the RAM in any one clock cycle.

The compiled hardware generates the following cycle for a write to external RAM:



The compiled hardware generates the following cycle for a read from external RAM:



This cycle may not be suitable for the RAM device in use. In particular, asynchronous static RAM may not work with the above cycle due to set-up and hold timing violations. For this reason, the `westart`, `welength` and `wegate` specifications may also be used with external RAM declarations.

10.3.2 Synchronous RAMs

SSRAM clocks

Handel-C timing semantics require that any assignment takes one clock cycle. Typically, SSRAMs have a latency of at least one clock cycle. Therefore, in order for accesses to a

SSRAM device to conform to Handel-C's one-clock-cycle-per-assignment rule, the SSRAM clock needs to be offset from the Handel-C clock. If the SSRAM has a latency of more than one clock cycle, its clock needs to be faster than the Handel-C clock, as well as being offset from it.

This is done by using an independent fast clock (RAMCLK) to match the SSRAM timings with the Handel-C timing constraints.

A fast external clock (CLK) is divided to provide the Handel-C clock (HCLK), and is also used to generate pulses to clock the SSRAM, where the pulses can be placed within a single HCLK cycle. This placed clock is the RAMCLK. It can be carried to an external SSRAM using the `clk` specification.

By default, the Handel-C compiler uses an inverted copy of the Handel-C clock to drive synchronous on-chip memories. This may mean you need to run your design at a lower clock frequency than you want to. You can increase the efficiency of your design by:

- using pipelined memory accesses, for certain on-chip SSRAMs. This is illustrated by the ***Pipelined on-chip SSRAM examples*** (see page 207), and ***the Pipelined on-chip SSRAM timing diag***rams (see page 205).
- using the clock position specifications to alter the position of the RAM clock relative to the Handel-C clock, to enable full memory accesses to be performed within 1 Handel-C clock cycle. For example, you might want to advance the write-clock, or delay the read-clock.

  This is most suitable for off-chip RAMs, and is illustrated by the ***Flow-through SSRAM example*** (see page 212) and the ***Pipelining off-chip SSRAM example*** (see page 214).

SSRAM devices supported

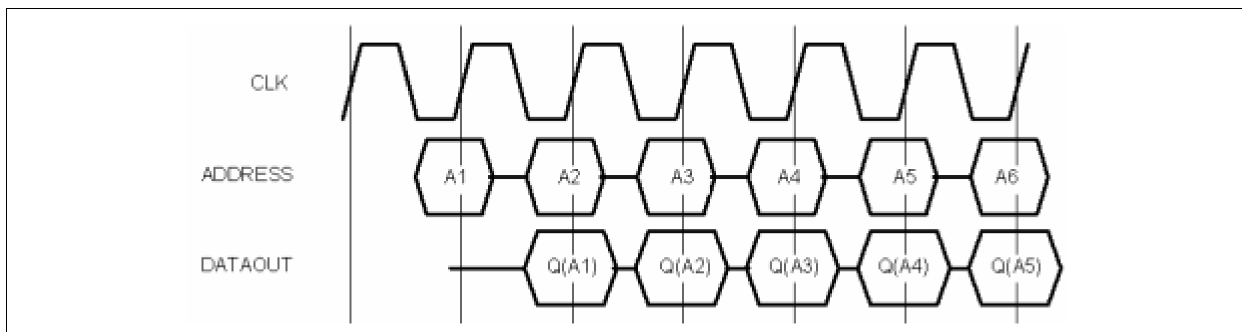
Handel-C supports ZBT-compatible (Zero Bus Turnaround) flow-through and pipelined output devices. DDR (double data rate) and QDR (quad data rate) devices are not supported directly; you can write your own interfaces.

SSRAM write-enable

The Handel-C compiler checks the `block` and `offchip` specifications to find out what type of RAM is being built and generates the appropriate write-enable signal (e.g. active low for ZBT SSRAM devices and active-high for block RAMs within Xilinx Virtex chips).

SSRAM read and write cycles

The inputs to most inputs to SSRAMs are captured on the rising edge of the input clock. During a read cycle there is a latency of at least one clock cycle between an address being captured at the input and data becoming available at the output. This is also true for the write cycle in many devices: an address is captured on one clock cycle, and data on the next. A diagram of a typical timing for a read (or write) cycle for an SSRAM device is shown below.



TIMING DIAGRAM: SSRAM READ AND WRITE

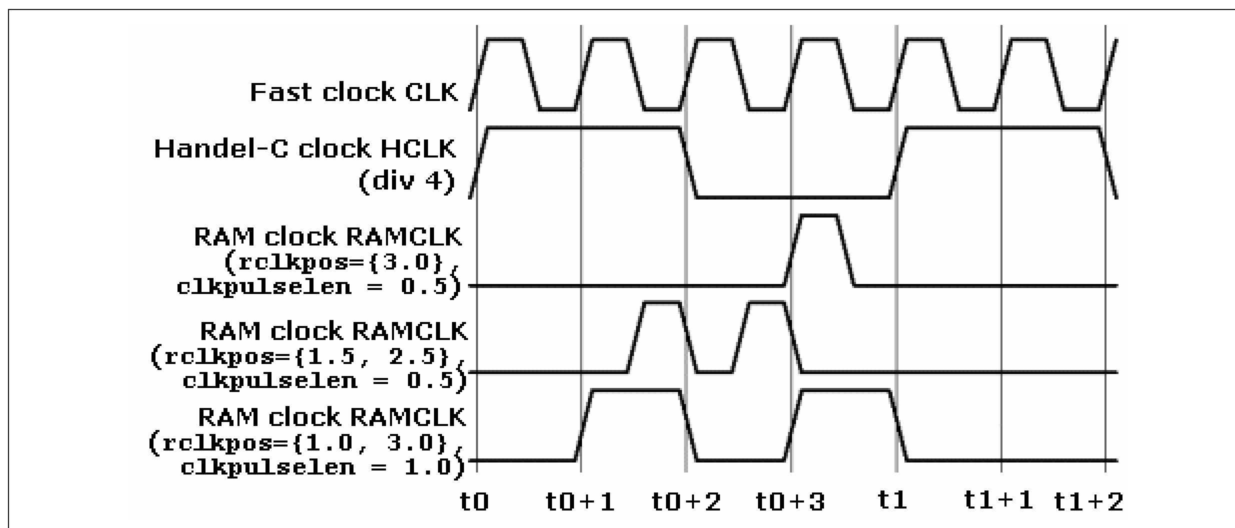
Specifying SSRAM timing

You can place the RAM clock pulses at different points within the Handel-C clock if the Handel-C clock is divided using `external_divide` or `internal_divide`.

If you have a fast undivided clock CLK, a divided clock HCLK, and you want to generate a RAM clock RAMCLK, the following apply:

- The SSRAM clock (RAMCLK) is generated from the fast clock (CLK) according to the specifications: `rclkpos`, `wclkpos` and `clkpulselen`. These specifications can be in whole or half cycles of the external clock (i.e. the specifications are in multiples of 0.5).
- `rclkpos` specifies the positions of the clock cycles of clock RAMCLK for a read cycle. These positions are specified in terms of cycles and half-cycles of CLK, counting forwards from a HCLK rising edge.
- `wclkpos` specifies the positions of the clock cycles of RAMCLK for a write cycle. These are also counted forward from an HCLK rising edge.

- `clkpulselen` specifies the length of the RAMCLK pulses in CLK cycles. This is specified once per RAM. It applies to both the read and write clocks.



TIMING DIAGRAM: SSRAM READ CYCLE USING GENERATED RAMCLK

The pulse positions and lengths are specified in cycles and half-cycles of CLK.

The `westart` and `welength` specifications are used to place the write enable strobe where it is required.

Pipelining on-chip SSRAM

By default, the DK compiler uses an inverted version of the main Handel-C clock to drive on-chip synchronous memories. This allows it to conform to Handel-C's timing semantics of 1 clock cycle per assignment. But it can potentially halve the maximum clock rate for a design.

Handel-C can pipeline accesses to on-chip SSRAMs if you write your code in a certain way. The effect is that the memory is driven by the main (non-inverted) Handel-C clock, potentially doubling the clock rate for the design, and accesses are performed with 1 clock cycle of latency.

Creating pipelined SSRAM accesses

For memory accesses to be pipelined, the following rules must be satisfied:

- The memory must always be read into an uninitialized register, and nowhere else.
- Nothing else must write to this register.

For multi-port memories, both rules must be satisfied for every readable port.

If these rules are satisfied, the compiler removes the output register and drives the memory with the main (non-inverted) Handel-C clock.

You can disable the transform by using the `-N-piperam` command line switch, or by de-selecting the **Enable memory pipelining transformations** box on the **Synthesis** tab in Project Settings.

The transform is effective for all forms of hardware output. Simulation is not affected.

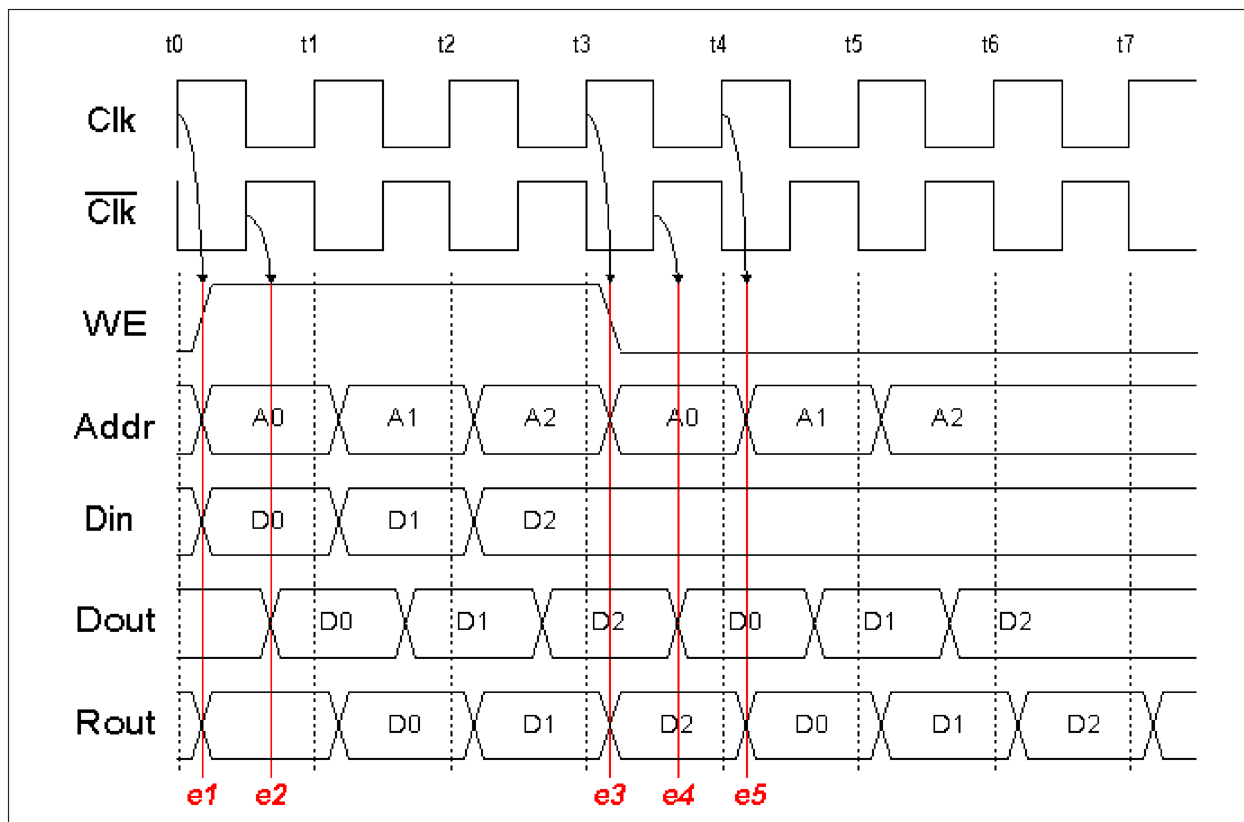
Devices supporting pipelined on-chip SSRAM

The transform only applies to certain devices and configurations:

Family	BlockRAM	EAB	M512	M4K	M-RAM
Actel ProAsic, Actel ProASIC+	yes	n/a	n/a	n/a	n/a
Altera Apex20K, Altera Apex20KE, Altera Apex20KC, Altera Excalibur ARM, Altera Mercury	n/a	yes	n/a	n/a	n/a
Altera ApexII	n/a	yes, except for single-port RAMs and true dual-port RAMs	n/a	n/a	n/a
Altera Stratix, Altera StratixGX, Altera Stratix II	n/a	n/a	yes, except for true dual-port RAMs;	yes, except for single-port RAMs and true dual-port RAMs	no
Altera Cyclone, Altera Cyclone II	n/a	n/a	no	except for single-port RAMs and true dual-port RAMs	no
Xilinx Virtex, Xilinx Spartan-II	no	n/a	n/a	n/a	n/a
Xilinx Virtex-II, Xilinx Virtex-II Pro, Xilinx Spartan-3 Xilinx Spartan-3E Xilinx Spartan-3L	yes	n/a	n/a	n/a	n/a

Pipelined on-chip SSRAM timing diagrams
The timing diagrams below illustrate the difference between pipelined accesses to SSRAM and non-pipelined accesses. The non-pipelined RAM can be transformed into a pipelined RAM if the memory is read into an uninitialized register reserved specifically for the use of the memory.

Non-pipelined access to RAM



3 write cycles are performed:

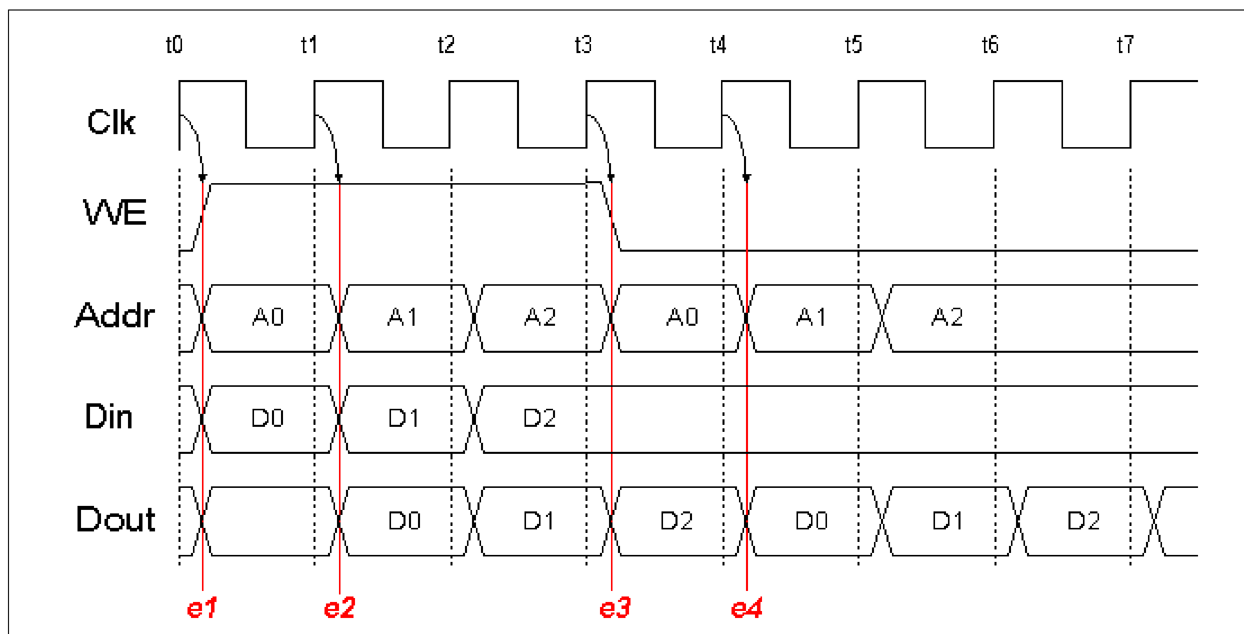
- At time t0, the rising edge of the main Handel-C clock `CLK` initiates a write cycle.
- At event e1, `WE` is asserted and `Addr` and `Din`, are set up, so that when the memory is next clocked, the data at `Din` will be written at the location specified in `Addr`.
- At time t0.5, the inverted clock rising edge clocks the memory, causing it to execute the write operation.
- At event e2, after the write operation has completed, the data that has been written becomes available at the output from the memory `Dout`.

Two further write cycles are performed, starting at time t1 and t2. This is followed by a read cycle:

- At time t3, the main Handel-C rising clock edge initiates a read cycle.
- At event e3, WE is de-asserted and Addr is set up, so that when the memory is next clocked, the location specified at Addr will be read.
- At time t3.5, the inverted clock rising edge clocks the memory, causing it to execute the read operation.
- At event e4, after the read operation has completed, the data that has been read becomes available at the output from the memory Dout.
- At time t4, the main Handel-C rising clock edge clocks the data that has been read from the memory into the pipeline register, as well as initiating the next read cycle.
- At event e5, after the write-to-register operation has completed, the data that has been written becomes available at the register output Rout.
- At time t5, the data that was read via the pipeline register (D0 in this case) is ready to be clocked into its destination.

Two further read cycles are performed, starting at time t4 and t5.

Pipelined access to RAM



3 write cycles are performed:

- At time t0, the main Handel-C rising clock edge initiates a write-cycle.
- At event e1, WE is asserted and Addr and Din are set up, meaning that when the memory is next clocked, the data at Din will be written at the location specified in Addr.
- At time t1, the main Handel-C rising clock edge clocks the memory, as well as initiating the next write cycle.

- At event e2, after the write operation has completed, the data that has been written becomes available at the output from the memory `Dout`.

Two further write cycles are performed, starting at time t1 and t2. This is followed by a read cycle:

- At time t3, the main Handel-C rising clock edge initiates a read cycle.
- At event e3, `WE` is de-asserted and `Addr` is set up, meaning that when the memory is next clocked, the location specified at `Addr` will be read.
- At time t4, the main Handel-C clock rising edge clocks the memory, as well as initiating the next read cycle.
- At event e4, after the read operation has completed, the data that has been read becomes available at the output from the memory `Dout`.
- At time t5, the data that was read (`D0` in this case) is ready to be clocked into its destination.

Two further read cycles are performed, starting at time t4 and t5.

Effect of performing a pipelining transform

The output from the pipeline register (`Rout`) in the non-pipelined version and the output from the memory (`Dout`) in the pipelined version are equivalent, showing that the transformation does not change the overall behaviour of the circuit. Valid data is available from the memory output one whole clock cycle later in the pipelined version, which is why the transform is only valid when there's a 'pipeline' register.

Pipelined SSRAM examples

The following examples demonstrate how you can ***pipeline accesses to on-chip SSRAM*** (see page 203). If the correct conditions are met, the RAM will use the main Handel-C clock instead of an inverted clock, and the output register will be removed.

SPRAM Example 1: transform is performed

```
void main(void)
{
    ram unsigned 4 rax[4] with {block = "BlockRAM"};
    static unsigned 2 i;
    unsigned 4 x;          // x is un-initialized
    interface bus_in(unsigned 4 i) I();
    interface bus_out() O(unsigned 4 o = x);

    while(1)
    {
        rax[i] = I.i;
        i++;
        x = rax[i];        // RAM only read into x
    }
}
```

SPRAM Example 2: transform is not performed (register is initialized)

```
void main(void)
{
    ram unsigned 4 rax[4] with {block = "BlockRAM"};
    static unsigned 2 i;
    static unsigned 4 x;   // x is initialized to zero
    interface bus_in(unsigned 4 i) I();
    interface bus_out() O(unsigned 4 o = x);
    while(1)
    {
        rax[i] = I.i;
        i++;
        x = rax[i];        // RAM only read into x
    }
}
```

SPRAM Example 3: transform is not performed (memory read into two destinations)

```
void main(void)
{
    ram unsigned 4 rax[4] with {block = "BlockRAM"};
    static unsigned 2 i;
    unsigned 4 x;          // x is un-initialized
    unsigned 4 y;          // y is un-initialized
    interface bus_in(unsigned 4 i) I();
    interface bus_out() O(unsigned 4 o = x);
    while(1)
    {
        rax[i] = I.i;
        i++;
        x = rax[i];        // RAM read into x...
        y = rax[i];        // ...but also into y
    }
}
```

SPRAM Example 4: transform is not performed (pipeline register written to from elsewhere)

```
void main(void)
{
    ram unsigned 4 rax[4] with {block = "BlockRAM"};
    static unsigned 2 i;
    unsigned 4 x;          // x is un-initialized
    interface bus_in(unsigned 4 i) I();
    interface bus_out() O(unsigned 4 o = x);
    while(1)
    {
        rax[i] = I.i;
        i++;
        x = rax[i];        // RAM only read into x...
        x = 1;             // ...but x also written to from elsewhere
    }
}
```

MPRAM Example 1: transform is performed

```
void main(void)
{
    mpram
    {
        ram unsigned 4 rax1[4];
        ram unsigned 4 rax2[4];
    } max with {block=1};
    static unsigned 2 i1, i2;
    unsigned 4 x;           // x is un-initialized
    unsigned 4 y;           // y is un-initialized
    interface bus_in(unsigned 4 i1) I1();
    interface bus_out() O1(unsigned 4 o1 = x);
    interface bus_in(unsigned 4 i2) I2();
    interface bus_out() O2(unsigned 4 o2 = y);
    while(1)
    {
        max.rax1[i1] = I1.i1;
        max.rax2[i2] = I2.i2;
        i1++;
        i2++;
        x = max.rax1[i1];   // mpram port rax1 only read into x
        y = max.rax2[i2];   // mpram port rax2 only read into y
    }
}
```

MPRAM Example 2: transform is not performed (port 'rax2' does not read into a register)

```
void main(void)
{
    mpram
    {
        ram unsigned 4 rax1[4];
        ram unsigned 4 rax2[4];
    } max with {block=1};
    static unsigned 2 i1, i2;
    unsigned 4 x;          // x is un-initialized
    interface bus_in(unsigned 4 i1) I1();
    interface bus_out() O1(unsigned 4 o1 = x);
    interface bus_in(unsigned 4 i2) I2();
    // port rax2 read directly into an interface, not a 'pipeline' register
    interface bus_out() O2(unsigned 4 o2 = max.rax2[i2]);

    while(1)
    {
        max.rax1[i1] = I1.i1;
        max.rax2[i2] = I2.i2;
        i1++;
        i2++;
        x = max.rax1[i1];   // mpram port rax1 only read into x...
    }
}
```

Targeting external synchronous RAMs

Off-chip synchronous SRAMs can be specified in exactly the same way as on-chip synchronous SRAMs, with the addition of the `rclkpos`, `wclkpos`, `clkpulselen` and `clk` specifications. `clk` specifies the pin on which the generated RAMCLK will appear, when the SSRAM in question is external (`offchip = 1`).

Example

```
macro expr addressPins = {Pin List...};
macro expr dataPins = {Pin List...};
macro expr csPins = {Pin List...};
macro expr wePins = {Pin List...};
macro expr oePins = {Pin List...};
macro expr clkPins = {Pin List...};

ram unsigned 32 ExtBank[1024] with {offchip = 1,
      addr = addressPins,
      data = dataPins,
      cs = csPins,
      we = wePins,
      oe = oePins,
      westart = 2,
      welength = 1,
      rclkpos = {1.5, 2.5},
      wclkpos = {1.5, 2.5, 3.5},
      clkpulselen = 0.5,
      clk = clkPins};
```

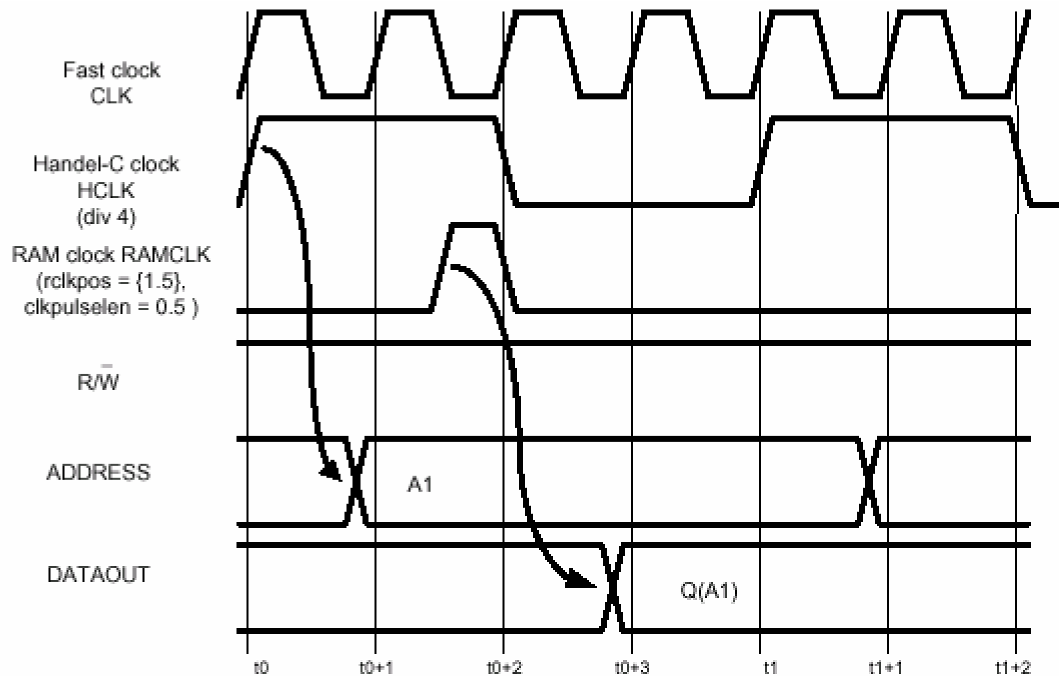
Flow-through SSRAM example

```
ram unsigned 18 FlowBank[1024]
    with {block = 1,
          westart = 2,
          welength = 1,
          rclkpos = {1.5},
          wclkpos = {2.5, 3.5},
          clkpulselen = 0.5};
```

This code instructs the compiler to build hardware to generate SSRAM control signals as shown below. It is also applicable for reading from block RAMs in Actel and Xilinx FPGAs and Altera ESB and tri-matrix memories.

Read cycle for a flow-through SSRAM

The timing diagram shows a read-cycle from a flow-through SSRAM.

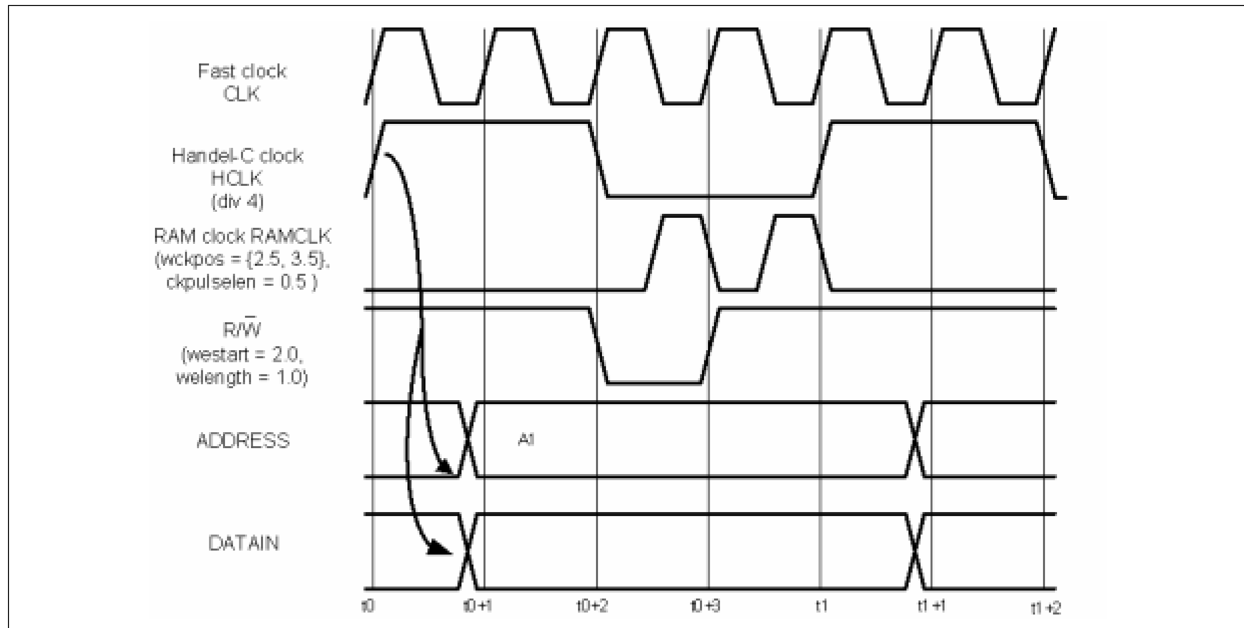


The rising HCLK edge at t0 initiates the read cycle. Some time later, the address A1 is set up, which is sampled somewhere in the middle of the HCLK cycle: t0+1.5 in this case. By the time the next HCLK rising edge occurs at t1, the data is available for reading. The cycle completes within one Handel-C clock cycle.

Write cycle for a flow-through SSRAM

Flow-through SSRAMs perform a "late" write cycle; the data is clocked in one clock cycle after the address is sampled.

The timing diagram shows the complete write cycle.



The HCLK rising edge at t0 initiates the write cycle, causing the ADDRESS and DATAIN signals to change. Two cycles of RAMCLK are needed to clock the new data into the RAM at the specified address: the first to sample the address, the second to sample the data. However, since we're not expecting to read from the RAM's output, we can wait until the last possible moment. In this case, the two rising edges of RAMCLK occur at t0+2.5 and t0+3.5.

The write enable signal must be low during the rising edge of RAMCLK that samples the address, but not during the one that samples the data. This can be done by setting `westart` and `welength` as shown. The entire cycle completes within a single Handel-C clock cycle.

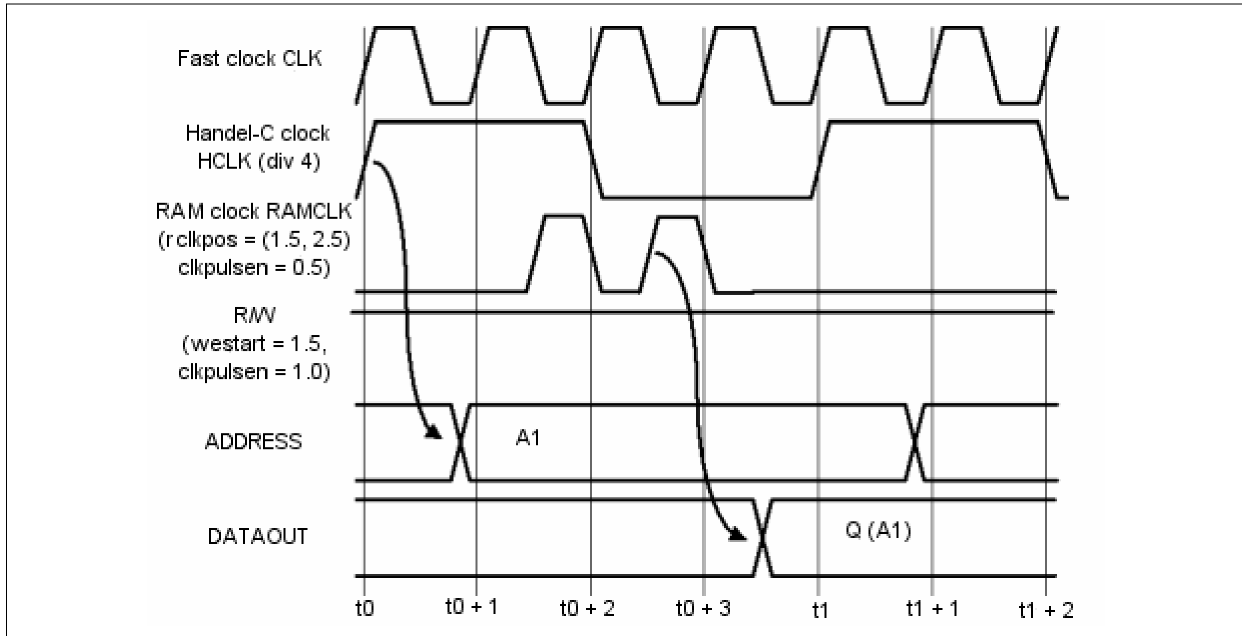
Pipelining off-chip SSRAM example

This method of pipelining SSRAM is most suitable for off-chip RAMs. For on-chip SSRAM, it is usually more efficient to use a pipelining transformation.

```
ram unsigned 18 PipeBank[1024]
    with {block = 1,
          westart = 1.5,
          welength = 1,
          rclkpos = {1.5, 2.5},
          wclkpos = {1.5, 2.5, 3.5},
          clkpulselen = 0.5};
```

Read cycle for a pipelined-output SSRAM

The timing diagram shows the read cycle

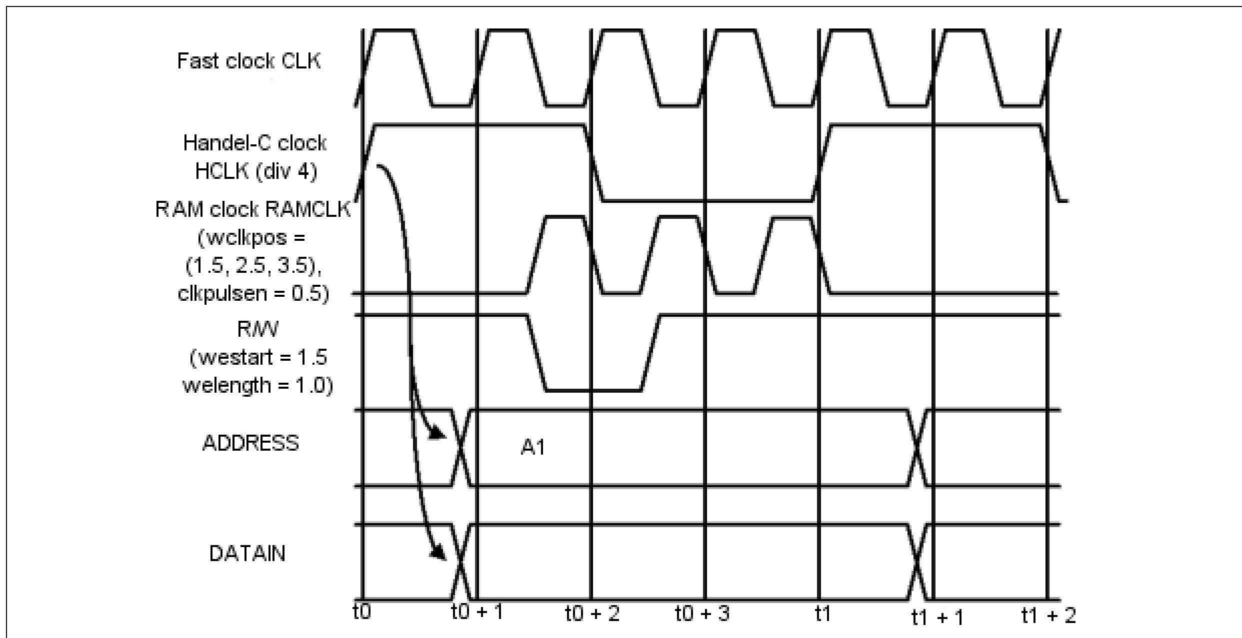


This read cycle is very similar to that for a flow through RAM. The rising HCLK edge at t0 initiates the read cycle. Some time later, the address A1 is set up, which is sampled somewhere near the middle of the HCLK cycle: (t0+1.5 in this case). The RAM contents at address A1 are then piped to the RAM's output register; it must be made available at the RAM output. A second RAMCLK pulse (at t0+2.5 in this case) is used to do this. By the time the next HCLK rising edge occurs at t1, the data is available for reading by the Handel-C design. The cycle completes within one Handel-C clock cycle.

Write cycle for a pipelined-output SSRAM

Pipelined-output SSRAMs perform a "late-late" write cycle. This means that data is written to memory two clock cycles after the address is sampled.

The timing diagram shows the complete cycle.



The HCLK rising edge at t0 initiates the write cycle, causing the ADDRESS and DATAIN signals to change. Three cycles of RAMCLK are needed to clock the new data into the RAM at the specified address: the first to sample the address and the third to sample the data. Since you will not read from the RAM on a write strobe, you can sample the data as late as possible to give the circuit maximum time to set up the data. In this case, the three rising edges of RAMCLK occur at t0+1.5, t0+2.5 and t0+3.5.

The write enable signal must be low during the rising edge of RAMCLK that samples the address, but not during the one that samples the data. This can be done by setting `westart` and `welength` as shown. The entire cycle completes within a single Handel-C clock cycle.

10.3.3 Targeting Stratix and Cyclone memory blocks

Altera Stratix, Stratix GX and Stratix II devices have 3 types of embedded memory: M512, M4K and M-RAM. Cyclone and Cyclone II devices only have M4K. You can specify what type of memory you want to build by using the `block` specification.

Type of memory	block specification
M512	`with {block = "M512"}`
M4K	`with {block = "M4K"}`
M-RAM	`with {block = "M-RAM"}`

If you do not use the `block` specification the memory is set to "AUTO" and Quartus determines the most appropriate memory type when you place and route.

All Stratix memories are fully synchronous. If you try to make them asynchronous, for example by using the `westart` and `welength` specifications, you will get a compiler error.

M-RAM cannot be initialized. This means that you cannot have a ROM built out of M-RAM. You will get a compiler error if you build a ROM using the `with {block = "M-RAM"}` specification.

M512 memory cannot be configured as a bi-directional dual-port MPRAM. If you try to create this, the compiler will issue a warning.

Example

```
set family = AlteraStratix;
set part = "EP1S10B672C7";
set clock = external;

ram unsigned 8 autoRam[16]; // Let Quartus select a suitable memory
structure
ram unsigned 8 m512Ram[16] with {block = "M512"}; // Use M512 blocks
ram unsigned 8 m4kRam[16] with {block = "M4K"};  // Use M4K blocks
ram unsigned 8 mRam[16] with {block = "M-RAM"};  // Use M-RAM blocks

void main(void)
{
    autoRam[0] = 1;
    m512Ram[0] = 1;
    m4kRam[0] = 1;
    mRam[0] = 1;

    ...etc...
}
```

10.3.4 Using on-chip RAMs in Actel devices

On-chip RAMs in Actel ProASIC and ProASIC+ devices use the embedded memory structures, which are of a fixed width and depth. These blocks can be combined to create deeper and wider memory spaces. When writing Handel-C programs, you must be careful not to exceed the number of memory blocks in the target device or the design will not place and route successfully. It is possible to use RAMs that do not match one of the width/depth combinations, but memory space may be wasted.

Synchronous and asynchronous access

Memory blocks in ProASIC and ProASIC+ parts can be configured to be either synchronous or asynchronous. If you do not apply any clock or write-enable specifications, Handel-C will create RAMs with a synchronous write port and asynchronous read port.

If you apply clock position specifications to the RAM, the read and write ports will both be synchronous.

If you apply any of the write-enable specifications (`westart`, `welength` or `wegate`) to the RAM, both write and read access will be asynchronous.

When declaring a memory as a MPRAM, if you only apply write-enable specifications to the read port AND you apply clock specifications to the write port, you will get a compiler error, as you cannot have an asynchronous write port and a synchronous read port.

Initialization

Actel memories may not be initialized.

10.3.5 Using on-chip RAMs in Altera devices

EAB structures

On-chip RAMs in Altera Flex10K devices use the EAB structures. These blocks can be configured in a number of data width/address width combinations. When writing Handel-C programs, you must be careful not to exceed the number of EAB blocks in the target device or the design will not place and route successfully. While it is possible to use RAMs that do not match one of the data width/address width combinations, EAB space may be wasted by such a RAM.

Synchronous and asynchronous access

RAM blocks in Flex, Apex, Excalibur and Mercury parts can be configured to be either synchronous or asynchronous. If you do not apply any clock or write-enable specifications, Handel-C will create RAMs with a synchronous write port and asynchronous read port as long as the target hardware supports it.

If you apply clock position specifications to the RAM, the read and write ports will both be synchronous.

If you apply any of the write-enable specifications (`westart`, `welength` or `wegate`) to the RAM, both write and read access will be asynchronous.

When declaring a memory as a MPRAM, if you only apply write-enable specifications to the read port AND you apply clock specifications to the write port, you will get a compiler error, as you cannot have an asynchronous write port and a synchronous read port.

Initialization

RAM/ROM initialization files with a `.mif` extension will be generated on compilation to feed into the Max Plus II or Quartus software. This process is transparent if they are in the same directory as the EDIF (`.edf` extension) file generated by the Handel-C compiler.

Creating RAMs without an inverted clock

If you declare a single-port RAM for Altera Flex, Apex 20, Mercury or Excalibur devices, the Handel-C compiler converts this into an MPRAM with a ROM port and a WOM port. This removes the inverted clock, and so increases the possible clock rate. If you want to remove the inverted clock from an on-chip memory on an ApexII device, you need to do this manually by creating an MPRAM instead of a RAM. The compiler does not do this automatically as the hardware created for an MPRAM is larger than that for a RAM on ApexII devices.

Stratix and Cyclone memories are totally synchronous, so creating an MPRAM with a ROM and a WOM port does not automatically result in the inverted clock being removed. Instead, you can pipeline the MPRAM, or you can customize the clock using the `rclkpos`, `wclkpos` and `clkpulselen` specifications.

10.3.6 Using on-chip RAMs in Xilinx devices

Handel-C supports the synchronous RAMs on Virtex series and Spartan-II and Spartan-3 parts directly simply by declaring a RAM or ROM. For example:

```
ram unsigned 6 x[34];
```

This will declare a RAM with 34 entries, each of which is 6 bits wide.

When writing Handel-C programs, you must be careful not to exceed the number of memory blocks in the target device or the design will not place and route successfully.

10.3.7 Using external ROMs

An external ROM is declared as an external RAM with an empty write enable pin list. For example:

```
ram unsigned 8 ExtROM[16384] with {
       offchip = 1,
       data = {"P1", "P2", "P3", "P4",
               "P5", "P6", "P7", "P8"},
       addr = {"P9", "P10", "P11", "P12",
               "P13", "P14", "P15", "P16",
               "P17", "P18", "P19", "P20",
               "P21", "P22"},
       we = {},
       oe = {"P24"},
       cs = {"P25"}};
```

Note that no `westart`, `welength` or `wegate` specification is required since there is no write strobe signal on a ROM device.

10.3.8 Connecting to RAMs in foreign code

You can create ports to connect to a RAM by using the `ports = 1` specification to your memory definition. This will generate VHDL, Verilog or EDIF wires which can be connected to a component created elsewhere. The ports specification cannot be used in conjunction with the `offchip=1` specification, but all other specifications will apply.

The interface generated will have separate read (output) and write (data) ports, write enable, data enable and clock wires. This ensures that it can be connected to any device. Pin names provided in the `addr`, `data`, `cs`, `we`, `oe`, and `clk` specifications will be passed through to the generated EDIF. They are not passed through to VHDL or Verilog, since VHDL and Verilog interfaces are generated as n-bit wide buses rather than n 1-bit wide wires. This means that it is ambiguous to specify a separate identifier for each wire. If they are used when compiling to VHDL or Verilog, the compiler issues a warning.

For VHDL or Verilog output, the compiler generates meaningful port names. For example, with the following RAM declaration compiled to VHDL:

```
ram unsigned 4 rax[4] with
        {ports = 1, data = dataPins, addr = addrPins,
         we = wePins, cs = csPins, oe = oePins};
```

the compiler will warn that all the pins specifications have been ignored, and will generate an interface in VHDL with the following ports:

```
component rax_SPPort
port(
rax_SPPort_addr: in unsigned(1 downto 0);
rax_SPPort_clk: in std_logic;
rax_SPPort_cs: in std_logic;
rax_SPPort_data_en: in std_logic;
rax_SPPort_data_in: out unsigned(3 downto 0);
rax_SPPort_data_out: in unsigned(3 downto 0);
rax_SPPort_oe: in std_logic;
rax_SPPort_we: in std_logic
);
```

The port names consist of the memory name (`rax` in this case), description of the memory type (`SPPort :` single port in this case) and an identifier describing the ports function.

A clock port will always be generated.

If you use the `ports` specification with an MPRAM, a separate interface will be generated for each port.

Generating an interface to a foreign code RAM

```
set family = XilinxVirtex;
set part = "V1000BG560-4";
set clock = external "C1";

unsigned 4 a;
ram unsigned 4 rax[4] with {ports = 1};

void main(void)
{
    static unsigned 2 i = 0;

    while(1)
    {
        par
        {
            i++;
            a++;
            rax[i] = a;
         }
         a = rax[i];
    }
}
```

The declaration of rax would produce wires

```
rax_SPPort_addr<0>        // Address
rax_SPPort_addr<1>
rax_SPPort_data_in<0>     // Data In
rax_SPPort_data_in<1>
rax_SPPort_data_in<2>
rax_SPPort_data_in<3>
rax_SPPort_data_out<0>    // Data Out
rax_SPPort_data_out<1>
rax_SPPort_data_out<2>
rax_SPPort_data_out<3>
rax_SPPort_data_en        // Data Enable
rax_SPPort_clk            // Clock
rax_SPPort_cs             // Chip Select
rax_SPPort_oe             // Output Enable
rax_SPPort_we             // Write Enable
```

Generating an interface to a foreign code MPRAM

```
set family = XilinxVirtex;
set part = "V1000BG560-4";
set clock = external "C1";

unsigned 4 a;

mpram Mpaz
{
    wom unsigned 4 wox[4];
    rom unsigned 4 rox[4];
} mox with {ports = 1};

void main(void)
{
    static unsigned 2 i = 0;

    while(1)
    {
        par
        {
            i++;
            a++;
            mox.wox[i] = a;
        }
        a = mox.rox[i];
    }
}
```

The declaration of the read only port `rox` would produce wires

```
mox_rox_addr_0    // Address
mox_rox_addr_1
mox_rox_clk       // Clock
mox_rox_cs        // Chip select
mox_rox_data_en   // Data enable
mox_rox_oe        // Output enable
mox_rox_we        // Write enable
mox_rox_data_in_0 // Data into Handel-C, out from foreign code memory
mox_rox_data_in_1
mox_rox_data_in_2
mox_rox_data_in_3
```

The declaration of the read only port `wox` would produce wires

```
mox_wox_addr_0      // Address
mox_wox_addr_1
mox_wox_clk         // Clock
mox_wox_cs          // Chip select
mox_wox_data_en     // Data enable
mox_wox_data_out_0  // Data out from Handel-C, into foreign code memory
mox_wox_data_out_1
mox_wox_data_out_2
mox_wox_data_out_3
mox_wox_oe          // Output enable
mox_wox_we          // Write enable
```

10.3.9 Using other RAMs

The interface to other types of RAM such as DRAM should be written by hand using interface declarations. Macro procedures can then be written to perform complex or even multi-cycle accesses to the external device.

11 Interfacing with external hardware

All off-chip accesses are based on the idea of a bus which is just a collection of external pins. Handel-C provides the ability to read the state of pins for input from the outside world and set the state of pins for writing to the outside world. Tri-state buses are also supported to allow bi-directional data transfers through the same pins.

The pins used may be defined in Handel-C by using pin specifications (e.g. `data`). If this is omitted, the pins will be left unconstrained and can be assigned by the place and route tools.

Note that Handel-C provides no information about the timing of the change of state of a signal within a Handel-C clock cycle. Timing analysis is available from the FPGA or PLD manufacturer's place-and-route tools.

Handel-C programs can also interface to external logic (other Handel-C programs, programs written in VHDL or Verilog etc.) by using user-defined interfaces or Handel-C ports.

Your license may not allow you to use interfaces. If this is the case you can only interface to external devices using macros provided in any Celoxica libraries you have licenses for, such as PAL.

11.1 Interface sorts

Handel-C provides a number of predefined interface sorts.

"bus-type" interfaces (`bus_*`) generate the hardware for buses connected to pins.

"port-type" interfaces (`port_*`) generate the hardware for floating ports (buses which are not connected to pins).

These can be of any width, and can carry signals between different sections of Handel-C code, or to software or hardware beyond the Handel-C program.

You can also define your own sorts to interface to external blocks of code ("generic" or custom interface sorts).

Predefined interface sorts

Sort identifier	Description
bus_in	Input bus from pins
bus_latch_in	Registered input bus from pins
bus_clock_in	Clocked input bus from pins
bus_out	Output bus to pins
bus_ts	Bi-directional tri-state bus
bus_ts_latch_in	Bi-directional tri-state bus with registered input
bus_ts_clock_in	Bi-directional tri-state bus with clocked input
port_in	Input port from logic
port_out	Output port to logic

Custom or generic interface sorts

You can define your own interface sorts to connect to non-Handel-C objects:

- Hardware descriptions written in another language.
  VHDL, Verilog and EDIF are currently supported. For a VHDL code interface, the interface sort would be the name of the VHDL entity. For a Verilog code interface, the interface sort would be the name of the Verilog module.

- Native PC object code used in simulation.
  Programs that run on your PC for simulation and connect to a Handel-C interface are known as plugins. There are special port specifications to enable you to connect user-defined interfaces with a plugin for simulation. These are `extlib`, `extfunc`, and `extinst`.

11.1.1 Reading from external pins bus_in

The `bus_in` interface sort allows Handel-C programs to read from external pins. Its general usage is:

```
interface bus_in(type portName)
    Name()
        with {data = {Pin List}};
```

Reading the bus is performed by accessing the identifier ***Name.portName*** as a variable which will return the value on those pins at that clock edge. If no input port name is given, the port name defaults to `in`.

Example

```
interface bus_in(int 4 To) InBus()
    with {data = {"P4", "P3", "P2", "P1"}};
int 4 x;

x = InBus.To;
```

This declares a bus connected to pins P1, P2, P3 and P4 where pin P4 is the most significant bit and pin P3 is the least significant bit.

The variable `x` is set to the value on the external pins. The type of `InBus.To` is `int 4` as specified in the type list after the `bus_in` keyword.

11.1.2 Registered reading from external pins: bus_latch_in

The `bus_latch_in` interface sort is similar to `bus_in` but allows the input to be registered on a condition. This may be required to sample the signal at particular times. Its general usage is:

```
interface bus_latch_in(type portName)
      Name(type conditionPortName=Condition)
          with {data = {Pin List}};
```

Reading the bus is performed by accessing the identifier ***`Name.portName`*** as a variable which will return the value on those pins at that clock edge. If no input port name is given, the port name defaults to `in`. ***`Condition`*** specifies a signal that is used to clock the input registers in the FPGA or PLD. The rising edge of this signal clocks the external signal into the internal value.

Example

```
unsigned 1 get;
int 4 x;

interface bus_latch_in(int 4 To)
        InBus(unsigned 1 condition = get)
        with {data = {"P4", "P3", "P2", "P1"}};

get = 0;
get = 1;      // Register the external value
x = InBus.To; // Read the registered value
```

11.1.3 Clocked reading from external pins: bus_clock_in

The bus_clock_in interface sort is similar to the bus_in interface sort but allows the input to be clocked continuously from the Handel-C global clock. This may be required to synchronize the signal to the Handel-C clock. Its general usage is:

```
interface bus_clock_in(type portName) Name()
   with {Specs};
```

Reading the bus is performed by accessing the identifier *Name.portName* as a variable which will return the value on those pins at that clock edge. If no input port name is given, the port name defaults to in. The rising edge of the Handel-C clock clocks the external signal into the internal value. For example:

```
interface bus_clock_in(int 4 InTo) InBus() with
            {data = {"P4", "P3", "P2", "P1"}};

x = InBus.InTo; // Read flip-flop value
```

11.1.4 Writing to external pins: bus_out

The bus_out interface sort allows Handel-C programs to write to external pins. Its general usage is:

```
interface bus_out()
   Name(type portName=Expression)
      with {data = {Pin List}};
```

A specific example is:

```
interface bus_out () OutBus(int 4 OutPort=x+y)
    with {data = {"P4", "P3", "P2", "P1"}};
```

This declares a bus connected to pins 1, 2, 3 and 4 where pin 4 is the most significant bit and pin 1 is the least significant bit. The value appearing on the external pins is the value of the expression x+y at all times.

11.1.5 Bidirectional data transfer: bus_ts

The bus_ts interface sort allows Handel-C programs to perform bi-directional off-chip communications via external pins. Its general usage is:

```
interface bus_ts (type inPortName)
    Name(type outPortName = Value,
        type conditionPortName = Condition)
   with {Specs};
```

Value is an expression giving the value to output on the pins. *Condition* is an expression giving the condition for driving the pins. When *Condition* is non-zero (i.e. true), the value of *Value* is driven on the pins. When the value of *Condition* is zero, the pins are tri-stated and the value of the external bus can be read using the identifier *Name. inPortName*. If *inPortName* is not defined, the port name defaults to `in`.

If you attempt to read from a tri-state bus when it is in write mode (i.e. condition is non-zero), you will get the value that you are writing to the bus.

Example

```
unsigned 1 condition;
int 4 x;

interface bus_ts(int 4 read)
    BiBus(int write=x+1,
    unsigned 1 enable= condition)
    with {data = {"P4", "P3", "P2", "P1"}};

condition = 0;  // Tri-state the pins
x = BiBus.read; // Read the value
condition = 1;  // Drive x+1 onto the pins
```

This example reads the value of the external bus into variable `x` and then drives the value of `x + 1` onto the external pins.

Take care when driving tri-state buses that the FPGA/PLD and another device on the bus cannot drive simultaneously as this may result in damage to one or both of them.

11.1.6 Bidirectional data transfer with registered input: bus_ts_latch_in

The `bus_ts_latch_in` interface sort allows Handel-C programs to perform bidirectional off-chip communications via external pins with inputs registered on a condition. Its general usage is:

```
interface bus_ts_latch_in (type inPortName)
    Name(type outPortName = Value,
        type conditionPortName = Condition,
        type clockPortName = Clock)
    with {Specs};
```

Value is an expression giving the value to output on the pins. *Condition* is an expression giving the condition for driving the pins. *Clock* is an expression giving the signal to clock the input from the pins. When *Condition* is non-zero (i.e. true), the value of *Value* is driven on the pins. When the value of *Condition* is zero, the pins are tri-stated and the registered value of the external bus can be read using the identifier *Name.inPortName*. If *inPortName* is not defined, the port name defaults to `in`.

The rising edge of the value of the third expression clocks the external values through to the internal values on the chip.

If you attempt to read from a tri-state bus when it is in write mode (i.e. condition is non-zero), you will get the value that you are writing to the bus.

Example

```
int 1 get;
unsigned 1 condition;
int 4 x;

interface bus_ts_latch_in(int 4 read)
    BiBus(int write = x+1,
    unsigned 1 enable = condition,
    unsigned 1 clock_port = get)
    with {data = {"P4", "P3", "P2", "P1"}};

condition = 0; // Tri-state external pins
get = 0;
get = 1;        // Register external value
x = BiBus.read; // Read registered value
condition = 1;  // Drive x+1 onto external pins
```

This example samples the external bus and reads the registered value into variable `x` and then drives the value of `x + 1` onto the external pins.

Take care when driving tri-state buses that the FPGA/PLD and another device on the bus cannot drive simultaneously as this may result in damage to one or both of them.

11.1.7 Bidirectional data transfer with clocked input: bus_ts_clock_in

The `bus_ts_clock_in` interface sort allows Handel-C programs to perform bidirectional off-chip communications via external pins with inputs clocked continuously with the Handel-C clock. Its general usage is:

```
interface bus_ts_clock_in (type inPortName)
    Name(type outPortName = Value,
        type conditionPortName = Condition)
    with {Specs};
```

Value is an expression giving the value to output on the pins. *Condition* is an expression giving the condition for driving the pins. When *Condition* is non-zero (i.e. true), the value of *Value* is driven on the pins. When the value of *Condition* is zero, the pins are tri-stated and the value of the external bus can be read using the identifier *Name*. *inPortName*. If *inPortName* is not defined, the port name defaults to `in`.

If you attempt to read from a tri-state bus when it is in write mode (i.e. condition is non-zero), you will get the value that you are writing to the bus.

The rising edge of the Handel-C clock reads the external values into the internal flip-flops on the chip. For example:

```
unsigned 1 condition;
int 4 x;

interface bus_ts_clock_in (int 4 read)
     BiBus(int 4 writePort=x+1,
           unsigned 1 enable=condition)
           with {data = {"P4", "P3", "P2", "P1"}};

condition = 0;  // Tri-state external pins
x = BiBus.read; // Read registered value
condition = 1;  // Drive x+1 onto external pins
```

This example reads the value from the flip-flop into variable `x` and then drives the value of `x + 1` onto the external pins.

Take care when driving tri-state buses that the FPGA/PLD and another device on the bus cannot drive simultaneously as this may result in damage to one or both of them.

11.1.8 Example hardware interface

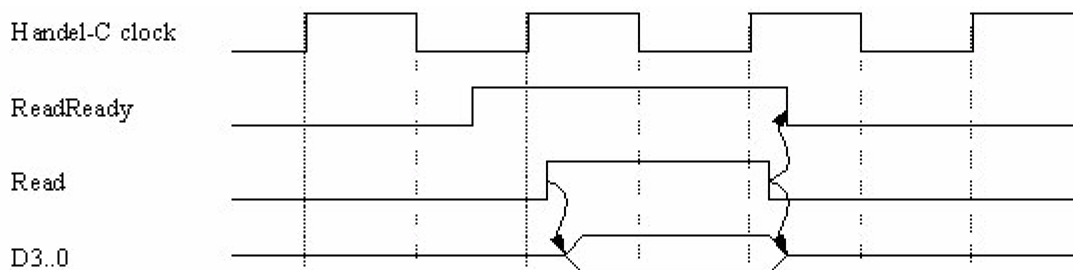
The example shows the use of buses. The scenario is of an external device connected to the FPGA/PLD which may be read from or written to. The device has a number of signals connected to the FPGA/PLD.

Signals connected

Signal Name	FPGA pin	Description
D3..0	1, 2, 3, 4	Data Bus
Write	5	Write strobe
Read	6	Read strobe
WriteRdy	7	Able to write to device
ReadRdy	8	Able to read from device

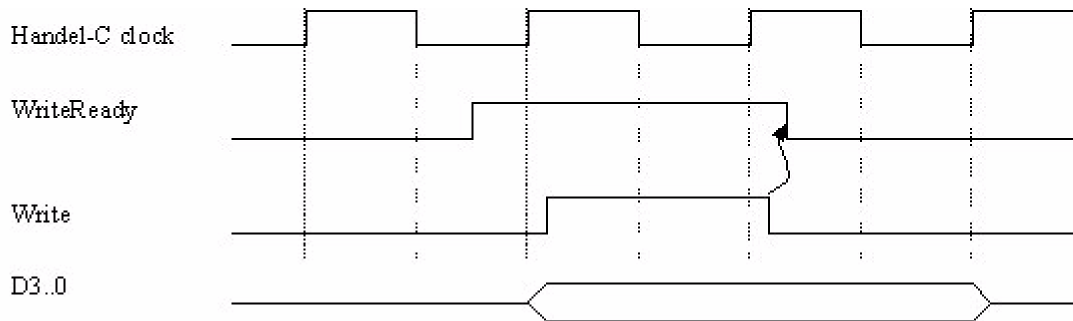
Read cycle timing

A read from the device is performed by waiting for ReadRdy to become active (high). The Read signal is then taken high for one clock cycle and the data sampled on the falling edge of the strobe.



Write cycle timing

A write to the device is performed by waiting for `WriteRdy` to become active (high). The `Write` signal is then taken high for one clock cycle while the data is driven to the device by the FPGA. The device samples the data on the falling edge of the Write signal.



Bus declarations

The first stage of the code declares the buses associated with each of the external signals.

```
int 4 Data;
int 1 En = 0;
interface bus_ts_clock_in(int 4 DataIn)
    dataB(int outPort=Data, int EnableSignal=En) with
    {data = {"P4", "P3", "P2", "P1"}};

int 1 Write = 0;
interface bus_out() writeB(int WriteSignal = Write) with
     {data = {"P5"}};

int 1 Read = 0;
interface bus_out() readB(int readSignal=Read) with
    {data = {"P6"}};

interface bus_clock_in(int 1 wr)
    WriteReady() with {data = {"P7"}};

interface bus_clock_in(int 1 readySignal)
    ReadReady() with {data = {"P8"}};
```

```
void main (void)
{
    int 4 Data, Reg;

    // Read word from external device
    while (ReadReady.readySignal == 0)
        delay;

    Read = 1; // Set the read strobe
    par
    {
        Data = dataB.DataIn; // Read the bus
        Read = 0; // Clear the read strobe
    }

    // Write one word back to external device
    Reg = Data + 1;
    while (WriteReady.wr == 0)
        delay;

    par
    {
        En = 1; // Drive the bus
        Write = 1; // Set the write strobe
    }

    Write = 0; // Clear the write strobe
    En = 0; // Stop driving the bus
}
```

Writing data

You can change the values on the output buses by setting the values of the `Data`, `Write` and `Read` variables. You can drive the data bus with the contents of `Data` by setting `En` to 1.

The variables that drive buses have been initialized to 0. That means that these variables must be static or global. This may be important when driving write strobes. Care should be taken during configuration that the FPGA pins are disconnected in some way from the external devices because the FPGA pins become tri-state during this time.

The main program

The main program reads a word from the external device before writing one word back.

```
void main (void)
{
    int 4 Data, Reg;

    // Read word from external device
    while (ReadReady.readySignal == 0)
        delay;
    Read = 1;      // Set the read strobe
    par
    {
        Data = dataB.DataIn; // Read the bus
        Read = 0;  // Clear the read strobe
    }

    // Write one word back to external device
    Reg = Data + 1;
    while (WriteReady.wr == 0)
        delay;
    par
    {
        En = 1;    // Drive the bus
        Write = 1; // Set the write strobe
    }
    Write = 0;     // Clear the write strobe
    En = 0;        // Stop driving the bus
}
```

Note that during the write phase, the data bus is driven for one clock cycle after the write strobe goes low to ensure that the data is stable across the falling edge of the strobe.

11.2 Simulating interfaces

You can combine the hardware and simulation versions of your program by using the `#ifdef` construct. For example:

```
#define SIMULATE

#ifdef SIMULATE
{
   ...
}
#else
{
   ...
}
#endif
```

There are several ways to simulate the reading and writing of data across an interface.

Bus-type and port-type interfaces

If you have a bus-type interface or a port-type interface (`port_in` or `port_out`) you can use the `infile` and `outfile` specifications to read and write data. (Bus-type interfaces are `bus_in`, `bus_latch_in`, `bus_clock_in`, `bus_out`, `bus_ts`, `bus_ts_latch_in` and `bus_ts_clock_in`).

For example:

```
set clock = external "P1";

unsigned 8 out;
interface port_in(unsigned 8 i) pi() with {infile = "in.txt"};
interface port_out() po(out) with {outfile = "out.txt"};
void main (void)
{
   do
   {
      out = pi.i;
   }while(out != 0);
}
```

`infile` and `outfile` can only connect to files with data in a simple format. If your data is more complex, you could write a C/C++ function and call it to bring in the data.

If you want to model the hardware as well as the functionality of your design, you will need to co-simulate your interface with a model of the component to which it will be connected (see below).

Generic interfaces

If you have written a custom (generic) interface, you will need to co-simulate the interface with a model of the component to which it will be connected in hardware. If you write the model in Handel-C, you can co-simulate it with your Handel-C interface using

dkconnect.dll. To synchronize the simulations, use dksync.dll. If your model is in VHDL or Verilog, you can co-simulate it with your Handel-C design using the Co-simulation Bridge for ModelSim provided in the Platform Developer's Kit.

11.3 Buses and the simulator

The Handel-C simulator cannot simulate buses directly, because the simulation of buses cannot determine when input and output should occur. The recommended process for debugging is:

For simple data, use a channel or a chanin/chanout to connect to a file. This is the simplest method.

For more complex buses/interfaces, write a C/C++ function and call it to bring in data. This allows you to operate on the data or read it in a complex format. This models functionality but not hardware.

To model buses accurately, use the Plugin Library to write a plugin which can be co-simulated. This is precise and allows you to read I/O signals using the Waveform Analyzer, but can be slow and cumbersome.

Using preprocessor definitions

By using the #define and #ifdef...#endif constructs of the preprocessor, it is possible to combine both the simulation and hardware versions of your program into one.

Channel example

```
#define SIMULATE
#ifdef SIMULATE
    input ? value;
#else
    value = BusIn.in;
#endif
```

External function call example

```
#define SIMULATE

#ifdef SIMULATE
    extern "C++" int 8 bus_input_function(void);
    data_in = bus_input_function();
#else
    interface bus_in(int 8 in) BusIn();
    data_in = BusIn.in;
#endif
```

Example with plugin

To simulate a tri-state bus:

```
interface bus_ts (int 32 in with
   {extlib = "MyPlugin.dll", extinst = "1", extfunc = "DataBusIn"})
   DataBus(int 32 out = DataOut with {extlib = "MyPlugin.dll",
      extinst = "1", extfunc = "DataBusOut"},
      int 1 enable = !WriteBus.in with {extlib = "MyPlugin.dll",
      extinst = "1", extfunc = "DataBusEnable"})
      with {data = pinList};
```

In this case, the functions `DataBusIn`, `DataBusOut` and `DataBusEnable` would be provided in the plugin `MyPlugin.dll` and called by the simulator. The `extlib`, `extfunc` and `extinst` specifications are ignored if compiled to HDL so the interface definition does not have to be within an `#ifdef`.

11.4 Merging pins

11.4.1 Merging clock pins

You can merge clock pins as long as:

- any pins specifications given to the two clocks match.
- there are no conflicts between any timing specifications given to the clocks.

For example, if you specified two clock domains in the same project with the following code:

```
set clock = external "C1" with {rate = 10}; //clock declaration in file
one.hcc
set clock = external "C1" with {rate = 20}; //clock declaration in file
two.hcc
```

you would get a compiler error, as the `rate` specifications don't match.

If one of the clocks is divided you need to divide the value of the `rate` specification to match. For example:

```
set clock = external "C1" with {rate = 10}; // file one.hcc
set clock = external_divide 3 "C1" with {rate = 3.3333333333333333}; //
file two.hcc
```

If you need to use decimal places to specify the rate for the divided clock, the compiler will round up the value to the nearest whole number as long as you use at least 16 decimal places (3 x3.3333333333333333 is rounded up to 10).

11.4.2 Merging input pins

Input pins can be merged so that pins can be read simultaneously into multiple variables. This can be done by specifying multiple interfaces (bus_in, bus_clock_in, bus_latch_in) which have some pins in common. If required, a different subset of pins can be specified for each instance of the interface. For example:

```
interface bus_in(int 8 wide) wideDataBus() with
   {data ={"P7", "P6", "P5", "P4", "P3",
      "P2", "P1", "P0"}};
interface bus_in(int 3 thin) thinDataBus() with
   {data ={"P5", "P4", "P3"}};
```

wideDataBus.wide would give the values of pins P0 – P7, whereas thinDataBus.thin would give the three bit value on pins P3, P4 and P5.

If the input pins have an intime specification, you need to ensure that these match.

11.4.3 Merging tri-state pins

Tri-state bus pins can be merged, though doing so will generate a compiler warning, as the compiler cannot detect whether the outputs for both pins might be enabled at the same time. If both outputs are enabled at the same time, the result is undefined. If you have used any intime and outtime specifications, make sure that they match.

You might wish to merge output pins for a tri-state bus if you wished to switch the circuit connections from one external piece of logic to another. For example:

```
int 1 en1, en2;
int 4 x, y;
interface bus_ts_clock_in (int 4 read)
      BiBus1(int 4 writePort=x+1, unsigned 1 enable = (en1==1))
        with {data = {"P4", "P3", "P2", "P1"}};
interface bus_ts_clock_in (int 4 read)
      BiBus2(int 4 writePort=y+1, unsigned 1 enable = (en2==1))
        with {data = {"P4", "P3", "P2", "P1"}};
```

Take care when driving tri-state buses that the FPGA/PLD and another device on the bus cannot drive simultaneously as this may result in damage to one or both of them.

11.5 Timing considerations for buses

bus_in interfaces

This form of bus is built with no register between the external pin and the points inside the FPGA or PLD where the data is used. If the value on the external pin changes asynchronously with the Handel-C clock then routing delays within the FPGA can cause the value to be read differently in different parts of the circuit. The solution to this problem is to use either a `bus_latch_in` or a `bus_clock_in` interface sort.

bus_out interfaces

The output value on pins cannot be guaranteed except at rising Handel-C clock edges. In between clock edges, the value may be in the process of changing. Since the routing delays through different parts of the logic of the output expression are different, some pins may change before others giving rise to intermediate values appearing on the pins. This is particularly apparent in deep combinational logic. Adding a flip-flop to the output (as shown in the `bus_out` example) will minimize these effects.

Race conditions within the combinational logic can lead to glitches on output pins between clock edges. When this happens, pins may glitch from 0 to 1 and back to zero or vice versa as signals propagate through the combinational logic. Adding a flip-flop at the output removes these effects.

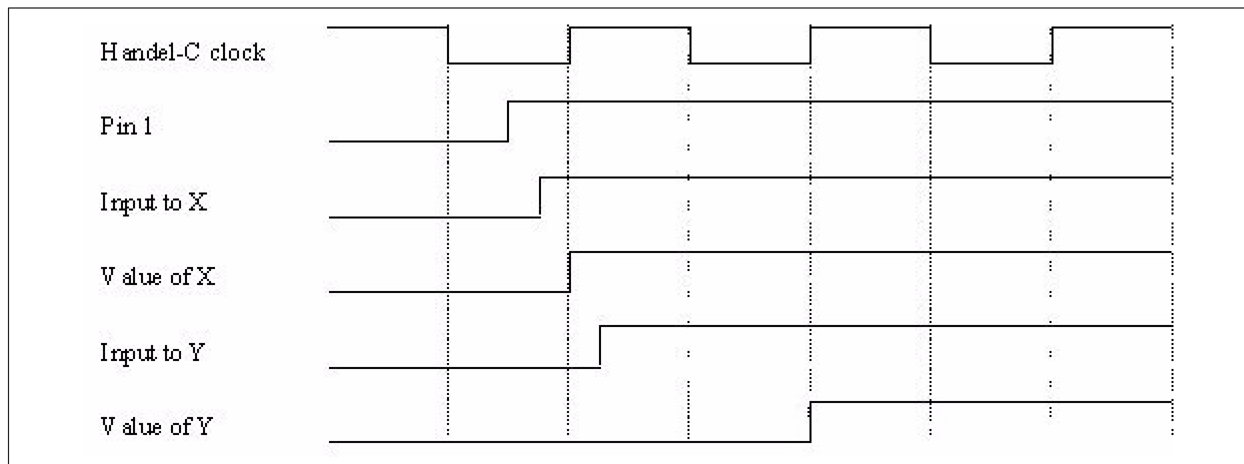
Bi-directional tri-state buses

The timing considerations for `bus_in` and `bus_out` interfaces should also be taken into account when using bi-directional tri-state buses since these are effectively a combination of an input bus and an output bus.

11.5.1 Example timing considerations for input buses

```
interface bus_in(int 1 read) a() with
   {data = {"P1"}};

par
{
    x = a.read;
    y = a.read;
}
```

Even though a.read is assigned to both x and y on the same clock cycle, if the delay from pin 1 to the flip-flop implementing the x variable is significantly different from that between pin 1 and the flip-flop implementing the y variable then x and y may end up with different values.



The delay between pin 1 and the input of y is slightly longer than the delay between pin 1 and the input to x. As a result, when the rising edge of the clock registers the values of x and y, there is one clock cycle when x and y have different values.

This effect can also occur in places that are more obscure.

```
interface bus_in(int 1 read) a() with
        {data = {"P1"}};

while (a.read==1)
{
    x = x + 1;
}
```

Although a.read is only apparently used once, the implementation of a while loop requires the signal to be routed to two different locations giving the same problem as before. The solution to this problem is to use either a bus_latch_in or a bus_clock_in interface sort.

The compiler will detect any occurrences of a pin feeding more than one register, and issue a warning.

11.5.2 Example timing considerations for output buses

```
int 8 x;
int 8 y;

interface bus_out() output(int out = x * y)
    with {data = {"P7", "P6", "P5", "P4",
        "P3", "P2", "P1", "P0"}};
```

A multiplier contains deep logic so some of the 8 pins may change before others leading to intermediate values. It is possible to minimize this effect (although not eliminate it completely) by adding a variable before the output. This effectively adds a flip-flop to the output. The above example then becomes:

```
int 8 x;
int 8 y;
int 8 z;

interface bus_out() output(int out = z)
    with {data = {"P7", "P6", "P5", "P4",
        "P3", "P2", "P1", "P0"}};

z = x * y;
```

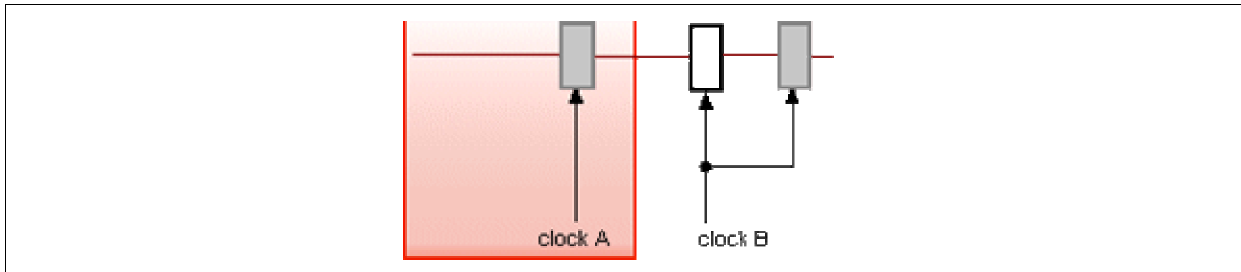
You must now take care to update the value of `z` whenever the value output on the bus must change.

11.6 Metastability

If the input of a flip-flop is connected to a signal which is not synchronous with the flip-flop's clock then its setup or hold time may be violated. This can result in the flip-flop entering a metastable state when it is clocked. The output of the flip-flop will then have an unpredictable value for an indeterminate period of time but will eventually become either 0 or 1.

In some circumstances (such as when two independent clocks are involved) metastability cannot be avoided. While a metastable flip-flop may remain so for any length of time, there is a high probability that it will enter a stable state after a relatively short delay.

The metastability characteristics of digital logic devices vary enormously. Refer to product data sheets for details.



The diagram shows flip-flops in separate clock domains. The central flip-flop receives data from the other clock domain. Its value is copied to the second flip-flop after 1 clock tick.

In that clock tick, it must resolve metastability and pass through any routing and output and setup delays.

11.6.1 Techniques to minimize the problem

- use extra registers to stabilize the data
- decouple the FPGA/PLD from external synchronous hardware by using external buffer storage

Stabilizing the data

The ideal system is designed such that when data is clocked into a register it is guaranteed to be stable.

The solution is to clock the data into the Handel-C program more than once, so it is clocked into one register, and the output of that register is then clocked into another register. On the first clock edge the data could be changing state so the output could be metastable for a short time after the clock. However, as long as the clock period is long relative to the possible metastable period, the second clock edge will clock stable data. Even more clock edges further reduce the possibility of metastable states entering the program, however the move from one clock to two clock ticks is the most significant and should be adequate for most systems.

The example below has 4 clock edges. The first is in the `bus_clock_in` procedure, and the next 3 are in the assignments to the variables `x`, `y`, and `z`.

```
int 4 x,y,z;

interface bus_clock_in(int 4 read) InBus() with
            {data = {"P4", "P3", "P2", "P1"}};

par
{
    while(1)
        x = InBus.read;

    while(1)
        y = x;

    {
        ......
        z = y;
    }
}
```

Designing the system to minimize the problem

Remember to keep the problem in perspective by examining the details of the board to estimate the probability of metastability. You can use external buffers to stabilize data from synchronous hardware before it is input to the FPGA.

Techniques to minimize the problem

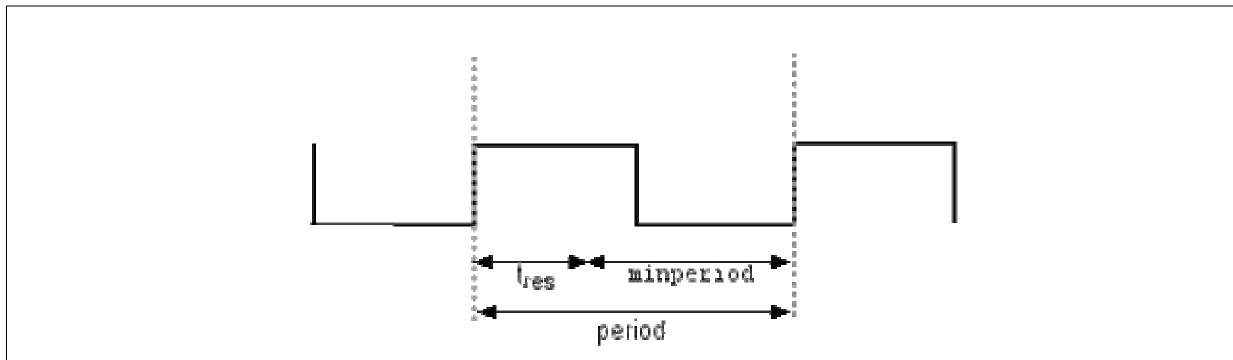
If using channels to communicate between clock domains, you may use clock specifications to balance speed and metastability issues

If using interfaces to communicate between clock domains, you can insert extra stabilizing flip-flops to reduce the likelihood of metastability being propagated through the circuit

Timing constraints used for channels across clock domains

Within a single clock tick, data transmitted from another asynchronous clock domain must settle (stop being metastable) and be routed to the next flip flop.

If you are using channels to communicate between clock domains, you can set the timing constraints which specify how long it is before you sample data (the amount of time for it to settle) OR the amount of time available for it to move onwards.

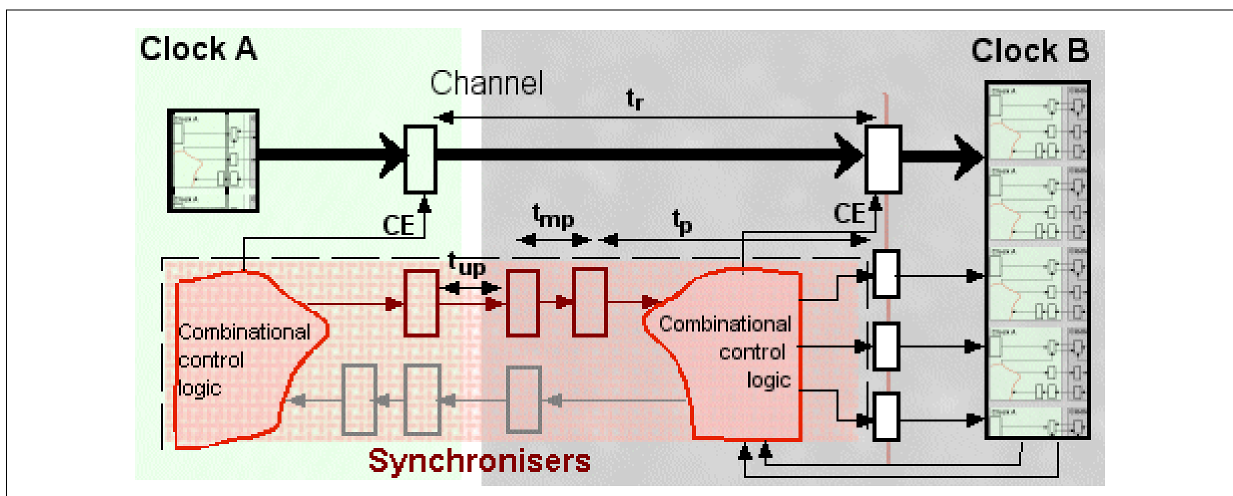


The amount of time used for it to settle is known as the resolution time (t_{res}). You can specify a maximum period for this by using the `resolutiontime` specification. A sensible value for `resolutiontime` is three-quarters of the clock period.

The amount of time left is the amount of time for the control signal to get from one flip-flop to the next, including all output, setup and routing delays. This is the `minperiod` specification. This would normally only be used if `paranoia` is set to 0.

How channels are designed to deal with metastability

When you use a channel to communicate across clock domains, synchronization hardware is built automatically.



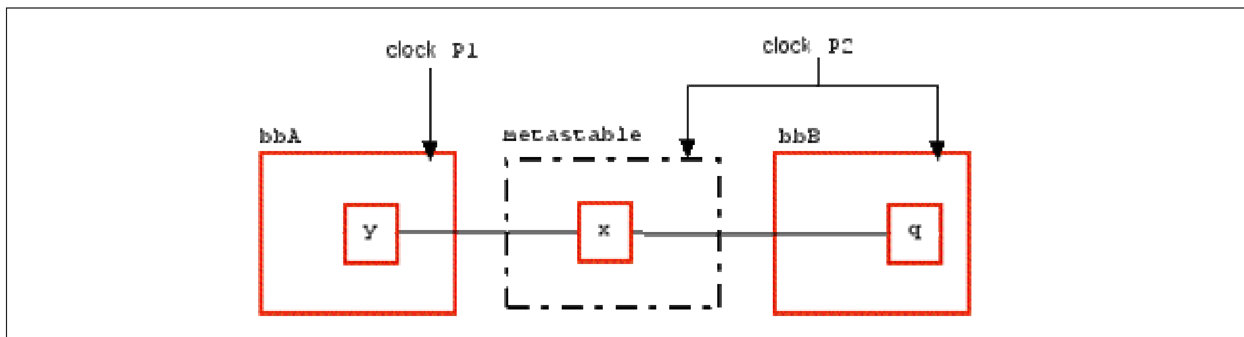
t_r	time to transfer between domains (`paranoia` in domain B + 1) X t_p
t_p	clock period in domain B
t_{up}	`unconstrainedperiod`

t_{mp} minperiod

The control signals are clocked through a number of flip-flops specified by paranoia. On each clock edge, the data is moved through another flip-flop, such that it is less likely to be metastable.

11.6.2 Using interfaces: External resynchronizing example

This example shows the three files required to connect two EDIF blocks (bbA and bbB) which use different clocks. The small files bbA.hcc and bbB.hcc compile to the EDIF code using the port_out from and port_in to interfaces. The metastable.hcc file connects the two together and generates one flip –flop that resynchronizes the data by reading the value from bbA into a variable.



File: metastable.hcc

```
/*
* Black box code to resynchronize
* Needs to be clocked from the reading clock
* (i.e. bbB.hcc's clock)
*/

int 1 x;
interface bbA(int 1 from) A();
interface bbB() B(int 1 to=x, unsigned 1 clk = __clock);

set clock = external "P1";
void main(void)
{
    while(1)
    {
        /*
         * stabilize the data by adding resynchronization FF
         */
        x = A.from;
    }
}
```

File: bbA.hcc

```
/*
* Domain bbA
* Compiles to bbA.edf
*/
interface port_in(unsigned 1 clk with { clockport = 1 }) clk();
set clock = internal clk.clk;
void main(void)
{
    int 1 y;
    interface port_out() from (int 1 from = y);
}
```

File: bbB.hcc

```
/*
*Domain bbB
* Compiles to bbB.edf
*/

set clock = external "P2";
void main(void)
{
    int 1 q;

    interface port_in(int 1 to) to();
    par
    {
        while(1)
        {
            q = to.to; // Read data
        }
    }
}
```

11.7 Ports: interfacing with external logic

Handel-C provides the interface sorts `port_in` and `port_out`. These allow you to have a set of wires, unconnected to pins, which you can use to connect to a simulated device or to another function within the FPGA or PLD. Handel-C supplies the interface declaration for these sorts, and you supply the instance definition.

port_in

For a `port_in`, you define the port(s) carrying data to the Handel-C code and any associated specifications.

```
interface port_in(Type data_TO_hc [with {port_specs}])
    Name() [with {Instance_specs}];
```

For example:

```
interface port_in(int 4 signals_to_HC) read();
```

You can then read the input data from the variable *Name.data_TO_hc*, in this case `read.signals_to_HC`

port_out

For a port_out, you define the port(s) carrying data from the Handel-C code, the expression to be output over those ports, and any associated specifications.

```
interface port_out() Name(Type data_FROM_hc =
    output_Expr[with {port_specs}])
        [with {Instance_specs}];
```

For example:

```
int X_out;
interface port_out()
   drive(int 4 signals_from_HC = X_out);
```

In this case, the width of X_out would be inferred to be 4, as that is the width of the port that the data is sent to.

Port names

The name of each port in a port_in or port_out interface must be different, as they will all be built to the top level of the design.

The examples below would both generate a compiler error.

Example 1:

```
interface port_in(unsigned 1 soggy) In1();
interface port_in(unsigned 1 soggy) In2();
```

Example 2:

```
interface port_in(unsigned 1 soggy) In1();
void main(void)
{
    interface port_in(unsigned 1 soggy) In2();
    ...
}
```

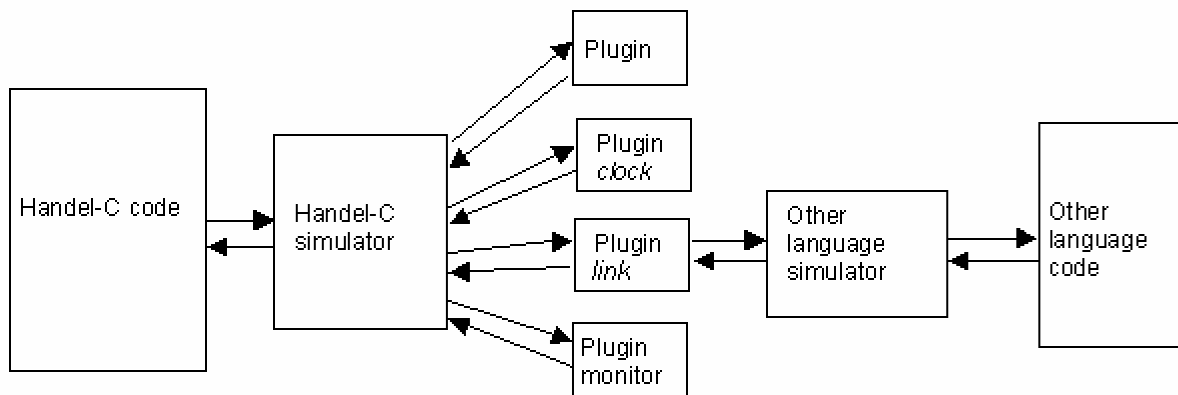
Both examples build two ports to the top level of the design called soggy. When they were integrated with external code, the PAR tools wouldn't know which soggy to use where.

11.8 Specifying the interface

You can specify your own interface format. This allows you to communicate with code written in another language such as VHDL, Verilog or EDIF and allows the Handel-C simulator to communicate with an external plugin program (e.g., a connection to a VHDL simulator).

The expected use for this is to allow you to incorporate bought-in or handcrafted pieces of low-level code in your high-level Handel-C program. It also allows your Handel-C program code to be incorporated within a large EDIF, VHDL or Verilog program. You can also use it to communicate with programs running on a PC that simulate external devices.

To use such a piece of code requires that you include an `interface` definition in the Handel-C code to connect it to the external code block. This interface definition also tells the simulator to call a plugin (which in turn may invoke a simulator for the foreign code).



11.9 Targeting ports to specific tools

When compiling to EDIF, Handel-C has the capacity to format the names of wires to external logic according to the different syntaxes used by any external components generated by foreign tools. You can do this using the `busformat` specification to a port. This allows you to specify how the bus name and wire number are formatted.

To specify a format, you use the syntax

`with {busformat = "`*`formatString`*`"}`

`formatstring` can be one of the following strings. `B` represents the bus name, and `I` represents the wire number.

```
BI
B_I
B[I]
B(I)
B<I>
```

`B` specifies a bus

`B[N:0]`, `B<N:0>` or `B(N:0)` specify a bus of width `(N+1)`.

Example format B[I]

```
interface port_in(int 4 signals_to_HC with
        {busformat="B[I]"}) read();
```

would produce wires

```
signals_to_HC[0]
signals_to_HC[1]
signals_to_HC[2]
signals_to_HC[3]
```

Example format B<I>

```
ram unsigned 4 rax[4] with
        {ports = 1, busformat="B<I>"};
```

would produce wires

```
rax_SPPort_addr<0>        // Address
rax_SPPort_addr<1>
rax_SPPort_data_in<0>     // Data In
rax_SPPort_data_in<1>
rax_SPPort_data_in<2>
rax_SPPort_data_in<3>
rax_SPPort_data_out<0>    // Data Out
rax_SPPort_data_out<1>
rax_SPPort_data_out<2>
rax_SPPort_data_out<3>
rax_SPPort_data_en        // Data Enable
rax_SPPort_clk            // Clock
rax_SPPort_cs             // Chip Select
rax_SPPort_oe             // Output Enable
rax_SPPort_we             // Data In
```

12 Object specifications

Handel-C provides the ability to add 'tags' to certain objects (variables, channels, ports, buses, RAMs, ROMs, mprams, clocks, resets and signals) to control their behaviour. These tags or specifications are listed after the definition of the object using the `with` keyword. All specifications can be applied to generic output. Others are only valid for simulation (Debug or Release) or for hardware output.

When defining multiple objects, the specification must be given at the end of the line and it applies to all objects defined on that line. For example:

```
extern unsigned x, y;
unsigned x, y with {show=0};
```

This attaches the `show` specification with a value of 0 to both `x` and `y` variables.

Specifications can only be applied to the definition of objects, not to declarations:

```
extern rom unsigned 32 SomeRom[1] with {Spec}; // Wrong; spec applied to de
claration
rom unsigned 32 SomeRom[1]={1} with {Spec}; // OK; spec applied to definiti
on
```

The `with` keyword takes one or more of the following attributes.

12.1 Summary of specifications

12.1.1 Compiler atttributes

These specifications are interpreted by the compiler.

Specification	Possible values	Default	Applies to	Meaning
`warn`	0, 1	1	variables memories channels interfaces clocks	Enable warnings for object
`extpath`	Name of port TO Handel-C on the same `interface`	None	port FROM Handel-C	Specify any direct logic (combinational logic) connections to another port

12.1.2 Simulator attributes

These specifications are interpreted by the simulator.

Specification	Possible values	Default	Applies to	Meaning
`show`	0, 1	1	variables channels o/p interfaces tri-state interfaces	Show variable during simulation
`base`	2, 8, 10, 16	10	variables chanouts o/p interfaces tri-state interfaces	Print variable in specified base
`infile`	Any valid filename	None	chanins i/p interfaces tri-state interfaces	Redirect from file
`outfile`	Any valid file name	None	chanouts o/p interfaces tri-state interfaces, variables	Redirect to file
`extlib`	Name of a plugin `.dll`	None	interface or port	Specify external plugin for simulator
`extfunc`	Name of a function within the plugin	`PlugInSet` or `PlugInGet` depending on port direction	interface or port	Specify external function within the simulator for this port
`extinst`	Instance name (with optional parameters)	None	interface or port	Specify simulation instance used

12.1.3 Clock attributes

These specifications apply to a clock, and affect the hardware built in that clock domain.

Specification	Possible Values	Default	Applies to	Meaning
`clockport`	0, 1	0 for a port on an interface, 1 for a clock declaration	ports on interfaces, external clocks	Mark port as feeding a clock. When applied to a generic interface port, it marks that port as feeding a clock. When applied to an external clock, it marks that clock as using a dedicated clock pin.
`minperiod`	Any time in nanoseconds	None	clocks with channels to other clock domains	minimum period for place and route tools to achieve between flip-flops
`paranoia`	0 or any positive integer (above 10 causes a warning)	1	clocks	specifies number of extra flip-flops used in stabilizing synchronization data
`rate`	Any floating-point frequency in MHz	None	clocks	Minimum frequency at which the clock in question should be capable of running
`resolutiontime`	Any time in nanoseconds	None	clocks with channels to other clock domains	Time for metastability to resolve on channels into clock domains
`retime`	0 or 1	1	clocks, variables	Prevent flip-flops in a specific clock domain or generated by a variable from being moved by the retimer
`unconstrainedperiod`	Any time in nanoseconds	None	clocks with channels to other clock domains	Constraint for compiler-generated control paths into clock domain

12.1.4 Channel attributes

This specification defines how channels are built.

Specification	Possible values	Default	Applies to	Meaning
`fifolength`	0 or any positive integer	0	channel	Create FIFO of specified length

12.1.5 Channel and memory attributes

This specification defines where memories and FIFOs are built.

Specification	Possible values	Default	Applies to	Meaning
`block`	"AUTO" for any device; "BlockRAM" for Actel; "LUT", "EAB", "M512", "M4K" or "M-RAM" for Altera; "BlockRAM" or "SelectRAM" for Xilinx	"AUTO"	memories (on-chip) FIFOs of two or more parts	Specify memory resource type to use for RAM/ROM

12.1.6 Memory attributes

These specification defines how memories are built.

Specification	Possible Values	Default	Applies to	Meaning
`offchip`	0, 1	0	memories	Set RAM/ROM to be off chip. Cannot be used in conjunction with ports
`ports`	0, 1	0	memories	Set RAM/ROM to be in external code. Cannot be used in conjunction with offchip
`wegate`	-1, 0, 1	0	RAMs	Place write enable signal
`westart`	in multiples of 0.5 to (clock division -0.5)	None	RAMs	Position write enable signal
`welength`	in multiples of 0.5 to clock division	None	RAMs	Set length of write enable signal
`rclkpos`	in multiples of 0.5 to (clock division -0.5)	None	memories	Set read cycle position of SSRAM clock
`wclkpos`	in multiples of 0.5 to (clock division -0.5)	None	memories	Set write cycle position of SSRAM clock
`clkpulselen`	in multiples of 0.5 to clock division	None	memories	Set pulse length of SSRAM clock
`clk`	Any valid pin list	None	memories (off-chip)	Set pins for external RAM or ROM clock
`addr`	Any valid pin list	None	memories (off-chip)	Set address pins
`oe`	Any valid pin list	None	memories (off-chip)	Set output enable pin(s)
`we`	Any valid pin list	None	RAMs (off-chip)	Set write enable pin(s)
`cs`	Any valid pin list	None	memories (off-chip)	Set chip select pin(s)

12.1.7 Interface and memory attributes

This specification defines how interfaces and memory connections are built.

Specification	Possible Values	Default	Applies to	Meaning
`speed`	0, 1, 2 (Actel ProASIC only) 0, 1 (Altera and Xilinx)	2 for Actel ProASIC and ProASIC+ 1 for Altera and Xilinx Virtex, Spartan-II/IIE/3/3E /3L series	o/p or tri-state interfaces	Set buffer speed
`intime`	Any floating-point delay (ns)	None	input port or interfaces or tri-state interfaces external RAMs	Maximum allowable delay between interface and variable
`outtime`	Any floating-point delay (ns)	None	output port or interfaces or tri-state interfaces external RAMs	Maximum allowable delay between variable and interface
`standard`	Specified keywords representing I/O standards	LVCMOS33 for ProASIC / ProASIC+ LVTTL for other devices	any external interface or external clock (dependent on FPGA type), and off-chip memories	I/O standard used (electrical characteristics)
`strength`	2, 4, 6, 8, 12, 16, 24 (mA) OR 0 (Min), -1 (Max)	Various, refer to device datasheets	external interfaces and off-chip memories	Signal strength.

`dci`	0, 0.5, 1	0 (No DCI)	external interfaces and external clocks (Virtex-II, Virtex-II Pro and Spartan-3/3E/3L only) and off-chip memories	Digital control impedance enabled (only valid with some standards)
`busformat`	Format string	BI	generic interfaces, port-type interfaces and ports to memories in external logic	Specify the way that wire names are formatted in EDIF
`pull`	0, 1	None	Xilinx and ApexII interfaces	Add pull up or pull down resistor(s)
`data`	Any valid pin list	None	memories interfaces	Set data pins

12.1.8 Interface attributes

These specifications defines how interfaces are built.

Specification	Possible values	Default	Applies to	Meaning
bind	0,1	0	interface, port	Bind component to work library
buffer	string value	Depends on target architecture and type of interface	bus-type interfaces, external clocks & resets	In EDIF: specify type of buffer to build
properties	string-value pair OR string-value-string triplet	None	generic interfaces	In EDIF: Parameterize instantiations of external black boxes In VHDL: Define generics In Verilog: Define parameters
quartus_proj_assign	string-value pair	None	bus-type interfaces, offchip RAM	In EDIF: specify Quartus project pins assignments
sc_type	string-value	bool for 1 bit wide ports, uint otherwise	port_in, port_out or generic interfaces	Create a SystemC port of a specified type
vhdl_type	string-value	std_logic for 1 bit wide ports, unsigned otherwise	port_in, port_out or generic interfaces	Create a VHDL port of a specified type

12.1.9 Examples

Specifications can be added to objects as follows:

```
unsigned 4 w with {show=0};
int 5 x with {show=0, base=2};
chanout char y with {outfile="output.dat"};
chanin int 8 z with {infile="input.dat"};
interface bus_clock_in(int 4 in) InBus() with
    { pull = 1, data = {"P4", "P3", "P2", "P1"} };
```

12.2 base specification

The `base` specification may be given to variable, output channel, output bus and tri-state bus declarations. You can only use it for simulation output (Debug or Release). The value that this specification is set to tell the Handel-C compiler which base to display the value of the object in. Valid bases are 2, 8, 10 and 16 for binary, octal, decimal and hexadecimal respectively.

The default value of this specification is 10. If you write `with {base = 0}` this is equivalent to not specifying a base.

Example

```
int 5 x with {base=2};
```

12.3 bind specification

The `bind` specification may be given to a user-defined interface that connects to a component in external logic. It only has meaning when instantiating an external block of code from Handel-C generated VHDL or Verilog. If `bind` is set to 1, it is assumed that the definition of the component exists in HDL elsewhere. If it is set to 0, it does not and the component is assumed to be a black box.

In VHDL, setting `bind` to 1 instantiates the component and generates a declaration of this component of which the definition is assumed to be within the work library. Setting `bind` to 0 (default) instantiates the component and generates a black box component declaration.

In Verilog, setting `bind` to 1 instantiates the component but does not declare it. Setting `bind` to 0 instantiates the component and generates a black box component declaration. This black box component declaration is an empty module, which merely describes the interfaces of the component.

VHDL example 1: with `bind` set to 0:

```
interface Bloo(unsigned 1 myin)
    B(unsigned 1 myout = x) with {bind = 0};
```

results in Handel-C generating this VHDL instantiation of the `Bloo` component:

```
component Bloo
port (
        myin : out std_logic;
        myout : in std_logic
      );
end component;
```

VHDL example 2: with `bind` set to 1:

```
interface Bloo(unsigned 1 myin)
    B(unsigned 1 myout = x) with {bind = 1};
```

results in Handel-C generating this VHDL instantiation/declaration of the `Bloo` component:

```
component Bloo
port (
        myin : out std_logic;
        myout : in std_logic
     );
end component;
for all : Bloo use entity work.Bloo;
```

In this case `Bloo` is bound to the work library.

Verilog example 1: with `bind` set to 0:

```
interface Bloo(unsigned 1 myin)
    B(unsigned 1 myout = x) with {bind = 0};
```

results in Handel-C generating this Verilog instantiation of the `Bloo` component:

```
module Bloo;
    input myin;
    output myout;
endmodule;

module MyModule;
    ...
    wire a, b;
    ...
    Bloo MyInstance (.myin(a), .myout(b));
    ...
endmodule;
```

Note that the code includes a black box declaration of `Bloo`.

Verilog example 2: with `bind` set to 1:

```
interface Bloo(unsigned 1 myin) B(unsigned 1 myout = x) with {bind = 1};
```

results in Handel-C generating this Verilog instantiation of the `Bloo` component:

```
module MyModule;
    ...
    wire a, b;
    ...
    Bloo MyInstance (.myin(a), .myout(b));
    ...
endmodule;
```

(The VHDL or Verilog synthesizer expects the declaration of `Bloo` to be provided in another block of HDL.)

12.4 block specification

The `block` specification may be given to a RAM or ROM declaration, for EDIF, VHDL or Verilog output. The `block` specification may also given to channels where `fifolength` is 2 or greater.

The specification takes a string to specify the type of block memory required. Possible values are:

- Actel devices: `"BlockRAM"`
- Altera devices: `"LUT"`, `"EAB"`, `"M512"`, `"M4K"`, `"M-RAM"` (`"EAB"` should be used for both EABs and ESBs)
- Xilinx devices: `"SelectRAM"`, `"BlockRAM"`
- All devices: `"AUTO"`. This is the same as not using the block specification, but can be used as a placeholder to pass in an active value.

For example:

```
ram int 8 a[15][43] with {block = "BlockRAM"};
```

```
chan <unsigned 1> ch with {fifolength=15, block ="SelectRAM"};
```

If you want to build a ROM from look-up tables (distributed memory) in Altera devices, you need to declare the ROM with `{block = "LUT"}`.

`"M512"`, `"M4K"` and `"M-RAM"` are used to specify memory blocks in Stratix and Cyclone devices.

The `block` specification has changed since DK1.1, although the old method, using `block = 1` to specify block RAMs, is still supported for backward compatibility.

Issues with Xilinx Virtex, VirtexE and Spartan-IIE

Due to the pipelined nature of Virtex and Spartan-IIE block RAM, if you attempt to read from one bank of block RAM and write the value into another on a single cycle, the value read is the value in block RAM on the previous clock cycle, not the current cycle.

Code example with timing issues

```
static ram unsigned 8 RAM1[4] = {0,1,2,3} with {block="BlockRAM"};
ram unsigned 8 RAM2[4] with {block="BlockRAM"};
signal s;
unsigned x;
unsigned i;

while(1)
{
    par
    {
        s = RAM1[i];
        RAM2[i] = s;
        x = s;
        i++;
    }
}
```

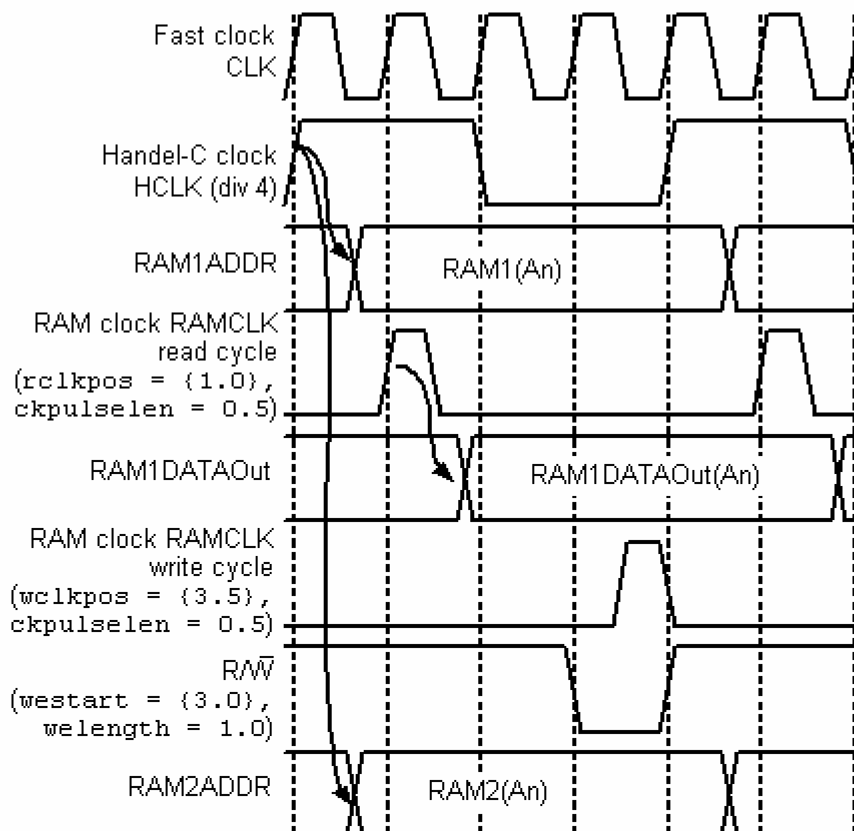
Here, x and RAM2[i] get different values. s changes on the falling edge. x is written to on the rising edge. RAM2[i] is written to on the falling edge.

Therefore, RAM2[i] gets the value of RAM1[i-1] and x gets the value of RAM1[i].

To alter this, you must use the rclkpos, wclkpos and clkpulselen specifications to set the RAM clock cycle positions.

Solution to timing problem

```
//divide CLK by four to give Handel-C clock
set clock = external_divide "C1" 4;

static ram unsigned 8 RAM1[4] = {0,1,2,3} with {block = "BlockRAM",
    rclkpos = {1.0},
    wclkpos = {3.5},
    clkpulselen = 0.5,
    westart = 3.0,
    welength = 1.0};

ram unsigned 8 RAM2[4] with {block = "BlockRAM",
    rclkpos = {1.0},
    wclkpos = {3.5},
    clkpulselen = 0.5,
    westart = 3.0,
    welength = 1.0};
```



HCLK initiates the parallel read from and write to the different blocks of RAM.

The settings of `rclkpos` and `clkpulselen` delay the read cycle until the address is stable. (Read clock pulse 1 CLK pulse after HCLK, held for 0.5 CLK pulses).

The settings of `wclkpos` and `clkpulselen` delays the write cycle until after the data has been read and is stable. The settings of `westart` and `welength` position the write enable appropriately.

12.5 buffer specification

The `buffer` specification can be applied to all bus-type interfaces and external clock/reset declarations. It accepts a string, and it specifies the type of buffer that should be built on the corresponding interface. "None" may be used to specify that no buffer should be built.

Example 1:

```
interface bus_in(unsigned 3 i) I() with {buffer = "IBUFG"};
```

builds a standard `bus_in` interface, where the buffer is of type `IBUFG`, specifying that the `bus_in` should feed a global buffer (for Xilinx) instead of a basic input buffer (for connecting to DCMs, for instance).

Example 2:

```
interface bus_in(unsigned 3 i) I() with {buffer = "None"};
```

builds a standard `bus_in` interface with no buffer. That is, any logic reading from `I.i` will be fed by pins directly.

Example 3:

```
set clock = external with {buffer = "GL25LP"};
```

specifies that the clock should use a low-power clock buffer for Actel.

Where no `buffer` spec is used, the default buffer type is used.

12.6 busformat specification

The `busformat` specification may be given to

- generic and port-type (`port_in` and `port_out`) interfaces (but not bus-type interfaces)
- port memories (memories using `with {ports = 1}` to connect to external code)

`busformat` specifications are ignored for VHDL and Verilog output and for bus-type interfaces (`bus_in`, `bus_ts` etc).

When compiled to EDIF, the `busformat` string defines the format of the wire names. Valid values for the `busformat` string are

```
BI B_I   B[I]   B(I)    B<I>
```

`B` represents the bus name and `I` the wire number. The default format is `BI`

If you want to specify a single port for the entire bus, use

```
B          B[N:0]  B<N:0>  B(N:0)
```

`B` specifies a bus without specifying a width and `B[N:0]` and `B<N:0>` specify a bus of width (N +1). A 6-bit port could therefore be generated as `port`, `port[5:0]`or `port<5:0>` depending on the value of `busformat`.

If `data` specifications are used with `busformat`, they are ignored and a warning is issued.

You can place the `busformat` specification after any port, or at the end of an interface statement. If you place a specification at the end of the interface declaration, it will apply to all ports in the declaration, except for any ports that have their own specification. For example:

```
interface Bloo (unsigned 4 in)
    InstBloo (unsigned 4 out = x
    with {busformat = "BI"})
    with {busformat = "B(I)"};
    // first port has spec B(I) and second port has spec BI
```

If you want to apply a `busformat` specification to a 1-bit wide bus, you need to place the specification after the port. If the specification is applied to the whole interface, it will be ignored for any 1-bit wide buses in the interface (to enable these to be used as signals etc.).

Examples

```
interface port_in(int 4 signals_to_HC with {busformat="B[I]"}) read();
```

creates four ports named `signals_to_HC[0]`, `signals_to_HC[1]`, `signals_to_HC[2]` and `signals_to_HC[3]`.

```
interface port_in(unsigned 6 myvar) MyFunction()
    with {busformat = "B[N:0]"};
```

creates a single 6-bit port: `myvar[5:0]`.

```
unsigned 6 x;
interface ExtThing(unsigned 6 myvar)
    Inst1ExtThing(unsigned 6 anothervar = x)
    with {busformat = "B[N:0]"};
```

creates two ports: `myvar[5:0]` and `anothervar[5:0]`.

```
interface ExtThing(unsigned 5 a,
    unsigned 1 b with {busformat = "B[I]"},
    unsigned 1 c)
    InstExtThing(unsigned 6 d)
    with {busformat = "B[I]"};
```

In this example, the `busformat` specification is applied to ports `a` and `d`, because they are more than 1-bit wide, and to port `b`, as this has an individual `busformat` specification, but not to port `c` as this is 1-bit wide and does not have an individual `busformat` specification.

12.7 Specifying the clock pin for SSRAM

The `clk` specification is used for external SSRAM or ROM declarations, for EDIF, VHDL or Verilog output. It specifies the pin(s) that carry the RAM/ROM clock to the external SSRAM/ROM. To use this specification, you must be using the `external_divide` or `internal_divide` clock types with a division factor of 2 or more, and you must use the `wclkpos`, `rclkpos` and `clkpulselen` specifications to define the clock that will appear at the specified pin(s).

Example

```
set clock = external_divide "C1" 4;

ram unsigned 4 ExtSyncMem[32] with
{
    offchip = 1,
    wclkpos = {2.5},
    rclkpos = {2.5},
    clkpulselen = 1,
    clk = {"P22"},
    westart = 2,
    welength = 1,
    we = {"P23"},
    cs = {"P24"},
    oe = {"P25"}
};

void main(void)
{
    static unsigned index;
    static unsigned data;

    ExtSyncMem[index] = data;
    etc...

    data = ExtSyncMem[index];
    etc...
}
```

The clock pattern defined by the `wclkpos`, `rclkpos` and `clkpulselen` specifications appears at pin "`P22`". The write enable strobe defined by `westart` and `welength` appears at pin "`P23`".

12.8 clockport specification

The `clockport` specification can be used when declaring a port on an interface, or when declaring a clock. You can use it for EDIF, VHDL or Verilog output.

Port declaration

You can use the `clockport` specification to indicate that a port on an interface is used to drive a clock in the Handel-C design. This is useful when the clock for the Handel-C design originates in an external 'black box' component. For example

```
unsigned 1 En;
interface BlackBox(unsigned 1 CLK with {clockport=1})
    Instance(unsigned 1 Enable = En);

set clock = internal Instance.CLK;
```

If you don't use the clockport specification you may end up with combinational loops.

Clock declaration

You can use the clockport specification, with {clockport=1}, when declaring external clocks to assign the clock to a dedicated clock input resource on the target device.

If you apply the clockport specification to Xilinx Virtex parts, you can use it to specify a particular "input" clock buffer.

If clockport is set to 0, the clock is assigned to a pin that is not a dedicated clock input and the I/O standard and dci specifications are not available.

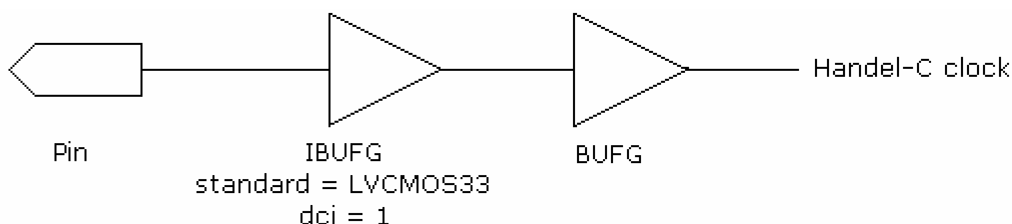
Example clock declarations

```
set family = XilinxVirtexII;
set clock = external with {standard = "LVCMOS33", dci = 1};
```

OR

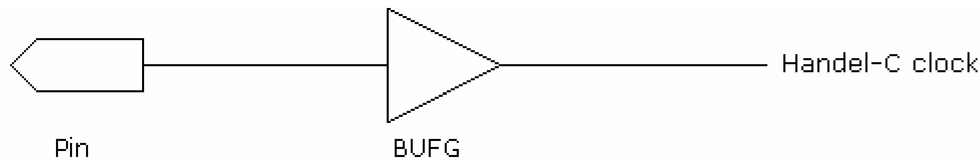
```
set family = XilinxVirtexII;
set clock = external with {clockport = 1, standard = "LVCMOS33", dci = 1};
```

both instruct the compiler to build an external clock interface, using a dedicated Virtex-II clock input (IBUFG) resource. That is, the clock interface logic built will be:




```
set family = XilinxVirtexII;
set clock = external with {clockport = 0, standard = "LVCMOS33", dci = 1};
```

This instructs the compiler to build an external clock interface, without using a dedicated Virtex-II clock input resource. That is, the clock interface logic built will be:



12.9 data specification (pin constraints)

The `data` specification can be used to constrain pin location or to name ports:

- When applied to bus-type interfaces or off-chip memories, `data` specifies pin locations as a list of pin numbers separated by commas. If you are using a differential I/O standard, the pins must be specified as pairs enclosed in braces.
- When applied to foreign code memories (using `with {ports=1}`), port-type interfaces and generic interfaces, `data` specifies port names as a list of names separated by commas

If the `data` specification is omitted for bus-type interfaces or off-chip memories, the place and route tools will assign the pins. The pins are listed in order MSB to LSB, but the LSB pin (rightmost element of list) is assigned first. If you do not assign all the pins used, the MSB pins remain unassigned.

If you are targeting EDIF output, the `data` specification can also be used for a `port_in` or `port_out` interface to specify the names of the ports to be exported. (This part of the `data` specification is ignored for VHDL or Verilog output.)

If you are compiling your Handel-C code to VHDL or Verilog, you can only use the `data` specification to constrain pin locations for Precision and Synplify style outputs. If you compile for ModelSim or Active-HDL, the `data` specification is ignored. In Precision VHDL or Verilog output styles, pin constraints are implemented using the `pin_number` attribute. In Synplify-style output, pin constraints are implemented using the `loc` attribute.

If the `busformat` specification is used as well as `data` specifications for port-type or generic interfaces, the `data` specifications are ignored and a warning is issued.

Bus-type interface example

```
macro expr dataPins = {"P3", "P2", "P1", "P0"};
interface bus_in(unsigned 4 inPort) hword()
    with {data = dataPins, intime = 5};
```

Port-type interface example

```
macro expr dataInNames = {"I3", "I2", "I1", "I0"};
macro expr dataOutNames = {"O3", "O2", "O2", "O1"};

unsigned 4 x;
interface port_in(unsigned 4 in) Ig()
    with {data = dataInNames};
interface port_out() Og(unsigned 4 out = x)
    with {data = dataOutNames};
```

Generic interface example

```
macro expr dataInNames = {"I3", "I2", "I1", "I0"};
macro expr dataOutNames = {"O3", "O2", "O2", "O1"};

unsigned 4 x;
interface Igator
    (
        unsigned 4 in with {data = dataInNames}
    )
    InstIgator
    (
        unsigned 4 out = x with {data = dataOutNames}
    );
```

12.10 dci specification

The dci specification may be used with the standard specification on external bus interfaces connected to pins (not port_in or port_out) to select whether Digital Controlled Impedance is to be used on all pins of that interface. You can also use it with the standard specification when declaring external clocks. The dci specification may also be applied to off-chip memories. The specification is only valid for EDIF, and is ignored for all other outputs.

The only devices that currently support DCI are Xilinx Virtex-II, Virtex-II Pro, Virtex-4 and Spartan-3/3E/3L. For more information on DCI, please refer to the Xilinx Data Book.

If you have used the clockport specification and set it to 0, dci specifications will be ignored. (The default for clockport is 1.)

Standards supporting dci are:

GTL	GTL+		
HSTL Class I	HSTL Class II	HSTL Class III	HSTL Class IV
LVCMOS33	LVCMOS25	LVCMOS18	LVCMOS15
SSTL2 Class I	SSTL2 Class II	SSTL3 Class I	SSTL3 Class II

The possible values for the dci specification are:

0	No DCI (default)
1	DCI with single termination
0.5	DCI with split termination. This can only be used with LVCMOS standards.

If dci is used on a device or standard that does not support it, a warning is issued and the specification is ignored.

Examples

```
// Use dci on all pins
interface bus_out() Eel(int 4 outPort = x)
   with {data = dataPins0, standard = "HSTL_I", dci=1};

//Use dci for clock pin
set clock = external "C1" with {standard = "HSTL_III", dci=1};
```

12.11 extinst, extlib, extfunc specifications

The extlib, extfunc and extinst specifications are used when connecting a Handel-C interface to a simulation .dll. There is a default value for extfunc, but extlib and extinst must both be specified.

Specification	Possible values	Default	Meaning
extlib	Name of a plugin .dll	None	Specify external plugin for simulator
extfunc	Name of a function within the plugin	PlugInSet or PlugInGet depending on port direction	Specify external function within the simulator for this port
extinst	Instance name (with optional parameters)	None	Specify simulation instance used

extlib

extlib takes the name of a .dll. It specifies that the named .dll plugin will be connected to the port or interface.

extfunc

extfunc specifies the name of an external function within the .dll.

On output ports, this function is called by the simulator to pass data from the Handel-C simulator to the plugin (default PlugInSet). It is guaranteed to be called every time the value on the port changes but may be called more often than that.

On input ports, this function is called by the simulator to get data from the plugin (default PlugInGet). It is guaranteed to be called at least once every clock cycle.

extinst

extinst takes a string, which is passed to the PlugInOpenInstance function within the plugin. If parameters must be passed to the .dll instance, they can be done so in the string. A new instance of the plugin will be generated for each unique extinst string.

Examples

```
interface bus_out() MyBusOut(outPort=MyOutExpr) with
    {extlib="pluginDemo.dll", extinst="0", extfunc="MyBusOut"};

interface TTL7446(unsigned 7 segments, unsigned 1 rbon)
    decode(unsigned 1 ltn=ltnVal, unsigned 1 rbin=rbinVal,
    unsigned 4 digit=digitVal, unsigned 1 bin=binVal)
    with {extlib="PluginModelSim.dll",
    extinst="decode; model=TTL7446_wrapper; delay=1"};
```

12.12 extpath specification

The `extpath` specification is used when connecting a Handel-C interface to external (black-box) logic. It is valid for any DK output.

`extpath` is used during simulation to tell the simulator about ports within the black box, so that it knows what order to update the ports in. It specifies that a Handel-C output port on an interface will have direct logic connections via the black box to one or more input ports on the same interface.

Its usage is

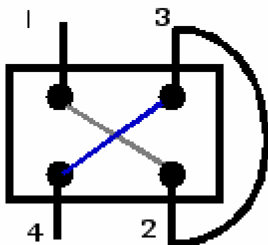
portName `with {extpath={`***portNameList***`}}`

portNameList is a comma-separated list of port names.

Example

```
interface blackBox
    (int 1 Two, int 1 Four)
    bb1(int 1 One = out with {extpath = {bb1.Two}},
    int 1 Three = bb1.Two with {extpath={bb1.Four}});
```

This example tells the compiler that there are direct connections within the black box between ports 1 and 2, and between ports 3 and 4. The interface also specifies an external connection from port 2 to port 3 (this connection is outside the black box).



12.13 fifolength specification

The `fifolength` specification converts a channel into a FIFO of the given length. If `fifolength` is two or greater the `block` specification can be used.

If `fifolength` is not a power of 2, and the `paranoia` specification is 0 or 1 (default), the FIFO will be created with low latency, else it will be created with a higher latency.

Example

```
int 8 chan_FIFO with {fifolength = 7, block = "MK4"}
//creates a FIFO in Altera Cyclone block RAM
```

12.14 infile and outfile specifications

The infile specification may be given to chanin, port_in, port_out, bus_in, bus_latch_in, bus_clock_in, bus_ts, bus_ts_latch_in and bus_ts_clock_in declarations. The outfile specification may be given to chanout, bus_out, bus_ts, bus_ts_latch_in and bus_ts_clock_in declarations. The strings that these specifications are set to will inform the simulator of the file that data should be read from (infile) or the file that data should be written to (outfile).

Note that when applying the outfile specification, it should not be given to multiple channels. For example, the following declarations are allowed, but it would be better to place them in separate files to avoid undefined results:

```
chanout int x, y with {outfile="out.dat"};
chanout unsigned a, b with {outfile="out.dat"};
```

The filename passed to infile and outfile is a standard string and follows all string rules, including the need to specify the backslash character as '\\'.

12.15 intime and outtime specifications

The intime specification may be given to an input port or bus, tri-state bus, foreign code memory or off-chip memory. The outtime specification may be given to an output port or bus, tri-state bus, foreign code memory or off-chip memory. The specifications are only valid for EDIF output.

intime specifies the maximum delay in ns allowed between an interface or memory interface and the sequential elements it feeds. outtime specifies the maximum delay in ns allowed between an interface or memory interface and the sequential elements it is fed from. They can be floating-point numbers. For example:

```
macro expr memoryPins = {"P13", "P12", "P11",
        "P10", "P9", "P8", "P7", "P6"};
macro expr dataPins = {"P4", "P3", "P2", "P1"};

interface bus_in(unsigned 4 dataIn) hword()
        with {data = dataPins, intime = 5};
interface port_out()
        new_hword(unsigned 4 out = hword.dataIn + 1)
        with {outtime = 5.2};
ram int 8 a[15][43] with {outtime = 5.2,
        offchip = 1,
        data = memoryPins};
```

When applied to Actel ProASIC devices, intime and outtime specifications cause Handel-C to generate a GCF file for the design. When an Altera device is the target, Handel-C generates ACF or TCL files. When applied to Xilinx chips, Handel-C generates a a Netlist

Constraints File (NCF). These files are used by the place-and-route tools to constrain the relevant paths.

12.16 Timing constraints example

This example shows the use of the `rate` specification and the `intime` and `outtime` specifications to constrain a design for speed. The use of these specifications causes the generation of a timing constraints file (with the type of file determined by the target platform).

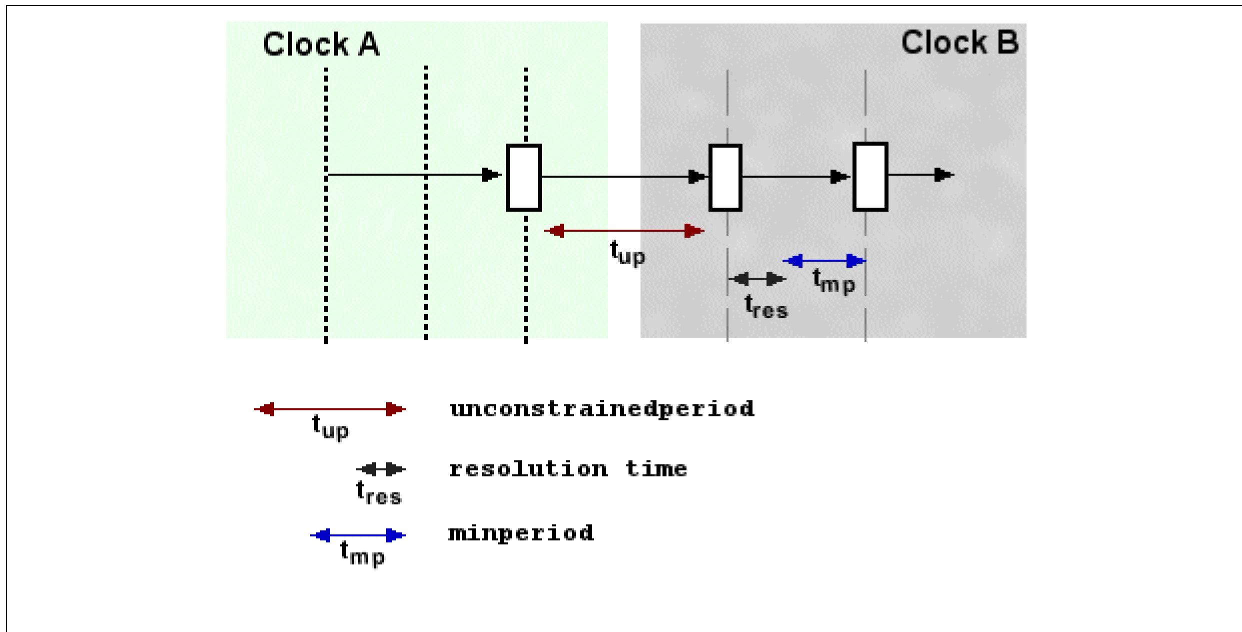
The design is constrained for a clock speed of 40MHz, with input data from two sources, taking a maximum of 5.5 and 5.0 nanoseconds, and output data taking a maximum of 4 nanoseconds to transmit.

```
// Clock
set clock = external "C13" with {rate = 40};

// Data path width
macro expr OpWidth = 8;

// Data pins
macro expr DataInA = {"D5","C5","E7","G8","H9","A5","A6","B5"};
macro expr DataInB = {"B6","D7","F8","E8","G9","F9","G10","H10"};
macro expr DataOut = {"B12","D12","D13","F13","G13","H13","H14","C14"};

// Data In/Out timing requirements
macro expr InTimeRequirementA = 5.5;
macro expr InTimeRequirementB = 5.0;
macro expr OutTimeRequirement = 4;

// Input data
interface bus_in(unsigned OpWidth dina) DINA() with
{
    data = DataInA,
    intime = InTimeRequirementA
};
interface bus_in(unsigned OpWidth dinb) DINB() with
{
    data = DataInB,
    intime = InTimeRequirementB
};

// Output data
unsigned result;
interface bus_out() DOUT(unsigned OpWidth dout = result) with
{
    data = DataOut,
    outtime = OutTimeRequirement
};

// Main program - pipelined multiplier
void main(void)
{
    unsigned xx[OpWidth];
    unsigned yy[OpWidth];
    unsigned rr[OpWidth];
```

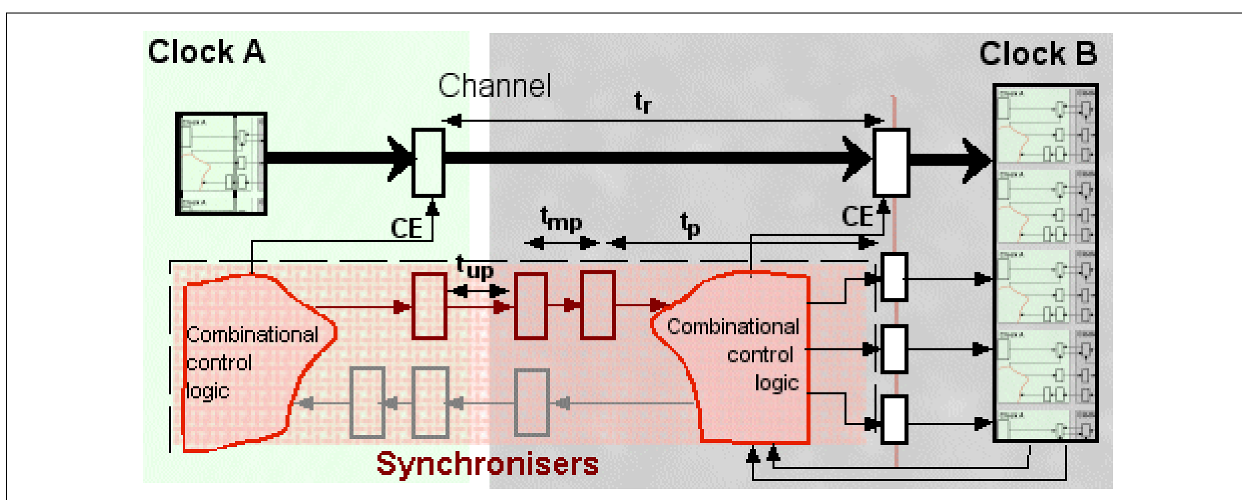
```
    while (1)
    {
        par
        {
            // Read operands from input interfaces
            xx[0] = DINA.dina;
            yy[0] = DINB.dinb;
            rr[0] = xx[0][0] ? yy[0] : 0;

            /*
             * Replicator: generates the pipeline stages of
             * the long multiplier, which are done in parallel.
             */
            par (Stage=1; Stage<OpWidth; Stage++)
            {
                xx[Stage] = xx[Stage-1] >> 1;
                yy[Stage] = yy[Stage-1] << 1;
                rr[Stage] = rr[Stage-1] + (xx[Stage][0] ? yy[Stage] : 0);
            }

            // Update result
            result = rr[OpWidth-1];
        }
    }
}
```

12.17 minperiod specification

The `minperiod` specification specifies the maximum delay in nanoseconds between flip-flops in a synchronizer. including output delay, setup time and skew at either end). Its value must be less than the clock period.



The higher the value for `minperiod`, the less time will be available within a clock tick for control signals to stabilize (`resolutiontime`). You may set the value of `minperiod` or `resolutiontime`, but not both. If `paranoia` has been set to 0, you should use `minperiod`.

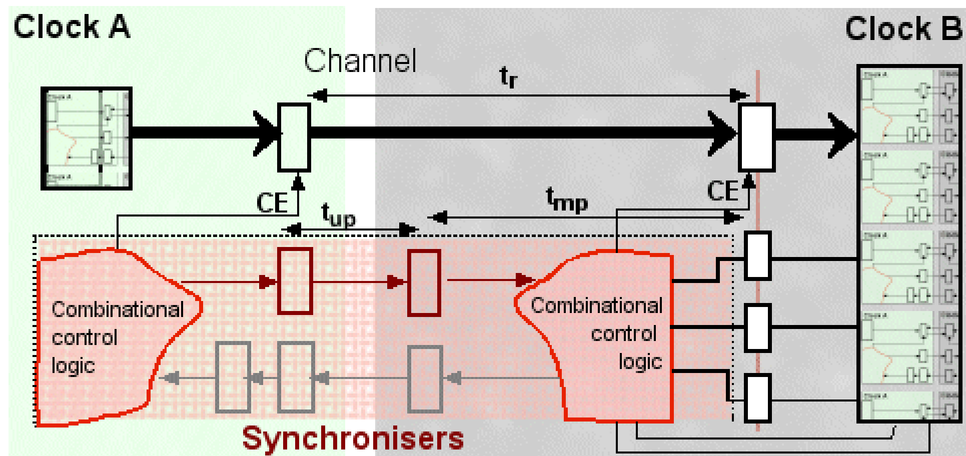


PERIODS WITH PARANOIA AT ITS DEFAULT OF 1

t_{mp} `minperiod`

t_{up} `unconstrainedperiod`

t_p clock period



PERIODS IF PARANOIA SET TO 0

In this case, it is possible that the control signal may be metastable within the first flip-flop, and if `minperiod` is inadequate, the metastability may be propagated into the rest of the circuit.

12.18 offchip specification

The `offchip` specification may be given to a RAM or ROM declaration (you cannot have offchip MPRAMs). When set to 1, the Handel-C compiler builds an external memory interface for the RAM or ROM using the pins listed in the `clk`, `addr`, `data`, `cs`, `we` and `oe` specifications. When set to 0, the Handel-C compiler builds the RAM or ROM on the FPGA or PLD and ignores any pins given with other specifications. You can use the `offchip` specification for EDIF, VHDL or Verilog output.

The compiler generates an error if the `ports` and `offchip` specification are both set to 1 for the same memory.

You cannot initialize an offchip RAM.

Example

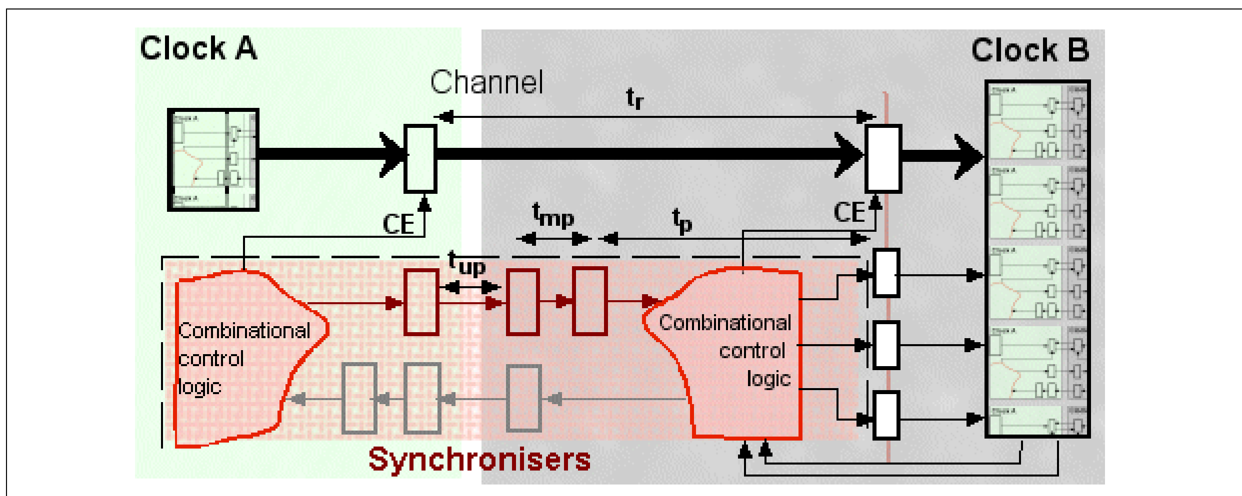
```
ram int 8 a[15][43] with {offchip = 1};
```

12.19 paranoia specification

The paranoia specification controls the number of flip-flops used in synchronization hardware for channels across clock domains. The higher the value for `paranoia`, the higher the stability and latency of the channels. The default is 1, which should be adequate in most cases. If latency is an issue, it is possible to set `paranoia` to 0, but the circuit is more likely to be metastable.

Circuit with paranoia set to default of 1

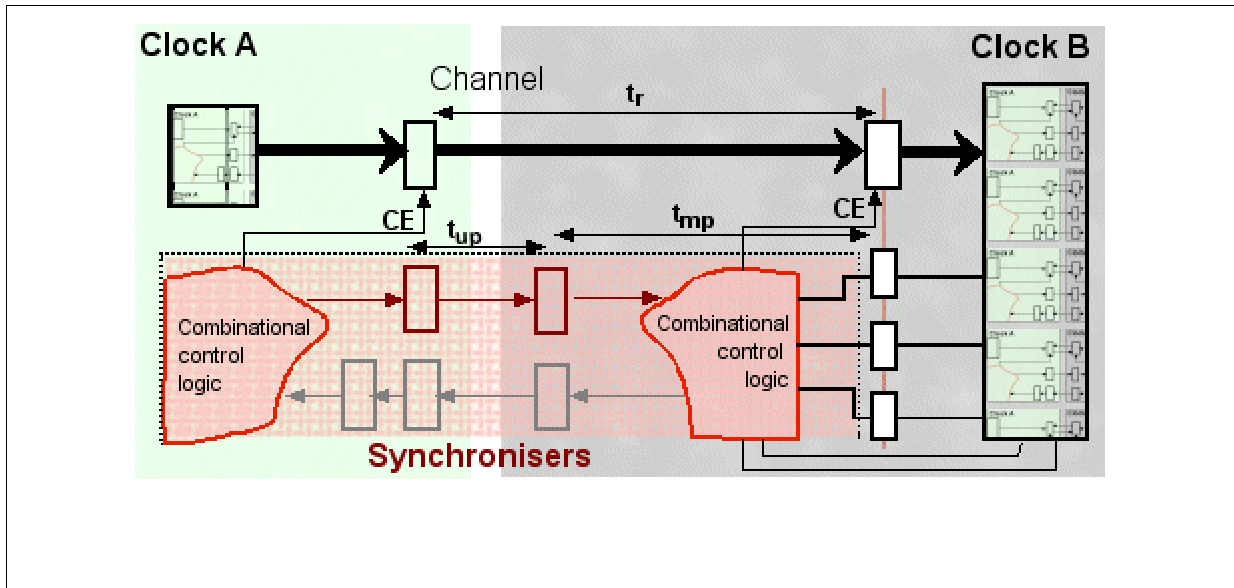
The diagram below shows a circuit with paranoia set to 1. In this case the synchronization data goes through one extra flip flop before generating the clock enable signal for the register.



TIMES WITH PARANOIA AT ITS DEFAULT OF 1

t_r	time to transfer between domains (`paranoia +1`) x t_p
t_{up}	`unconstrainedperiod`
t_{mp}	`minperiod`
t_p	clock period

Circuit showing constraints if paranoia is set to 0



TIMES WITH PARANOIA SET TO 0

12.20 Pin specifications

The `addr`, `data`, `we`, `cs` and `oe` specifications each take a list of device pins and are used to define the connections between the FPGA/PLD and external devices. The specifications only have meaning for EDIF, VHDL and Verilog output. If the specifications are omitted, the place and route tools will assign the pins. The specifications apply to the following objects:

Specification	Meaning	Input bus	Output bus	Tri-state bus	RAM	ROM
`addr`	Address pins	-	-	-	•	•
`data`	Data pins	•	•	•	•	•
`we`	Write Enable pin	-	-	-	•	-
`cs`	Chip Select pin	-	-	-	•	•
`oe`	Output Enable pin	-	-	-	•	•
`clk`	Clock pin	-	-	-	•	•

Pin lists are always given in the order most significant to least significant. Multiple write enable, chip select and output enable pins can be given to allow external RAMs and ROMs to be constructed from multiple devices. For example, when using two 4-bit wide chips to make an 8-bit wide RAM, the following declaration could be used:

```
ram unsigned 8 ExtRAM[256]
    with {offchip=1,
    addr={"P1", "P2", "P3", "P4", "P5", "P6", "P7", "P8"},
    data={"P9", "P10", "P11", "P12", "P13", "P14", "P15", "P16"},
    we={"P17", "P18"},
    cs={"P19", "P20"},
    oe={"P21", "P22"}
};
```

12.21 ports specification

The ports specification may be given to a RAM, ROM or MPRAM declaration and is valid for EDIF, VHDL and Verilog output. When set to 1 the compiler builds an external memory interface, allowing you to connect to dedicated memory resources on an FPGA/PLD or to connect to RAMs in external code. You can only use "simple" types for memories with the ports specification (e.g. int, unsigned; not array or struct).

The compiler generates an error if the ports and offchip specification are both set to 1 for the same memory. All other specifications can be applied.

If you use the ports specification with an MPRAM, a separate interface will be generated for each port.

You cannot initialize a memory that uses the ports specification.

Examples

```
mpram
{
     ram <unsigned 8> ReadWrite[256]; // Read/write port
     rom <unsigned 8> Read[256];      // Read only port
} Joan with {ports = 1, busformat = "B<I>"};
```

generates EDIF ports with names prefixed by Joan_Read and Joan_ReadWrite. For example:

```
(interface
    (port Joan_Read_addr<0> (direction INPUT))
    (port Joan_Read_addr<1> (direction INPUT))

    ......

(interface
    (port Joan_ReadWrite_addr<0> (direction INPUT))
    (port Joan_ReadWrite_addr<1> (direction INPUT))
    .....
```

12.22 properties specification

The `properties` specification can be given to generic interfaces.

If you are generating EDIF, it is used to parameterize instantiations of external black boxes. Each valid property is propagated through to the EDIF netlist as an EDIF property.

If you are generating VHDL or Verilog, the result of the `properties` specification depends on the value of the `bind` specification. When the `bind` specification has a value of 1, it is used to define generics (VHDL) or parameters (Verilog) when creating a user-defined interface to an existing VHDL or Verilog code block. When the `bind` specification is 0, the `properties` specification is used to define attributes for black boxes.

Properties are specified as a list of property items, where each item comprises two or three values:

{*property_name*, *property_value* [, *property_type*]}

- *property_name* is a string
- *property_value* can be a string or an integer
- *property_type* is optional, with 3 possible values (all strings): `"integer"`, `"boolean"` or `"string"`

If your property is a boolean, you need to specify 0 (false) or 1 (true) as the property value, and specify `"boolean"` as the type.

If your property is an integer or string, the type can be inferred from the property value and you do not need to specify it.

Compiler warnings are issued if illegal values are entered, or if there is a mismatch between the property type and property value.

EDIF Example

```
unsigned 6 x;
interface ExtThing(unsigned 6 myvar)
    Inst1ExtThing(unsigned 6 anothervar = x)

    with {properties = {{"LPM_TYPE", "LPM_RAM_DQ"},

    {"LPM_WIDTH", 6, "integer"}}, busformat = "B[N:0]"};
```

This interface will generate an EDIF block with the following EDIF properties: `LPM_TYPE` and `LPM_WIDTH`.

VHDL/Verilog example (bind = 1)

```
interface ExtThing (unsigned 6 myvar)
    Inst1ExtThing(unsigned 6 anothervar = x)
    with {bind = 1,
    properties = {{"prop1", 0, "integer"},
                  {"prop2", "SomeString", "string"},
                  {"prop3", 0, "boolean"},
                  {"prop4", 1, "boolean"}}};
```

For Verilog, this interface will generate the instantiation:

```
ExtThing #(0, // prop1
           "SomeString", // prop2
           0, // prop3
           1) // prop4
InstanceN (.anothervar(x_Out),
           .myvar(W_10))
```

For VHDL, the interface will generate the following component declaration:

```
  component ExtThing
  generic (
           prop1 : integer := 0;
           prop2 : string := "SomeString";
           prop3 : boolean := false;
           prop4 : boolean := true
          );
  port (
        anothervar : in unsigned(5 downto 0);
        myvar : out unsigned(5 downto 0)
       );
end component;
```

and the following component instantiation:

```
InstanceN : ExtThing
  generic map (prop1 => 0,
               prop2 => "SomeString",
               prop3 => false,
               prop4 => true)
  port map (anothervar => x_Out,
            myvar => globals_W_10
  );
```

VHDL/Verilog example (bind = 0)

When the bind specification has a value of 0, attributes are produced instead of generics or parameters, for example:

```
interface ExtThing (unsigned 6 myvar)
    Inst1ExtThing(unsigned 6 anothervar = x)
    with {bind = 0,
    properties = {{"prop1", 0, "integer"},
                  {"prop2", "SomeString", "string"},
                  {"prop3", 0, "boolean"},
                  {"prop4", 1, "boolean"}}};
```

For Verilog and Precision as an output style, this interface will generate a module instantiation with the following Precision attributes:

```
// pragma attribute InstanceN prop1 0
// pragma attribute InstanceN prop2 SomeString
// pragma attribute InstanceN.prop3 0
// pragma attribute InstanceN prop4 1
```

For VHDL, the interface will generate a component instantiation with the following VHDL attributes:

```
attribute prop1: integer;
attribute prop2: string;
attribute prop3: boolean;
attribute prop4: boolean;
attribute prop1 of InstanceN : label is 0;
attribute prop2 of InstanceN : label is "SomeString";
attribute prop3 of InstanceN : label is false;
attribute prop4 of InstanceN : label is true;
```

For Verilog the `properties` specification with a `bind` specification value of 0 is only supported for Precision output style

12.23 pull specification

The `pull` specification may be given to an input or tri-state bus. It is only valid for EDIF output. When set to 1, a pull up resistor is added to each of the pins of the bus. When

set to 0, a pull down resistor is added to each of the pins of the bus. When this specification is not given for a bus, no pull up or pull down resistor is used.

Actel ProASIC and ProASIC+ devices have a pull-up resistor but no pull-down resistor. Refer to the appropriate data sheet for details.

Most Altera devices do not have pull-up or pull-down resistors. ApexII, Mercury, Stratix and Cyclone devices have a pull-up resistor but no pull-down resistor. Refer to the appropriate data sheet for details.

Refer to the Xilinx FPGA data sheet for details of pull up and pull down resistors.

By default, no pull up or pull down resistors are attached to the pins.

Example

```
interface bus_clock_in(int 4 in) InBus() with
    { pull = 1,
      data = {"P4", "P3", "P2", "P1"}
    };
```

12.24 quartus_proj_assign specification

The `quartus_proj_assign` specification can be given to bus-type interfaces or offchip RAM for EDIF output. It allows you to specify Quartus project pins assignments.

Assignments are specified as a list of pairs of items enclosed in braces. The items are strings, and enclosed in quotes. The first item in each pair specifies the item you are assigning, and the second item specifies its value:

{"***assignment_name***", "***assignment_value***"}

Example

```
interface bus_out() MyBusOut(unsigned 3 outPort = MyOutExpr)
   with {quartus_proj_assign = {{"TERMINATION", "Series"},
   {"ENABLE_BUS_HOLD_CIRCUITRY", "On"}},
   standard = "HSTL_I", strength = -1}
```

12.25 rate specification

The `rate` specification may be given to a clock, and is used to specify the frequency (in MHz) at which the clock will need to be driven. The specification only applies to EDIF output (it is ignored for other outputs). The `rate` specification causes Handel-C to generate one of the following:

- a Gate-field Constraints File (GCF) for Actel ProASIC and ProASIC+
- an Assignments and Constraints File (ACF) for use with Max+PlusII for non-Apex Altera devices

- a TCL script (for use with Quartus) for Altera Apex, Cyclone and Stratix devices
- a Netlist Constraints File (NCF) for Xilinx devices

The place-and-route tools then use these timing requirements to constrain the relevant paths so that the part of the design connected to the clock in question can be clocked at the specified rate. In the example below, the clock will need to run at 17.5MHz.

```
set clock = external_divide "D17" 4
    with {rate = 17.5};
```

When `rate` is applied to a divided clock (as shown), it is the divided clock that will be constrained by the specification, not the external clock. Undivided clocks are also constrained to the appropriate value as calculated from the specified rate and the division factor.

12.26 rclkpos, wclkpos and clkpulselen specifications (SSRAM timing)

The `rclkpos`, `wclkpos` and `clkpulselen` may be given to internal or external SSRAM declarations. They are valid for EDIF, VHDL and Verilog outputs. They are specified as floating-point numbers in multiples of 0.5. To use these specifications, you must be using the `external_divide` or `internal_divide` clock types with a division factor of 2 or more.

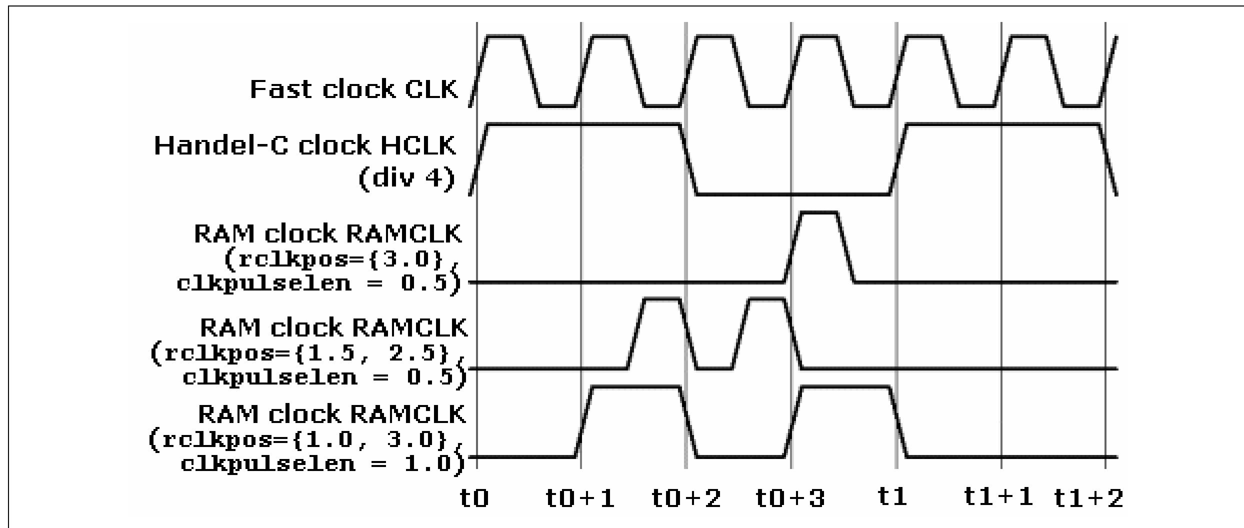
`rclkpos` specifies the positions of the clock cycles of the RAM clock for a read cycle. These positions are specified in terms of cycles of a fast external clock, counting forwards from the rising edge of the divided Handel-C clock rising edge. You need to write the value(s) for the specification in braces. For example, `with {rclkpos = {1.5}}`.

`wclkpos` specifies the positions of the clock cycles of the RAM clock, for a write cycle. You need to write the value(s) for the specification in braces. For example, `with {wclkpos = {1.5, 2.5}}`.

`clkpulselen` specifies the length of the pulses of the RAM clock, in terms of cycles of a fast external clock.

`rclkpos`, `wclkpos` and `clkpulselen` can be applied to the whole of a RAM or MPRAM, or to individual ports within a memory. Specifications applied to the whole memory will apply to each port that does not have its own specification. If you apply `rclkpos` or `wclkpos` to the whole memory, the compiler will issue a warning as `rclkpos` only applies to the read port(s) and `wclkpos` only applied to the write port(s). However, the memory will build correctly.

Illustration



Examples

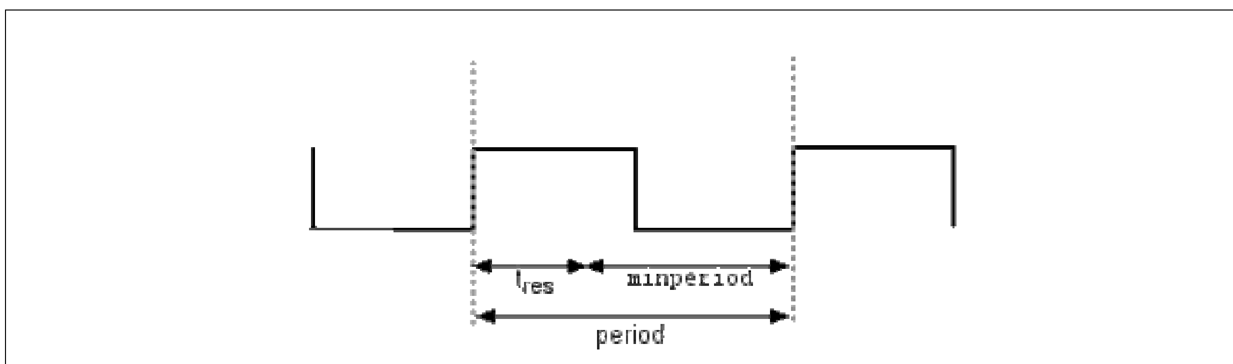
- Applying RAM clock specifications to ports:

```
mpram
{
    rom int 1 ro[16]
        with {rclkpos = {1}, clkpulselen = 0.5};
    wom int 1 wo[16]
      with {wclkpos = {1.5}, clkpulselen = 0.5};
} Mympram;
```

- Pipelined-out SSRAM timing
- Flow through SSRAM
- Targeting external synchronous RAMs

12.27 resolutiontime specification

The `resolutiontime` specification specifies the maximum time in nanoseconds for metastability to resolve in the channel synchronization hardware. It is needed when you are using channels to communicate between multiple clock domains. The higher the value for `resolutiontime` the less time will be available within a clock tick for combinational logic in the synchronizer. This only matters if you have set `paranoia` to 0.

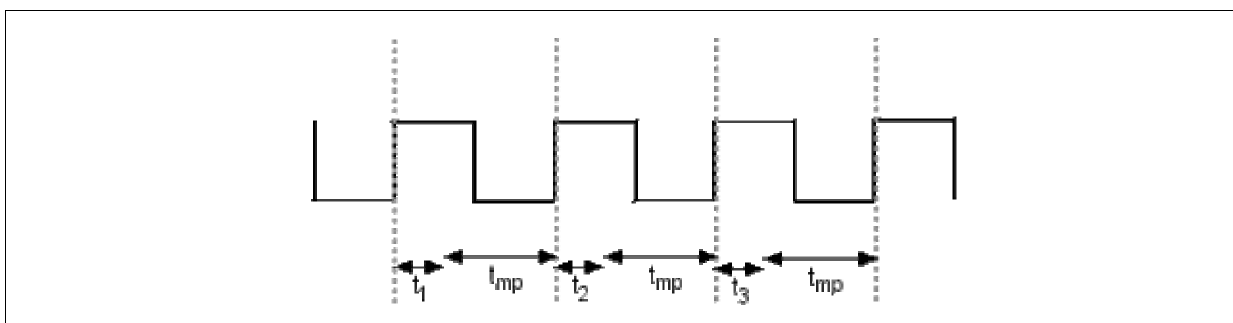


Its value must be less than (clock period x `paranoia`)where `paranoia` is > 0. If `paranoia` has been set to 0, you should use `minperiod` rather than `resolutiontime`.

Either `resolutiontime` or `minperiod` may be set, but not both.

Achieving a given value of resolution time

If you need a higher value of resolution time, you can increase the value of the `paranoia` specification. The resultant value for `minperiod` will be clock period - (`resolutiontime`/`paranoia`).



RESOLUTION TIME SUMMED OVER THREE CLOCK TICKS WHEN PARANOIA = 3

12.28 retime specification

In some circumstances it is desirable to prevent some flip-flops in a circuit from being moved by the retimer. This often occurs when writing interfaces to devices external to

the FPGA or to other IP on the FPGA. The `retime` specification can be added to any variable declaration to lock the position of the flip-flops generated by that variable.

```
unsigned 16 In1, In2 with { retime = 0 };
unsigned 8 SomeOtherVar;
```

For instance in the code above, variables `In1` and `In2` are prevented from moving whereas `SomeOtherVar` can be moved as required by the retimer to meet the specified clock rate.

To disable all flip-flops from being retimed in a specific clock domain, the `retime` specification can be applied to a clock, for instance:

```
set clock = external "D17" with { retime = 0 };
```

12.29 sc_type specification

The `sc_type` specification may be given to `port_in`, `port_out` or generic interfaces to specify the type of a port in SystemC.

Valid string values of this specification are:

sc_int sc_uint bool sc_logic sc_lv

The default type of a port is `bool` if the port is 1 bit wide, `sc_uint` otherwise. You can apply the `sc_type` specification to individual ports. If you place the specification at the end of the interface statement, it will be applied to all the ports.

Example 1: Handel-C ports in SystemC without sc_type specification set:

```
int X_out;
interface port_in(int 1 To) read();
interface port_out() drive(int 4 From = X_out);
```

results in Handel-C generating the SystemC ports:

```
sc_in< bool > To;
sc_out< sc_uint<4> > From;
```

Example 2: Handel-C ports in SystemC with sc_type specification set:

```
int X_out;
interface port_in(int 1 To) read() with {sc_type = "sc_logic"};
interface port_out() drive(int 4 From = X_out} with {sc_type = "sc_int"};
```

results in Handel-C generating the SystemC ports:

```
sc_in< sc_logic > To;
sc_out< sc_int<4> > From;
```

12.30 show specification

The show specification may be given to variable, channel, output bus and tri-state bus declarations. When set to 0, this specification tells the Handel-C simulator not to list this object in its output. This means that it will not appear in the Variables debug window in the GUI, but it can be seen in the Watch window.

The default value of this specification is 1.

```
int 5 x with {show=0};
```

12.31 speed specification

The speed specification may be given to an output or tri-state bus. It only applies to EDIF output. The value of this specification controls the slew rate of the output buffer for the pins on the bus.

For Actel ProASIC and ProASIC+ devices there are three possible values: 0 (slow), 1 (normal) and 2 (fast – default value).

For Altera devices, Xilinx Virtex series and Xilinx Spartan-II and Spartan-3 series, 0 is slow, 1 is fast, and the default value is 1. Refer to the Altera or Xilinx data sheets for details of slew rate control.

Example

```
interface bus_out()
    drive(int 4 signals_from_HC = X_out) with {speed=0};
```

12.32 standard specification

The standard specification may be applied to any external bus interface (not port_in or port_out) connected to pins to select the I/O standard to be used on all pins of that interface. It may also be applied to external clocks and to off-chip memories. If the standard supports it, you can use the strength specification to set the drive current and the dci specification to set digital controlled impedance. The standard specification only applies to EDIF output (it is ignored for other outputs).

standard and dci specifications are ignored if you have used the clockport specification and set it to 0. (The default for clockport is 1.)

Different device families support different standards. Consult the data sheet for a specific device for details of which standard it supports. The compiler will issue errors if a non-supported standard is selected for a particular device, or if the standard specification is used on a family not supporting selectable I/O standards.

12.32.1 Available I/O standards

I/O standard	Handel-C keyword	I/O standard	Handel-C keyword	I/O standard	Handel-C keyword
LVTTL	`"LVTTL"`	HSTL (1.8v) Class I	`"HSTL18_II"`	LVDS (2.5V) see note 1	`"LVDS25"`

LVCMOS (3.3 V)	"LVCMOS33"	HSTL (1.8v) Class II	"HSTL18_II"	LVDS (3.3V)	"LVDS33"
LVCMOS (2.5 V)	"LVCMOS25"	HSTL (1.8v) Class III	"HSTL18_III"	BLVDS (2.5V) see note 1	"BLVDS25"
LVCMOS (1.8 V)	"LVCMOS18"	HSTL (1.8v) Class IV	"HSTL18_IV"	LVPECL (3.3V) see note 1	"LVPECL"
LVCMOS (1.5 V)	"LVCMOS15"	SSTL (2.5v) Class I	"SSTL2_I"	LVDCI (3.3 V) - see note 2	"LVDCI_33"
LVCMOS (1.2 V)	"LVCMOS12"	SSTL (2.5v) Class II	"SSTL2_II"	LVDCI (2.5V) - see note 2	"LVDCI_25"
PCI (33 MHz, 3.3 V)	"PCI33_3"	SSTL(3.3v) Class I	"SSTL3_I"	LVDCI (1.8 V) - see note 2	"LVDCI_18"
PCI (33 MHz, 5.0 V)	"PCI33_5"	SSTL (3.3v) Class II	"SSTL3_II"	LVDCI (1.5 V) - see note 2	"LVDCI_15"
PCI (66 MHz, 3.3 V)	"PCI66_3"	SSTL (1.8v) Class I	"SSTL18_I"	LVDCI (3.3 V, split termination) - see note 3	"LVDCI_DV2_33"
PCI-X	"PCIX"	SSTL (1.8v) Class II	"SSTL18_II"	LVDCI (2.5 V, split termination) - see note 3	"LVDCI_DV2_25"
GTL	"GTL"	CTT	"CTT"	LVDCI (1.8 V, split termination) - see note 3	"LVDCI_DV2_18"
GTL+	"GTL+"	AGP (1x)	"AGP-1X"	LVDCI (1.5 V, split termination) - see note 3	"LVDCI_DV2_15"
HSTL (1.5v) Class I	"HSTL_I"	AGP (2x)	"AGP-2X"		
HSTL (1.5v) Class II	"HSTL_II"			HyperTransport	"HyperTransport"

HSTL (1.5v) Class III	"HSTL_III"
HSTL (1.5v) Class IV	"HSTL_IV"

Notes:

1. The only differential I/Os supported for tri-state interfaces are BLVDS25 on the VirtexII, VirtexII-Pro and Virtex-4 and LVDS25 and LVPECL33 on the VirtexE.
2. LVDCI standards are equivalent to using LVCMOS standards with a dci specification of 1
3. LVDCI split termination standards are equivalent to using LVCMOS standards with a dci specification of 0.5

If no I/O standard is specified, the default for Actel ProASIC and ProASIC+ is LVCMOS33 (with drive strength "High" or "Max"). The default for all other devices is LVTTL (with a drive current of 12mA in the case of Xilinx families supporting Select I/O).

Examples

```
set clock = external "C1" with {standard = "HSTL_III"};
interface bus_out()
    Eel(int 4 outPort=x)
    with {data = dataPins0, standard = "HSTL_I"};

interface bus_ts(unsigned 3)
    Baboon(unsigned 3 ape1 = y, unsigned 1 ape2 = en)
    with {data = dataPinsT, standard = "LVTTL", strength = 24};
```

12.32.2 I/O standards supported by different chips

You can specify the I/O standard for a particular device using the standard specification. Consult the manufacturer's datasheet for the standards supported by a particular chip.

- Spartan, Spartan XL and Flex10 series devices do not support selectable standards.
- Actel ProASIC and ProASIC+ only support the LVCMOS33 (default) and LVCMOS25 standards.
- If you are using differential I/Os with Mercury devices, you need to use the dedicated pins interfacing to the HSDI (high-speed differential interface)

12.32.3 I/O standard details

The following input/output standards are available in Handel-C. To select a standard, use the standard specification.

AGP (1x, 2x) – Advanced Graphics Port

The AGP standard is specified by the Advanced Graphics Port Interface Specification Revision 2.0 introduced by Intel Corporation for graphics applications. AGP is a voltage-referenced standard requiring a reference voltage of 1.32 V, an input/output source voltage of 3.3 V and no termination. This standard requires a Differential Amplifier input buffer and a Push-Pull output buffer.

BLVDS - Bus Low Voltage Differential Signal

BLVDS is a differential I/O scheme, although it is not currently defined by any IEEE/EIA/TIA industry standards. Unlike LVDS and LVPECL, which are intended for point-to-point communications, BLVDS allows for bi-directional data transfer over the same set of transmitter-receiver pin pairs (also known as transceivers). It thus enables transmission of high-speed differential signals over multipoint backplanes. Due to the bi-directional transfer capability, 50 ohm termination resistors are needed at both ends of the transmission line.

CTT – Center Tap Terminated

The CTT standard is a 3.3V memory bus standard, specified by JEDEC Standard JESD 8-4, Center-Tap-Terminated (CTT) Low-Level, High-Speed Interface Standard for Digital Integrated Circuits, and sponsored by Fujitsu. CTT is a voltage-referenced standard requiring a reference voltage of 1.5 V, an input/output source voltage of 3.3 V and a termination voltage of 1.5 V. The CTT standard is a superset of LVTTL and LVCMOS. CTT receivers are compatible with LVCMOS and LVTTL standards. CTT drivers, when un-terminated, are compatible with the AC and DC specifications for LVCMOS and LVTTL. This standard requires a Differential Amplifier input buffer and a Push-Pull output buffer.

GTL+ – Gunning Transceiver Logic Plus

The GTL+ standard is a high-speed bus standard (JESD 8-3) first used by Intel Corporation for interfacing with the Pentium Pro processor and is often used for processor interfacing or communication across a backplane. GTL+ is a voltage-referenced standard requiring a 1.0 V input reference voltage and board termination voltage of 1.5 V. The GTL+ standard is an open-drain standard that requires a minimum input/output source voltage of 3.0 V.

HSTL – High-speed Transceiver Logic

The HSTL standard, specified by JEDEC Standard JESD 8-6, High-Speed Transceiver Logic (HSTL), is a 1.5 V output buffer supply voltage based interface standard for digital integrated circuits. This is a voltage-referenced standard, and has four variations or classes. Classes I & II require a reference voltage of 0.75 V and a termination voltage of 0.75 V; classes III & IV require a reference voltage of 0.9 V and a termination voltage of 1.5 V. All four classes require an input/output source voltage of 1.5 V. This standard requires a Differential Amplifier input buffer and a Push-Pull output buffer.

HyperTransport

HyperTransport technology is a differential high-speed, high-performance I/O interface standard. It is a point-to-point standard requiring a 2.5-V VCCIO, in which each HyperTransport technology bus consists of two point-to-point unidirectional links. Each link is 2 to 32 bits. The HyperTransport technology standard does not require an input reference voltage. However, it does require a 100-ohm termination resistor between the two signals at the input buffer.

LVCMOS (3.3 V) – 3.3 Volt Low-Voltage CMOS

This standard is an extension of the LVCMOS standard and is defined in JEDEC Standard JESD 8-A, Interface Standard for Nominal 3.0 V/3.3 V Supply Digital Integrated Circuits. This is a single-ended general-purpose standard also used for 3.3V applications. It uses a 5V-tolerant CMOS input buffer and a Push-Pull output buffer. This standard requires a 3.3V input/output source voltage, but does not require the use of a reference voltage or a board termination voltage.

LVCMOS (2.5 V) – 2.5 Volt Low-Voltage CMOS

This standard is an extension of the LVCMOS standard and is documented by JEDEC Standard JESD 8-5, 2.5 V ± 0.2 V (Normal Range) and 1.7 V to 2.7 V (Wide Range) Power Supply Voltage and Interface Standard for Non-terminated Digital Integrated Circuit. This is a single-ended general-purpose standard, used for 2.5V (or lower) applications. It uses a 5V-tolerant CMOS input buffer and a Push-Pull output buffer. This standard requires a 2.5V input/output source voltage, but does not require the use of a reference voltage or a board termination voltage. Altera documentation refers to this standard as simply "2.5 V".

LVCMOS (1.8 V) – 1.8 Volt Low-Voltage CMOS

This standard is an extension of the LVCMOS standard and is documented by JEDEC Standard JESD 8-7, 1.8 V ± 0.15 V (Normal Range) and 1.2 V to 1.95 V (Wide Range) Power Supply Voltage and Interface Standard for Non-terminated Digital Integrated

Circuit. This is a single-ended general-purpose standard, used for 1.8V power supply levels and reduced input and output thresholds. It uses a 5V-tolerant CMOS input buffer and a Push-Pull output buffer. This standard does not require the use of a reference voltage or a board termination voltage. Altera documentation refers to this standard as simply "1.8 V".

LVCMOS (1.5 V) – 1.5 Volt Low-Voltage CMOS

This standard is an extension of the LVCMOS standard. This is a single-ended general-purpose standard, used for 1.5V applications. It uses a 5V-tolerant CMOS input buffer and a Push-Pull output buffer. This standard does not require the use of a reference voltage or a board termination voltage. Altera documentation refers to this standard as simply "1.5 V".

LVCMOS (1.2 V) - 1.2 Volt Low-Voltage CMOS

This standard is an extension of the LVCMOS standard. This is a single-ended general-purpose standard, used for 1.2V applications. It uses a 5V-tolerant CMOS input buffer and a Push-Pull output buffer. This standard does not require the use of a reference voltage or a board termination voltage.

LVDCI - Low Voltage Digital Controlled Impedance

Xilinx Virtex II devices are able to provide controlled impedance input buffers and output drivers that eliminate reflections without an external source termination. Output drivers can be configured as controlled impedance drivers, or as controlled impedance drivers with half impedance. Inputs can be configured to have termination to V_{cco} or to $V_{cco}/2$ (split termination), where V_{cco} is the input/output source voltage. All of these are available at four voltage levels: 1.5 V, 1.8 V, 2.5 V and 3.3 V. For further details, please refer to the Xilinx Data Book.

LVDS – Low Voltage Differential Signal

LVDS is a differential I/O standard. It requires that one data bit be carried through two signal lines. The LVDS I/O standard is used for very high-performance, low-power-consumption data transfer. Two key industry standards define LVDS: IEEE 1596.3 SCI-LVDS and ANSI/TIA/EIA-644. Both standards have similar key features, but the IEEE standard supports a maximum data transfer of 250 Mbps. The use of a reference voltage or a board termination voltage is not required, but a 100 ohm termination resistor is required between the two traces at the input buffer.

LVPECL – Low Voltage Positive Emitter Coupled Logic

LVPECL is a differential I/O standard. It requires that one data bit be carried through two signal lines. The LVPECL standard is similar to LVDS. In LVPECL, the voltage swing between the two differential signals is approximately 850 mV. The use of a reference voltage or a board termination voltage is not required, but an external termination resistor is required.

LVTTL – Low Voltage TTL

The Low-Voltage TTL, or LVTTL standard is a single ended, general purpose standard for 3.3V applications that uses an LVTTL input buffer and a Push-Pull output buffer. The LVTTL interface is defined by JEDEC Standard JESD 8-A, Interface Standard for Nominal 3.0 V/3.3 V Supply Digital Integrated Circuits. This standard requires a 3.3V output source voltage, but does not require the use of a reference voltage or a termination voltage.

PCI (33 MHz, 3.3 V) & PCI (66 MHz, 3.3 V) – 3.3 Volt PCI

The PCI standard specifies support for 33 MHz, 66 MHz and 133 MHz PCI bus applications. It uses a LVTTL input buffer and a Push-Pull output buffer. This standard requires a 3.3V input output source voltage, but not the use of input reference voltages or termination.

PCI (33 MHz, 5.0 V) – 5.0 Volt PCI

Some Xilinx devices may be configured in this mode (an extension of the 3.3 Volt PCI standard), which makes them 5V tolerant. No Altera devices currently support this mode.

PCI-X

The PCI-X standard is an enhanced version of the PCI standard that can support higher average bandwidth and has more stringent requirements.

SSTL2 – Stub Series Terminated Logic for 2.5 V

The SSTL2 standard, specified by JEDEC Standard JESD 8-9, Stub-Series Terminated Logic for 2.5 Volts (SSTL-2), is a general purpose 2.5 V memory bus standard sponsored by Hitachi and IBM. This is a voltage-referenced standard, and has two variations or classes, both of which require a reference voltage of 1.25 V, an input/output source voltage of 2.5 V and a termination voltage of 1.25 V. This standard requires a Differential Amplifier input buffer and a Push-Pull output buffer. SSTL2 is used for high-speed SDRAM interfaces.

SSTL3 – Stub Series Terminated Logic for 3.3 V

The SSTL2 standard, specified by JEDEC Standard JESD 8-8, Stub-Series Terminated Logic for 3.3 Volts (SSTL-3), is a general purpose 3.3 V memory bus standard sponsored by Hitachi and IBM. This is a voltage-referenced standard, and has two variations or classes, both of which require a reference voltage of 1.5 V, an input/output source voltage of 3.3 V and a termination voltage of 1.5 V. This standard requires a Differential Amplifier input buffer and an Push-Pull output buffer. SSTL3 is used for high-speed SDRAM interfaces.

SSTL18 - Stub Series Terminated Logic for 1.8 V

The SSTL18 standard, specified by JEDEC Preliminary Standard JC42.3, is a general purpose 1.8V memory bus standard. This is a voltage-referenced standard, and has two variations or classes, both of which require a reference voltage of 0.90 V, an input/output source voltage of 1.8 V and a termination voltage of 0.90 V. This standard requires a Differential Amplifier input buffer and a Push-Pull output buffer. SSTL18 is used for high-speed SDRAM interfaces.

GTL – Gunning Transceiver Logic Terminated

The GTL standard is a high-speed bus standard (JESD 8-3) invented by Xerox. Xilinx has implemented the terminated variation for this standard (Altera has not). This standard requires a differential amplifier input buffer and an Open Drain output buffer.

12.32.4 Differential I/O standards

Differential I/O standards can be used with bus-type interfaces, offchip memories and external clocks in EDIF output. They are specified using the standard specification. The differential I/O standards supported by Handel-C are LVDS25, LVDS33, BLVDS25, LVPECL33 and HyperTransport.

If you want to build a tri-state interface, you can use only the BLVDS25 standard.

To specify pins for a bus_type interface with a differential I/O, use the `data` specification. Pins are specified in pairs enclosed in braces:

```
interface bus_in (unsigned 2 datain) I()
    with {standard = "LVDS25",
    data = {{"P1", "P2"}, {"P3", "P4"}}}
```

The first pin in a pair is the positive one. You can omit the second pin of each pair, but you still need to enclose the single pins within braces.

You also need to specify pair of pins enclosed in braces for pin specifications for offchip memories (addr, we, cs, oe and clk) when you are using a differential I/O. For example:

```
ram unsigned 4 ExtRAM[256]
    with {offchip=1, standard = "LVPECL33",
    addr={{"P1", "P2"}, {"P3", "P4"}, {"P5", "P6"}, {"P7", "P8"}},
    data={{"P9", "P10"}, {"P11", "P12"}, {"P13", "P14"}, {"P15", "P16"}},
    we={{"P17", "P18"}},
    cs={{"P19", "P20"}},
    oe={{"P21", "P22"}}
    };
```

If you use a differential I/O for an external clock, the pins are specified using the set clock construct, rather than the data specification:

```
set clock = external {"C1", "C2"}
    with {standard = "LVDS25"}
```

The standard specification is ignored for VHDL and Verilog output, but if you have used a data specification with pairs of pins, and then build the code for VHDL or Verilog output, the first pin in each pair will be assigned and the other pin will be ignored.

12.33 std_logic_vector specification

The std_logic_vector specification may be given to port_in, port_out or generic interfaces, where you want to use a std_logic_vector port instead of an unsigned port in VHDL. Set std_logic_vector to 1 if you want to:

- instantiate an external block of code in Handel-C generated VHDL, and the external block uses one or more std_logic_vector ports
- produce a block of VHDL that will be linked into another VHDL block that uses one or more std_logic_vector ports.

The default value for std_logic_vector is 0. You can apply the std_logic_vector specification to an individual port. If you place the specification at the end of the interface statement, it will be applied to all the ports.

The std_logic_vector specification is ignored for all outputs except for VHDL

Example 1: Handel-C instantiation of a Bloo component with std_logic_vector set to 0 (default):

```
interface Bloo(unsigned 1 myin) B(unsigned 4 myout = x) with
{std_logic_vector = 0};
```

results in Handel-C generating this VHDL instantiation of the Bloo component:

```
component Bloo
port (
        myin : out std_logic;
        myout : in unsigned (3 downto 0)
   );
   end component;
```

Example 2: Handel-C instantiation of a Bloo component with std_logic_vector set to 1:

```
interface Bloo(unsigned 1 myin)
    B(unsigned 4 myout = x) with {std_logic_vector = 1};
```

results in Handel-C generating this VHDL instantiation of the `Bloo` component:

```
component Bloo
port (
        myin : out std_logic_vector (0 downto 0);
        myout : in std_logic_vector (3 downto 0)
     );
end component;
```

12.34 strength specification

The `strength` specification may be used in conjunction with the `standard` specification on any external bus interface (not `port_in` or `port_out`) connected to pins to select the drive current (in mA) to be used on all pins of that interface. It may also be applied to off-chip memories. You can only use the `strength` specification for EDIF output.

Different device families support different values. The compiler will issue warnings if a non-supported value is selected for a particular device. Check the device datasheet to confirm what values it supports

The following standards do not support drive strength selection: PCI, GTL, HSTL III, HSTL IV, CTT, AGP(1x), AGP(2x), LVDS, LVPECL, LVDCI and BLVDS.

The following devices do not support drive strength selection for any standards: Excalibur, Apex 20, Apex 20KE and Apex 20KC.

Example

```
interface bus_out() Eel(int 4 outPort = x)
   with {data = dataPins0, standard = "HSTL_I", strength = -1};
interface bus_ts(unsigned 3 inPort) Baboon(ape1 = y, ape2 = en)
    with {data = dataPinsT, standard = "LVTTL",
          strength = 24};
```

12.35 synchronous specification

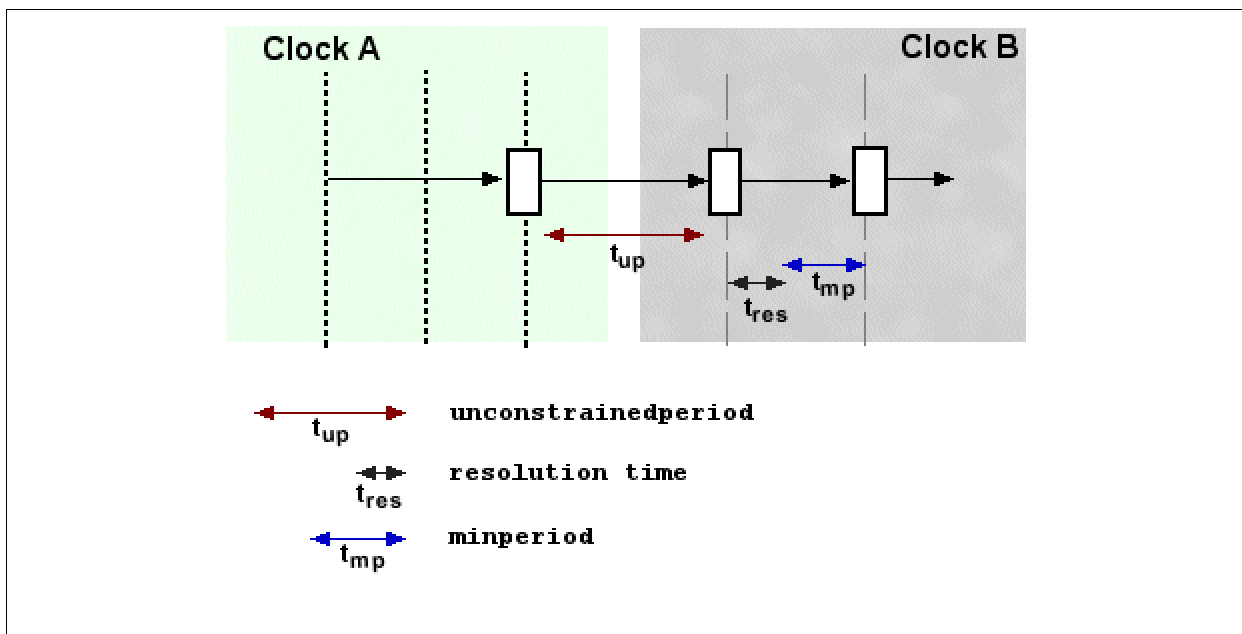
The `synchronous` specification may be given to a reset signal. The value of this specification controls whether the reset is synchronous (occurs on next clock tick) or asynchronous (occurs immediately). The default is asynchronous (0)

Example

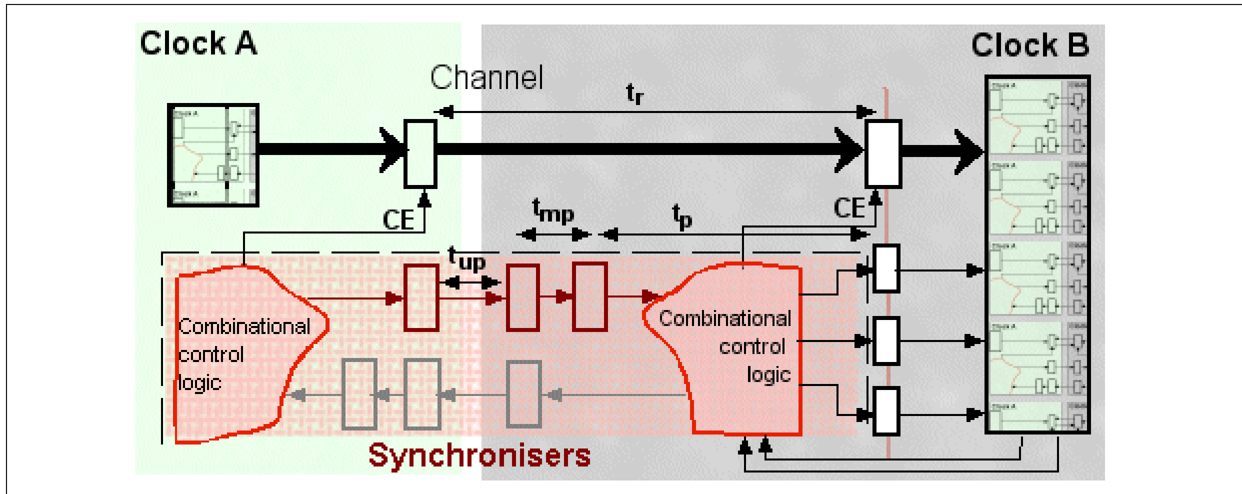
```
set reset  = external with {synchronous=1};
```

12.36 unconstrainedperiod specification

The `unconstrainedperiod` specification gives the maximum period in nanoseconds on channel control paths between clock domains. If this specification is not used, the place and route tools may generate a warning for affected paths crossing the clock domain.



If the specification is used, it applies to unconstrained paths into the clock domain. The diagram below shows where it is used.



TIMES WITH PARANOIA AT ITS DEFAULT OF 1

t_r: time to transfer between domains (`paranoia +1`) x t_p

t_{up}: unconstrained period

t_{mp}: minperiod

t_p: clock period

```
set clock = external with {unconstrained period=10};
```

12.37 vhdl_type specification

The `vhdl_type` specification may be given to `port_in`, `port_out` or generic interfaces to specify the type of a port in VHDL.

Valid string values of this specification are:

unsigned signed std_logic std_logic_vector

The default type of a port is `std_logic` if the port is 1 bit wide, `unsigned` otherwise. You can apply the `vhdl_type` specification to individual ports. If you place the specification at the end of the interface statement, it will be applied to all the ports.

The `vhdl_type` specification replaces the deprecated `std_logic_vector` specification

Example 1: Handel-C instantiation of a Bloo component without vhdl_type specification set:

```
interface Bloo(unsigned 1 myin) B(unsigned 4 myout = x);
```

results in Handel-C generating this VHDL instantiation of the `Bloo` component:

```
COMPONENT Bloo
PORT (
        myin : OUT std_logic;
        myout : IN unsigned (3 DOWNTO 0)
   );
END COMPONENT;
```

Example 2: Handel-C instantiation of a Bloo component with vhdl_type applied to entire interface:

```
interface Bloo(unsigned 1 myin)
          B(unsigned 4 myout = x) with {vhdl_type = "std_logic_vector"};
```

results in Handel-C generating this VHDL instantiation of the `Bloo` component:

```
COMPONENT Bloo
PORT (
        myin : OUT std_logic_vector (0 DOWNTO 0);
        myout : IN std_logic_vector (3 DOWNTO 0)
     );
END COMPONENT;
```

Example 3: Handel-C instantiation of a Bloo component with vhdl_type applied to individual ports:

```
interface Bloo(unsigned 1 myin with {vhdl_type = "std_logic_vector"} )
          B(unsigned 4 myout = x with {vhdl_type = "signed"} );
```

results in Handel-C generating this VHDL instantiation of the `Bloo` component:

```
COMPONENT Bloo
PORT (
        myin : OUT std_logic_vector (0 DOWNTO 0);
        myout : IN signed (3 DOWNTO 0)
     );
END COMPONENT;
```

12.38 warn specification

The `warn` specification may be given to a variable, RAM, ROM, channel, bus or clock. It can be used for any DK output. When set to zero, certain non-crucial warnings will be disabled for that object. When set to one (the default value), all warnings for that object will be enabled.

```
int 5 x with {warn=0};
```

12.39 wegate specification

The `wegate` specification may be given to external or internal RAM declarations to place the write-enable strobe. You can only use this specification with an undivided clock. If it is used in the absence of SRAM clock specifications (`rclkpos`, `wclkpos` and `clkpulselen`), it forces the generation of an asynchronous memory or memory port. If you have a divided clock, use the `westart` and `welength` specifications instead. The `wegate` specification is valid for EDIF, VHDL and Verilog output.

When the `wegate` specification is set to 0, the write strobe will appear throughout the Handel-C clock cycle. When set to -1, the write strobe will appear only in the first half of the Handel-C clock cycle. When set to 1, the write strobe will appear only in the second half of the Handel-C clock cycle.

You can apply the specification to the whole of a RAM or MPRAM, or to individual write ports within an MPRAM. Specifications applied to individual ports take precedence over specifications applied to the whole memory. Specifications applied to the whole memory apply to each port that does not have its own specification.

12.40 westart and welength specifications

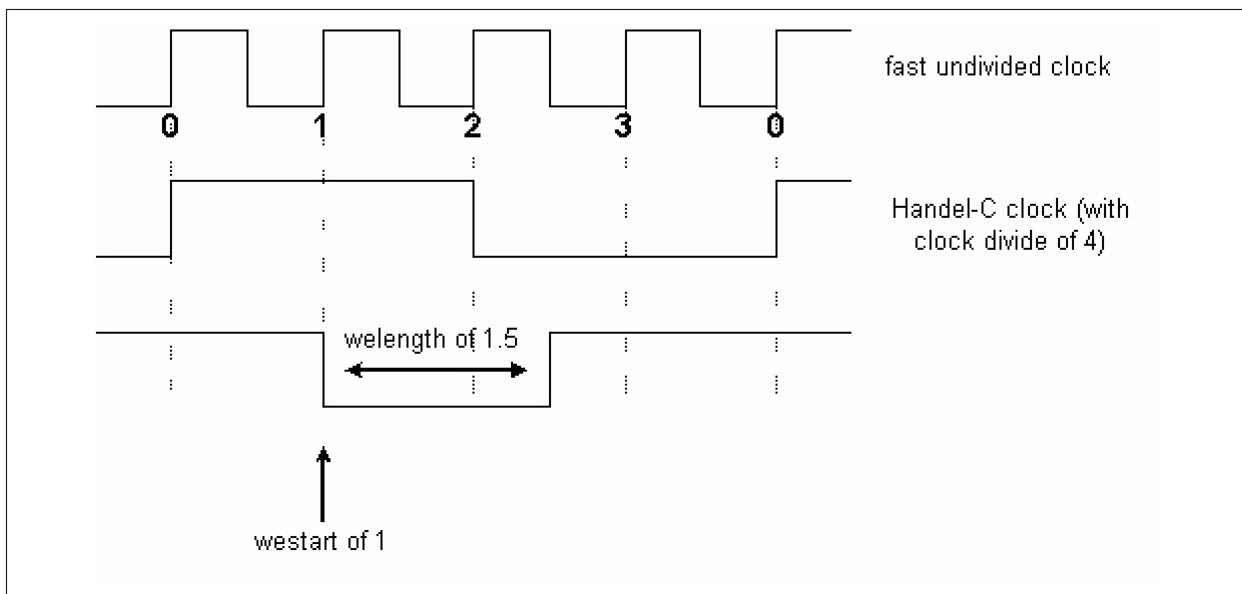
The `westart` and `welength` specifications position the write enable strobe within the Handel-C clock cycle. If they are used in the absence of SRAM clock specifications (`rclkpos`, `wclkpos` and `clkpulselen`), they force the generation of an asynchronous memory or memory port. The specifications may be given to internal or external RAM declarations. You can only use these specifications together with `external_divide` or `internal_divide` clock types with a division factor greater than 1. If you have an undivided clock, use the `wegate` specification instead. `westart` and `welength` are valid for EDIF, VHDL and Verilog output.

`westart` is used to specify the starting position of the write enable strobe, and `welength` is used to specify its length. For both of these specifications, a unit value corresponds to a single cycle of the fast clock which has been divided in order to generate the Handel-C clock. The size of `welength` and `westart` can be given in multiples of 0.5, but (`westart + welength`) must not exceed the clock divide.

You can apply the specification to the whole of a RAM or MPRAM, or to individual write ports within a memory. Specifications applied to the whole memory will apply to each port that does not have its own specification.

Examples

```
//applying the specifications to the whole RAM
set clock = external_divide "P78" 4;
ram unsigned 6 x[34] with {westart = 1, welength = 1.5};
```



WRITE ENABLE STROBE WITH A WESTART OF 1, A WELENGTH OF 1.5, AND A CLOCK DIVIDE OF 4

```
//applying the specifications to ports
mpram
{
    wom unsigned 6 r[32}
        with {westart = 1, welength = 1.5};
    wom unsigned 6 s[32];
    rom unsigned 6 t[32];
    rom unsigned 6 u[32};

} with {westart = 1.5, welength = 0.5};
```

This example would result in a compiler warning as the specifications at the end would be applied to all ports that do not have their own specification (`s`, `t` and `u`). `t` and `u` are read-only ports and therefore cannot have write-enable specifications. However, the `mpram` would build correctly with the first set of specifications applied to port `r` and the second set to port `s`.

13 Handel-C preprocessor

The preprocessor is invoked by the Handel-C compiler as the first stage in the compilation process, and is used to manipulate the text of source code files. Correct use of this tool can simplify code development and the subsequent maintenance process. There are a number of functions performed by the preprocessor:

- Macro substitution
- File inclusion
- Conditional compilation
- Line splicing
- Line control
- Concatenation
- Error generation
- Predefined macro substitution

Communication with the preprocessor occurs through the use of *directives*. Directives are lines within source code which begin with the # character, followed by an identifier known as the *directive name*. For example, the directive to define a macro is '#define'.

13.1 Preprocessor macros

Simple macros

The preprocessor supports several types of macros. Simple macros (or manifest constants) involve the simplest form of macro substitution and are defined with the form:

`#define` *`name sequence-subsitute`*

Any occurrences of the token *name* found in the source code are replaced with the token sequence *sequence-substitute*, which may include spaces. All leading and trailing white spaces around the replacement sequence are removed. For example:

```
#define FOO 1024
#define loop_forever while (1)
```

Parameterized macros

You can also define macros with arguments. This allows replacement text to be passed as parameters. For example:

```
#define mul(A, B) A*B
```

This will replace

```
x = mul (2, 3);
```

with

```
x = 2 * 3;
```

Take care to preserve the intended order of evaluation when passing parameters. For example the line

```
x = mul (a - 2, 3);
```

will be expanded into

```
x = a - 2 * 3;
```

The multiplication is evaluated first, then the result subtracted from variable a. This is almost certainly not the intention, and errors of this type may be difficult to locate.

If a parameter name is preceded by a # when declared as part of a macro, it is expanded into a quoted string by the preprocessor. E.g., if a macro is defined:

```
#define quickassert(X) assert (width(X)==1,0 "Width of " #X " is not
1!\n");
```

The line:

```
quickassert(length);
```

will expand into:

```
assert (width(X)==1,0 "Width of length is not 1!\n");
```

Undefining identifiers

To undefine an identifier, the `#undef` directive may be used. E.g.

```
#undef FOO
```

Note that no error will occur if the identifier has not previously been defined.

Preprocessor directives cannot be used unexpanded in a library; use macro procedures instead.

13.2 File inclusion

File inclusion makes it possible to easily manage and reuse declarations, macro definitions, and other code. The feature is helpful when writing general purpose functions and declarations which can be reused for a number of designs. File inclusion is achieved using directives of the form:

```
#include "filename"
```

or

```
#include <filename>
```

Such lines are replaced by the contents of the file indicated by *filename*. If the file name is enclosed by quotation marks, the preprocessor looks for the file in the directory containing source code for the current design. If the file cannot be found there, or the file name is enclosed with angular brackets, the search examines user-defined include file directories (specified using Tools>Options>Directories), and the main Agility include file directory.

13.3 Conditional compilation

Conditional directives

You can control preprocessing with conditional directives. These statements can add a great deal of flexibility to source code. For example, they may be used to alter the behaviour of a design, depending upon whether a macro definition is present. Conditional statements must begin with an #if directive and an expression to be evaluated, and end with the #endif directive. Valid directives are:

```
#if expression

#elif expression
#else

#endif
```

Example

```
#if a==b
    // include this section if a is equal to b
#elif a>b
    // include this section if a is greater than b
#else
    // otherwise include this section
#endif
```

If the expression is evaluated to be zero, then any text following the directive will be discarded until a subsequent #elif, #else, or #endif statement is encountered; otherwise the lines will be included as normal. Note that each directive should be placed individually on its own line starting at column 0.

A useful application for conditional directives is easy exclusion of code without the use of comments. For example:

```
#if (0)
    // Code for debugging purposes
#endif
    // Code continues
```

By amending the above evaluation to (1), the code can quickly be included during compilation.

Conditional definition

To test for the existence of a macro definition, use the following directives:

`#ifdef` ***identifier*** (equivalent to #if defined (identifier))

`#ifndef` ***identifier*** (equivalent to #if !defined (identifier))

These are used in the same way as `#if`, but are followed by an identifier, rather than an expression. The `#ifndef` directive is often used to ensure that source code is only included once during compilation. E.g.

```
#ifndef UTILCODE
#define UTILCODE

// Utility code is written here

#endif
```

13.4 Line control

A directive of the form:

`#line` ***integer***

instructs the compiler that the next source line is the line number specified by ***integer***. If a filename token is also present:

`#line` ***integer*** "***filename***"

the compiler will additionally regard ***filename*** as the name of the current input file.

13.5 Concatenation in macros

If a macro is defined with a token sequence containing a ## operator, each instance of ## is removed (along with any surrounding white space), thus concatenating adjacent tokens into one. For example, if the macro below was declared:

```
    #define million(X) X ## e6
```

then

```
    i = million (3);
```

is expanded into:

```
    i = 3e6;
```

Take care when specifying parameters. In the example above, if 3e6 was passed instead of 3, then the line would be expanded into:

```
    i = 3e6e6;
```

which would result in an error.

13.6 Error generation

Fatal error messages may be reported during preprocessing using the directive:

`#error` ***`error_message`***

This may be useful with conditional compilation if your design only supports certain combinations of parameter definitions.

13.7 Predefined macro substitution

The preprocessor contains a number of useful predefined macros which may be placed into source code:

`_ _FILE_ _`	Expands to the name of the current file being compiled
`_ _LINE_ _`	Expands to the number of the current source line
`_ _TIME_ _`	Expands to the current time of compilation in the form hh:mm:ss
`_ _DATE_ _`	Expands to the current date of compilation in the form mmm dd yyyy

13.8 Line splicing

You can splice multiple lines together by placing a backslash character ('\') followed by a carriage return between them. This feature allows you to break lines for aesthetic purposes when writing code, which are then joined by the preprocessor prior to compilation. For example, if a macro is defined:

```
#define ERRORCHECK(error)  \
    if (error!=0)          \
    return (error)
```

The line:

```
ERRORCHECK(i);
```

Expands to:

```
if (i!=0)
    return i;
```

14 Language syntax

The complete Handel-C language syntax is given in BNF-like notation.

The overall syntax for the program is:

program ::= {*external_declaration*}

```
void main(void)
{
   {declaration}
   {statement}
}
```

Language

```
external_declaration ::= function_definition
             | declaration
             | set_statement
```

14.1 Language syntax conventions

BNF (Backus-Naur Format) is a way to describe the syntax of file formats. It consists of definitions of the form

identifier ::= definition

The identifier is a word which describes this part of the syntax.
The ::= represents "consists of".
The definition lists the permitted contents of the identifier.

The conventions used in this language reference are:

- Terminal symbols are set in typewriter font `like this`.
- Non-terminal symbols are set in italic font *like this*.
- Square brackets [...] denote optional components.
- Braces {...} denotes zero, one or more repetitions of the enclosed components.
- Braces with a trailing plus sign {...}+ denote one or several repetitions of the enclosed components.
- Parentheses (...) denote grouping.

14.2 Keyword summary

The keywords listed below are reserved and cannot be used for any other purpose.

<table>
<tr><th>Keyword</th><th>Meaning</th><th>ANSI-C/C++ ?</th></tr>
<tr><td>=</td><td>assignment operator</td><td>Yes</td></tr>
<tr><td>;</td><td>statement terminator</td><td>Yes</td></tr>
<tr><td>,</td><td>comma operator</td><td>Yes</td></tr>
<tr><td>{ }</td><td>code block delimiters</td><td>Yes</td></tr>
<tr><td><></td><td>type clarifier</td><td>No</td></tr>
<tr><td>(</td><td>open delimiter</td><td>Yes</td></tr>
<tr><td>)</td><td>close delimiter</td><td>Yes</td></tr>
<tr><td>[ ]</td><td>array index delimiters, bit selection</td><td>Yes</td></tr>
<tr><td>[:]</td><td>bit range selection</td><td>No</td></tr>
<tr><td>!</td><td>logical NOT operator</td><td>Yes</td></tr>
<tr><td>!</td><td>output to channel</td><td>No</td></tr>
<tr><td>+</td><td>addition operator</td><td>Yes</td></tr>
<tr><td>-</td><td>subtraction operator</td><td>Yes</td></tr>
<tr><td>-</td><td>unary minus operator</td><td>Yes</td></tr>
<tr><td>*</td><td>multiplication operator</td><td>Yes</td></tr>
<tr><td>/</td><td>division operator</td><td>Yes</td></tr>
<tr><td>%</td><td>modulo operator</td><td>Yes</td></tr>
<tr><td>\\</td><td>drop LSBs</td><td>No</td></tr>
<tr><td><-</td><td>take LSBs</td><td>No</td></tr>
<tr><td>?</td><td>read from channel</td><td>No</td></tr>
<tr><td>?</td><td>conditional expression</td><td>Yes</td></tr>
<tr><td>^</td><td>Bitwise XOR</td><td>Yes</td></tr>
<tr><td>&</td><td>Bitwise AND</td><td>Yes</td></tr>
<tr><td>|</td><td>Bitwise OR</td><td>Yes</td></tr>
<tr><td>~</td><td>bitwise NOT</td><td>Yes</td></tr>
<tr><td>&&</td><td>Logical AND</td><td>Yes[1]</td></tr>
<tr><td>||</td><td>Logical OR</td><td>Yes[1]</td></tr>
<tr><td>.</td><td>structure member operator</td><td>Yes</td></tr>
<tr><td><<</td><td>left-shift operator</td><td>Yes</td></tr>
<tr><td>>></td><td>right shift operator</td><td>Yes</td></tr>
<tr><td><</td><td>less than operator</td><td>Yes[1]</td></tr>
<tr><td>></td><td>greater than operator</td><td>Yes[1]</td></tr>
<tr><td><=</td><td>less or equal operator</td><td>Not standard[1]</td></tr>
</table>

>=	greater or equal operator	Not standard[1]
==	equality operator	Not standard[1]
!=	inequality operator	Not standard[1]
++	increment operator	Not standard
--	decrement operator	Not standard
+=	assignment operator	Not standard
-=	assignment operator	Not standard
*=	assignment operator	Not standard
/=	assignment operator	Not standard
%=	assignment operator	Not standard
<<=	assignment operator	Not standard
>>=	assignment operator	Not standard
&=	assignment operator	Not standard
|=	assignment operator	Not standard
^=	assignment operator	Not standard
...	Reserved. Not valid in Handel-C, but can be used for C/C++ calls.	Yes
->	structure pointer operator	Yes
@	concatenation operator	No

[1] Note, the results of these tests are a single bit `unsigned int`

Keyword	Meaning	ANSI-C/C++ ?
assert	diagnostic macro to print to stderr	Not standard
auto	auto variable	Yes
break	immediate exit from code block	Yes
case	selection within switch and prialt	Yes
chan	define channel variable	No
chanin	simulator channel in	No
chanout	simulator channel out	No
char	8-bit variable	Yes
clock	define clock	No
const	specify that variable's value will not change	Yes
continue	force next iteration of loop	Yes
default	default case within switch, prialt	Yes
delay	wait one clock cycle	No
do	start do while loop	Yes
double	Reserved. Not valid in Handel-C	C-only
else	conditional execution	Yes
enum	enumeration constant	Yes
expr	define macro as expression	No
extern	define global variable	Yes
external	clock from device pin	No
external_divide	clock from device pin with integer division	No
family	define target device's family	No
float	Reserved. Not valid in Handel-C	C-only
for	for loop iteration	Yes
goto	jump to specified label	Yes
if	conditional execution	Yes
ifselect	conditional compilation on compile-time selection	No
in	define scope for local macro expression declaration	No
inline	declaration of inline function	No

`int`	definable width variable	Yes
`interface`	declaration of off-chip interface	No
`internal`	use internal clock	No
`internal_divid`	internal clock with integer division	No
`intwidth`	set integer width	No
`let`	start declaration of local macro expression	No
`long`	declare 32-bit variable	Yes
`macro`	declare a macro	No
`mpram`	declare a multi-port RAM	No
`par`	execute statements in parallel	No
`part`	define target hardware	No
`prialt`	execute first ready channel	No
`proc`	define macro as procedure	No
`ram`	declare a RAM (array)	No
`register`	declare register variable	Yes
`releasesema` (***semaphore***)	free ***semaphore***	No
`reset`	reset design	No
`return`	return from function	Yes
`rom`	declare a ROM (array)	No
`select`	select expression or macro expr at compile time	No
`sema`	declare a semaphore	No
`set`	specify device family or part, int width, target, reset or clock	No
`seq`	execute statements in sequence	No
`shared`	declare a shared expression	No
`short`	declare 16-bit variable	Yes
`signal`	declare a signal object	No
`signed`	declare a signed variable	Yes
`sizeof`	Reserved. Not valid in Handel-C	Yes
`static`	specify variable with limited scope	Yes
`struct`	declare a structure variable	Yes
`switch`	switch statement (between cases)	Yes

`try reset(Condition) {...}`	execute statements if Condition is true during execution within related try block	No
`trysema`	Test if semaphore owned. Take if not.	No
`typedef`	define type	Yes
`typeof`	return type of expression	No
`undefined`	specify a variable of undefined width	No
`union`	Reserved. Not valid in Handel-C	Yes
`unsigned`	declare an unsigned variable	Yes
`void`	specify void return type,	Yes
`volatile`	declare volatile variable	Yes
`while`	loop statement	Yes
`width`	return integer width	No
`with`	specify interface, signals, channels, RAM and ROM types, variables etc.	No
`wom`	declare a WOM (array)	No

The following character sequences are also reserved:

```
/*    */    //    #    "    '
```

14.3 Constant expressions

The following constants are available in Handel-C

- Identifiers
- Integer constant
- Character constants
- String constant
- Floating-point constants

14.3.1 Identifiers: syntax

```
identifier ::= letter {letter | 0...9}

letter ::= A...Z | a...z | _
```

14.3.2 Integer constants: syntax

```
integer_constant ::= [-]{1...9}+{0...9}
    | [-](0x | 0X){0...9 | A...F | a...f}+
    | [-](0){0...7}
    | [-](0b | 0B){0...1}+
```

14.3.3 Character constants: syntax

character is any printable character or any of the following escape codes.

Escape code	ASCII value	Meaning
\a	7	Bell (alert)
\b	8	Backspace
\f	12	Form feed
\t	9	Horizontal tab
\n	10	New line
\v	11	Vertical tab
\r	13	Carriage return
\"	-	Double quote mark
\0	0	String terminator
\\	-	Backslash
\’	-	Single quote mark
\?	-	Question mark

14.3.4 Strings: syntax

```
string ::= "{character}"
```

14.3.5 Floating-point constants: syntax

```
float_constant::=
    [{0...9}+].{0...9}+[(e | E)[+|-]{0...9}+][f | F | l | L]
    | {0...9}+.[(e | E)[+|-]{0...9}+][f | F | l | L]
    | {0...9}+(e | E)[+|-]{0...9}+[f | F | l | L]
```

14.4 Functions and declarations: syntax

```
function_definition ::= declaration_specifiers declarator
        compound_ statement [ with  initializer ;]
    | declarator compound_statement [ with initializer ;]

declaration ::= declaration_specifiers [ init_declarator_list ] [with
initializer ] ;
    | interface_declaration
    | macro_declaration

declaration_specifiers ::= storage_class_specifier [ declaration_specifiers
]
    | type_specifier [ declaration_specifiers]
    | type_qualifier [ declaration_specifiers]

storage_class_specifier ::= auto
    | register
    | inline
    | typedef
    | extern
    | static

type_specifier ::= void
    | char
    | short
    | int
    | long
    | float
    | double
    | signed
    | unsigned
    | typeof ( expression )
    | signal_specifier
    | channel_specifier
    | ram_specifier
    | struct_or_union_specifier
    | enum_specifier
    | typedef_name

type_qualifier ::= const
    | volatile
```

```
typedef_name ::= identifier

init_declarator_list ::= declarator
[= initializer] { ,declarator [= initializer]}
```

14.5 Macro/shared exprs/procs: syntax

```
macro_declaration ::= macro_proc_decl
    | macro_expr_decl

macro_proc_decl ::= [ static | extern]
    macro_proc_spec identifier
    [ ( [ macro_param{, macro_param} ] )] statement
    [ with initializer ;]

macro_expr_decl ::= [ static | extern]
    macro_expr_spec identifier
    [ ( [macro_param{, macro_param} ] )] ;
    |  [ static | extern] macro_expr_spec identifier
    [ ( [macro_param{, macro_param} ] )] = let_initializer
    [with initializer ] ;

macro_proc_spec ::= macro proc

macro_expr_spec ::= macro expr
    | shared expr

let_initializer ::= initializer
    | let macro_expr_decl in let_initializer

macro_param ::= identifier
```

14.6 Interfaces: syntax

```
interface_declaration ::= interface identifier
( [int_parameter_declaration
        { , int_parameter_declaration} ] ) identifier
        ([ assignment_expr_spec {, assignment_expr_spec } ] )
        [with initializer];
    | interface_type_declarator
    | old_style_interface_declarator

interface_type_declarator :: = interface identifier
        ( [ int_parameter_proto{ , int_parameter_proto}] )
        identifier ( [int_init_parameter_declaration
        { ,
int_init_parameter_declaration} ] )
```

This format is deprecated but retained for compatibility reasons:

```
old_style_interface_declarator ::= interface identifier
    ( [int_parameter_declaration  {, int_parameter_declaration}] )
    identifier ([ assignment_expr_spec {,assignment_expr_spec})
    [with  initializer ] ;

interface ::= [ static  | extern] interface

 int_parameter_proto::= declaration_specifiers
    | declaration_specifiers declarator
    | declaration_specifiers abstract_declarator
    | declaration_specifiers width

int_parameter_declaration ::= declaration_specifiers [with initializer ]
    | declaration_specifiers declarator [with initializer ]
    | declaration_specifiers abstract_declarator [ with initializer ]
    | declaration_specifiers width [with initializer ]

int_init_parameter_declaration ::= int_parameter_declaration
    | declaration_specifiers declarator [ = initializer
] [with initializer ]

assignment_expr_spec ::= assignment_expression [with initializer ]
```

14.7 Structures and unions: syntax

```
struct_or_union_specifier ::= aggregate_form [ identifier] {
{struct_declaration}+ }
    | aggregate_form identifier

aggregate_form ::= struct
    | union
    | mpram


struct_declaration ::= { type_specifier | type_qualifier}+
    {struct_declarator} + [with initializer ];

struct_declarator ::= declarator
      | [declarator]: constant_expression
```

The current version of Handel-C does not support unions.

14.8 Enumerated types: syntax

```
enum_specifier ::= enum [ identifier ] { enumerator {,[ enumerator]} }
    | enum identifier
enumerator ::= identifier
    | identifier = constant_expression
```

14.9 Signal specifiers: syntax

```
signal_specifier ::= signal < type_name >
    | signal
```

14.10 Channel syntax

```
channel_specifier ::= chan [ < type_name > ]
    | chanin  [ < type_name > ]
    | chanout  [ < type_name > ]
```

14.11 Ram specifiers: syntax

```
ram_specifier ::= ram [ < type_name > ]
    | rom  [ < type_name > ]
    | wom [ < type_name > ]
```

14.12 Declarators: syntax

```
declarator ::= [ width ] pointer direct_declarator


width ::= undefined
    | primary_expression


direct_declarator ::= identifier
    | ( pointer direct_declarator )
    | direct_declarator [ [constant_expression] ]
    | direct_declarator ( [ {parameter_declaration}+ ] )


pointer ::= *
    | * type_qualifier
    | * pointer
    | * type_qualifier pointer
```

14.13 Function parameters: syntax

```
parameter_declaration ::= declaration_specifiers
    | declaration_specifiers width
    | declaration_specifiers abstract_declarator
    | declaration_specifiers declarator
```

14.14 Type names and abstract declarators: syntax

```
type_name ::= { type_specifier | type_qualifier}+
    |  { type_specifier | type_qualifier}+ abstract_declarator
    |  { type_specifier | type_qualifier}+ width

abstract_declarator ::= [width] pointer direct_abstract_declarator

direct_abstract_declarator ::= ( pointer direct_abstract_declarator )
    |  [direct_abstract_declarator][ [constant_expression] ]
    |  [direct_abstract_declarator] ( [ {parameter_declaration}+ ] )
```

14.15 Statements: syntax

```
statement ::= semi_statement ;
    | non_semi_statement

semi_statement ::= expression_statement
    | do statement while ( expression )
    | jump_statement
    | assert ( constant_expression [, assignment_expression {,
assignment_expression}] )
    | delay
    | channel_statement
    | set_statement

non_semi_statement ::= labelled_statement
    | compound_statement
    | selection_statement
    | iteration_statement
```

The following statements can appear in `for` start/end conditions:

```
for_statement ::= non_semi_statement
    | expression_statement
    | do statement while ( expression )
    | assert ( constant_expression , constant_expression
        [, assignment_expression{, assignment_expression}] )
    | delay
    | channel_statement
```

These are the statements that can appear in prialt blocks:

```
prialt_statement ::= semi_statement ;
      |   non_semi_prialt_statement

non_semi_prialt_statement ::= prialt_labelled_statement
    | compound_statement
    | selection_statement
    | iteration_statement

labelled_statement ::= identifier : statement
    | case constant_expression : statement
      |   default : statement

prialt_labeled_statement ::= identifier : prialt_statement
    | case channel_statement : prialt_statement
    | default : prialt_statement

expression_statement ::= [ expression ]

channel_statement ::= unary_expression ! expression
    | logical_or_expression ? expression

jump_statement ::= goto identifier
    | continue
    | break
    | return
    | return expression

selection_statement ::= if ( expression ) statement if
    | if ( expression ) statement else statement
    | ifselect ( constant_expression ) statement if
    | ifselect ( constant_expression ) statement else statement
    | switch ( expression ) statement
    | prialt { [{ prialt_statement }+] }

set_statement ::= set part = STRING
    | set clock = clock
    | set family = identifier
    | set intwidth = constant_expression
    | set intwidth = undefined
    | set reset = reset

clock ::= internal expression [with initializer ]
```

```
    | external expression [with initializer ]
    | internal_divide expression expression [with initializer ]
    | external_divide expression expression [with initializer ]

reset ::= internal expression
    | external expression

iteration_statement ::= while ( expression ) statement
    | for ( [for_statement] ; [ expression] ;
        [ for_statement] ) statement
```

14.15.1 Compound statements with replicators

```
compound_statement ::= [seq | par] {{ declaration} {statement} }
    |   [seq | par] ( [repl_macro_param{, repl_macro_param}]
;constant_expression;
    [repl_update_param {, repl_update_param}] ) {{declaration} {statement}
}
```

14.16 Replicators: syntax

Replicator initialization definitions

```
repl_macro_param ::= repl_param = initializer
    | ( repl_param = initializer )
```

Replicator update definitions

```
repl_update_param ::= repl_update_param_body
    | ( repl_update_param )

repl_update_param_body ::= repl_param assignment_operator initializer
    |  ++ repl_param
    |  repl_param ++
    |  -- repl_param
    |  repl_param --

repl_param ::= identifier
    | ( repl_param )
```

14.17 Expressions: syntax

```
constant_expression ::= assignment_expression

expression ::= assignment_expression
    | expression, assignment_expression}

assignment_expression ::= conditional_expression
    | unary_expression assignment_operator assignment_expression

assignment_operator ::= =  | *=  |  /=  |  %=  |  +=  |  -=  |  <<=  |  >>=
|  &=
     |  ^=  |  |=

initializer ::= assignment_expression

conditional_expression ::= logical_or_expression
    | logical_or_expression ? expression : conditional_expression

logical_or_expression ::= logical_and_expression
    | logical_or_expression || logical_and_expression

logical_and_expression ::= inclusive_or_expression
    | logical_and_expression && inclusive_or_expression

inclusive_or_expression ::= exclusive_or_expression
    | inclusive_or_expression | exclusive_or_expression

exclusive_or_expression ::= and_expression
    | exclusive_or_expression ^ and_expression

and_expression ::= equality_expression
    | and_expression & equality_expression

equality_expression ::= relational_expression
    | equality_expression == relational_expression
    | equality_expression != relational_expression

relational_expression ::= cat_expression
    | relational_expression < cat_expression
    | relational_expression > cat_expression
    | relational_expression <=  cat_expression
```

```
    | relational_expression >= cat_expression

cat_expression ::= shift_expression
    | cat_expression @ shift_expression

shift_expression ::= additive_expression
    | shift_expression << additive_expression
    | shift_expression >> additive_expression

additive_expression ::= multiplicative_expression
    | additive_expression + multiplicative_expression
    | additive_expression - multiplicative_expression

multiplicative_expression ::= take_drop_expression
    | multiplicative_expression * take_drop_expression
    | multiplicative_expression / take_drop_expression
    | multiplicative_expression % take_drop_expression

take_drop_expression ::= cast_expression
    | take_drop_expression <- cast_expression
    | take_drop_expression \\ cast_expression

cast_expression ::= unary_expression
    | ( type_name ) cast_expression

unary_expression ::= postfix_expression
    |  ++ unary_expression
    |  -- unary_expression
    |  unary_operator cast_expression
    |  sizeof unary_expression
    |  sizeof ( type_name )
    |  width ( expression )

unary_operator ::= &  |  +  |  -  |  ~  |  !  |  *

postfix_expression ::= select_expression
    | postfix_expression [ expression ]
    | postfix_expression [ expression : expression ]
    | postfix_expression [ : expression ]
    | postfix_expression [ expression : ]
    | postfix_expression [ ]
```

```
    | postfix_expression ( [assignment_expression     {,
assignment_expression}] )
    | postfix_expression . identifier
    | postfix_expression -> identifier
    | postfix_expression ++
    | postfix_expression --

select_expression ::= primary_expression
    | select ( constant_expression , constant_expression ,
        constant_expression )

primary_expression ::= identifier
    | constant
    | ( expression )
    | { }
    | {[initializer {, initializer}[, ] ]}

constant ::= integer_constant
    | character_constant
    | string_constant

integer_constant ::= NUMBER

character_constant ::= CHARACTER

string_constant ::= STRING
```

15 Index

-- (postfix and prefix operators) 95
- (subtraction) 104
! 44, 107
!= 106
(macro concatenation)............. 306
#define 303
#elif 305
#else 305
#endif 305
#ifdef 305
#ifndef 305
#include 41, 116, 304
#undef 303
% (modulo)................................. 104
(line breaker) 307
*/ (comments delimiter) 10
. (structure member operator) 109
/ (division) 104
/* (comments delimiter) 10
// (comments delimiter) 10
?...44, 109
@ (concatenation)......................... 102
[ ] (bit selection)........................... 103
\ (drop operator)........................... 102
^ (bitwise XOR) 108
__clock 162
__isfamily() construct 182
+ (addition) 104
++ (prefix and postfix operators) 95
< (less than) 106
<- (take operator)......................... 102
<< (shift operator)........................ 102
<= (less than or equal).................. 106
<> (type qualifier) 72
= (assignment)............................... 81
== (equal to) 106
> (greater than) 106
-> (structure pointer operator)........ 109
>= (greater than or equal) 106
>> (shift operator)........................ 102
1.2V 286, 292
1.5V 286, 292
1.8V 286, 291
2.5V 286, 291
3.3V 286, 291
33MHz 3.3V...........................286, 293
33MHz 5.0V...........................286, 293
66MHz 3.3V...........................286, 290
abstract declarators 321
ACF files 269, 281
Actel....................................180, 183
  devices 180, 183
  on-chip RAM 210
  specifying reset pin 186
addition 104
addr 276
AGP....................................286, 290
AGP I/O standard 286, 289, 290
  AGP-1X 286
  AGP-2X 286
algorithms...............................22, 178
  debugging 22, 178
Altera 180, 183, 211
  devices 180, 183
  on-chip RAM 211
  ROMs 256
ampersand (address operator) 43
ANSI-C 15, 25
  calling from Handel-C 65
  compared to Handel-C 15, 25
Apex devices 180, 183
  arithmetic operators 104

constraints files 281
I/O standards supported 289, 296
mprams 59
RAM 211
architectural types 44
arithmetic operators 104
arrays 34, 36, 46, 56, 119
channels
arrays 46
functions 119, 120, 121
indices 36
multi-dimensional 34, 56
pointers to 34
assert 99
assertion failed 99
assignments 81
asterisk (indirection operator) 43
asynchronous RAM 187, 192, 300
divided clock 300
examples 188, 190
generating 300
timing 187
undivided clock 300
asynchronous reset 186, 297
attributes 245
auto 64
base specification 254
basic concepts 5
bidirectional data transfers 221, 222, 224
clocked input 224
registered input 222
binary 30
bind specification 254
bit fields 38
bit manipulation 101
operators 101
bit selection 103
bitwise logical operators 108
bitwise AND 108
bitwise NOT 108
bitwise OR 108
bitwise XOR 108
block RAM 256
block specification 256
blocks 178, 256
data transfer 178
BLVDS 286, 290
BLVDS I/O standard 286, 290, 294
break 78, 86, 88, 89
breaking lines 307
buffer specification 259
BUFG 262
bus_clock_in 218, 221
bus_in 218, 219
bus_latch_in 218, 220
bus_out 218, 221
bus_ts 218, 221
bus_ts_clock_in 218, 224
bus_ts_latch_in 218, 222
buses 50, 52, 219, 220, 221, 230, 233, 243
bidirectional 221, 222, 224
clocked 221
input 219
latched 220
naming 243
read/write 221
read/write clocked 224
registered 220, 222
simulating 230
specification 50, 52
timing 233, 235
write 221
busformat specification 243, 259

C language ... 15, 25
compared to Handel-C 15, 25
C++ ... 65
calling from Handel-C 65
type mapping in Handel-C 65
case ... 88
casting ... 17, 39, 96, 97
chan ... 44, 319
chanin ... 177, 319
channels ... 6, 44, 46
arrays 46
between clock domains 162, 164, 166, 168, 238
chanin and chanout 177
channels
arrays ... 46
communication 6, 44, 162, 164
examples 164
metastability 237, 238
reading from 44
restrictions 46
simulating 177
simultaneous access 46
specifying 319
syntax 319
writing to 44
chanout ... 177, 319
char ... 32
character constants ... 315
chips ... 180
clk ... 204, 261
clkpulselen ... 195, 282
clock cycles used ... 141, 147
clock pin specifications ... 261
clock position specifications ... 282
clock rate ... 281
clocked reading from external pins ... 221
clockport specification ... 262
clocks ... 141, 160, 161, 162
clock domains 162, 172
clock pins 231, 261
current 162
cycles 195
dummy 160
external 161
external resynchronization 172
fast 187, 193
internal 161
inverted 193, 196
locating 160
multiple 160, 162, 172
period 154
position specifications 282
reading from external pins 221
resynchronizing 172
simulation 160
source 160
specifying 160
SSRAM 193, 196
SSRAMs 193, 195, 196, 261
combinational loops ... 90, 149, 262
comments ... 10
communication ... 6, 44, 162, 164
between clock domains 162, 164
channels 6, 44, 164
comparison ... 106, 107
implicit 107
operators 106
signed/unsigned 107
compile-time ... 70, 99, 131, 132
messages 99
selection 70, 131, 132
complex declarations ... 69, 71
complex expressions ... 72, 95

compound statements with replicators 324
concatenation 102, 306
operator 102
preprocessor 306
conditional compilations 305
conditional directives 305
conditional execution (if ... else)........ 84
conditional operator................ 109, 131
const 71
constant expressions 314
constant macro expressions............ 130
constants 30, 315
binary 30
character 315
decimal 30
hexadecimal 30
manifest 303
octal 30
constraints 264, 269, 270, 281
files 269, 281
pins 264
timing 269, 270, 281
continue 82
conversion 17, 96, 97
cs 276
CTT 286, 290
CTT I/O standard 286, 289, 290
current clock 162
Cyclone devices 183, 209
constraints files 281
I/O standards supported 289, 296
mprams 57, 59
pull-up resistors 280
RAMs 211
targeting embedded memory 209
data 22, 178
file format 178
input and output 22
data specification 264
dci specification 265
DDR devices 193
debug 99
assertions 99
decimal...30
declarations.......................... 118, 320
disambiguating 72
functions 118
interfaces 49
mpram 57
RAM 54
ROM 54
syntax 320
declarators 320
default 78, 88
defining the clock 160
delay 90, 141, 147
device specifiers........................... 183
devices 179, 180, 182, 183, 218
detecting current device 182
external 218
specifying 183
differential................................... 294
differential I/O standards................ 294
Digital Controlled Impedance 265
disambiguator................................72
division 104
do ... while 86, 89
does not equal 106
domains.........................162, 168, 172
channel timing 168
multiple clocks 162, 172
double 16, 309
drop operator 102

EAB................................180, 211, 256
EDIF....................................243, 259
buses 259
wire names 243
efficiency 154
else...84
enum 37, 319
enumerated types 37, 319
equal to 106
error generation.......................99, 307
ESB....................................180, 256
examples
asynchronous RAM 188, 190
between clock domains 164
function pointers 123
functions 114, 120, 121
interfacing to hardware 225
macros 114
mprams 61
optimizing code 123, 154, 157
prialt 79
SSRAM 200, 205, 207
targeting external RAM 192, 204
targeting ports to specific tools 243
timing 141, 152, 270
Excalibur devices 180, 183
I/O standards supported 289, 296
RAM 211
exit from code block 89
expressions 23, 95, 314
comparison with ANSI-C 23
complex 72
constant 314
shared 135, 136
syntax 325
timing 95
extern (external variables) 65
extern (linking to C/C++ code) 65
external clocks................161, 187, 189
external hardware 218
external ROMs 212
external variables...........................65
external_divide 160, 161
extfunc 266
extinst 266
extlib.. 266
extpath....................................... 268
families..........................179, 180, 183
recognized 180, 183
fast external clock 187
fifolength 268
FIFOs 6, 268
code example 46, 145
timing 144
files.. 304
including 304
reading and writing 178
timing constraints 269, 281
Flex devices...........................180, 183
constraints files 281
I/O standards supported 289
RAMs 211
float 16, 309
floating-point arithmetic 16
floating-point constants................. 315
for loops 86, 89
differences from ANSI-C 86
formatting bus and wire names 243
FPGA devices 179, 180
function calls 127
parallel 127
simultaneous 127
functions........... 19, 111, 114, 117, 118
arrays 119, 120, 121

clock cycles 114, 141
compared to macros 111, 113, 114
definitions and declarations 118
differences to ANSI-C 19
examples 114
inline 68
parameters 320
pointers 122, 123
prototypes 118
restrictions 117, 127, 129
returning macro expr 71
scope 119
shared 127
syntax 316, 320
GCF files 269, 281
generic interfaces.......................... 218
generics (VHDL)............................ 278
getting started 5
goto 83
greater than 106
greater than or equal to 106
GTL I/O standard 286, 289, 290, 294
GTL 286, 290, 294
GTL+ 286, 290
Handel-C5, 12, 14, 15, 25, 29, 75, 95, 111, 245, 309
code 5
compared to ANSI-C 15, 25
expressions 95
functions 111
getting started 5
keywords 309
macros 111
object specifications 245
operators 12
programs 5
statements 75
syntax 309
types 14
values and widths 29
Handel-C preprocessor 303
hardware 177
interfaces 177, 225
hexadecimal 30
HSTL I/O standard...........286, 289, 291
Class I 286
Class II 286
Class III 286
Class IV 286
I/O standards .. 265, 286, 289, 290, 294
differential 294
I/O standards supported 289, 296
IBUFG.. 262
identifiers.................................... 314
if...else 84
ifselect.. 132
implicit compares 107
in (let...in) 137
indirection operator 43
indirection techniques 39, 43
inferring widths...............................33
infile.. 269
initialization............................. 59, 73
MPRAM 59
RAM and ROM 54, 210, 211
structures 36
variables 73
inline 64, 68
input 220, 221, 269, 290, 294
clocked 221
files 269
latched 220
standards 286, 290, 294
int...31

integer 31, 315
constants 315
range 31
syntax 315
interfaces.....................49, 50, 52, 218
bidirectional buses 221, 222, 224
bus_* interfaces 219, 220, 221, 222, 224
customized 218, 242
debugging 228
declaration 49, 51
definition 50, 52
format 242
generic 218, 242
overview 49
pointers 42
port_* interfaces 241
simulating 228
sorts 218
specification 50, 52
syntax 318
types 218
interfacing......................177, 218, 241
with external hardware 218
with external logic 218, 241
with memory 186, 213
with the simulator 177
internal clocks 160, 161
internal RAM and ROM 54
internal_divide.......................160, 161
intime..................................269, 270
intwidth 33
inverted clocks................193, 196, 273
ISO-C 15, 25
calling from Handel-C 65
compared to Handel-C 15, 25
keywords 309
labels 83
language basics 9
language summary..........9, 15, 25, 309
language syntax............................ 309
latch... 220
register 220
latency 48, 157, 169
left shift 102
less than 106
less than or equal to 106
let ... in 137
line control 306
line splicing 307
loc attribute................................. 264
locating the clock 160
logic depth 154
reducing 154
logic types.....................................31
logical operators 107
long 32
loops 20, 85, 86, 89, 149
combinational 149
do ... while 86
for loops 86
termination 89
while loops 85
LVCMOS I/O standard286, 289, 291, 292
1.2V 286, 292
1.5V 286, 292
1.8V 286, 291
2.5V 286, 291
3.3V 286, 291
LVDCI I/O standard ..265, 286, 289, 292
1.5V 286
1.8V 286
2.5V 286
3.3V 286

split termination 265, 286
LVDS I/O standard ...286, 289, 292, 294
LVPECL I/O standard 286, 289, 293, 294
LVTTL I/O standard 286, 289, 293
macro expressions71, 130, 131, 133, 135, 136
in widths 71
macro procedures 138, 139
macros 111, 130, 131, 138, 303
compared to functions 111, 113, 114
differences to ANSI-C 19
examples 114
introduction 130
parameterized 131, 138
preprocessor 303
recursion 133, 135, 136
substitution 303, 307
syntax 317
main function 9
malloc 22
manifest constants 303
mapping of different width ports........ 59
maximum clock rate 154
member operators......................... 109
memory. 54, 57, 62, 186, 187, 193, 196
Actel 210
allocation 22
Altera 211
asynchronous 187, 192, 300
block 256
initialization 54
multi-port 57
off-chip 274
on Cyclone devices 209
on Stratix devices 209
on-chip 180, 210, 211, 212
pipelining 196, 207
RAM 54, 186
restrictions 97
ROM 54, 186
simultaneous access 97
specifications 256, 274
synchronous 193, 204, 261, 282
type 256
WOM 62
Xilinx 212
Mercury devices 180, 183
I/O standards supported 289, 296
mprams 57
pull up resistors 280
RAM 211
merging pins 231, 232
metastability.......................... 168, 235
channels across clock domains 168, 237, 238
clock domains 168
examples 172
external resynchronization 172
stabilizing data in interfaces 236
MIF files...................................... 211
minperiod 167, 273
modulo arithmetic 104
mpram (multi-ported RAM) 57, 59, 61
multidimensional arrays 56
multi-file projects 116
multiple 160, 162
channel timing issues 163, 166
communicating between clock domains 162, 172
multiple clocks 160, 162
multiplication................................ 104
-N option 196
NCF files 269, 281
not equal to................................. 106
object specifications....................... 245

octal 30
oe 276
offchip 274
on-chip RAMs 180, 210, 211, 212
operators12, 101, 102, 104, 106, 107, 108, 109
  arithmetic 104
  bit manipulation 101
  bitwise logical 108
  comparison 106
  concatenation 102
  conditional 109
  drop 102
  logical 107
  precedence 12
  relational 106, 107
  shift 102
  summary 12
  take 102
  trysema 92
  width 104
optimizing code............... 123, 154, 157
  examples 123, 154, 157
outfile.. 269
output 269, 286, 290, 294
  files 269
  standards 286, 290, 294
outtime............................... 269, 270
overflow 29
overview of Handel-C5, 12, 14, 15, 25, 29, 36, 75, 95, 111, 245, 309
padding 29, 38, 102
par...75, 76
parallel 5, 9, 44, 75, 151
  access to variables 151
  branch synchronization 6, 44, 164
  execution 75
  functions 119
  programs 5
  statements 75
  structure 9
parameterized macro expressions.... 131
parameters........................... 111, 278
  functions 111
  macros 111
  Verilog 278
paranoia 169, 275
PCI I/O standard 286, 289, 293
  33MHz 3.3V 286, 293
  33MHz 5.0V 286, 293
  66MHz 3.3V 286, 290
  PCI-X 286, 293
pin specifications.................... 264, 276
  omitting 276
pin_number attribute 264
pinouts 276
  specifying 276
pins............... 218, 231, 232, 264, 276
  constraining 264
  merging 231, 232
  naming 264
  reset 186
  specifying 276
  tri-state 232
pipelining 76, 157, 196, 207
  examples 207
PLD devices 179, 180, 183
pointers 39, 41, 42, 122
  addresses 41
  casting 17
  declaration 39
  operations 39
  to arrays 34
  to functions 122, 123
  to interfaces 42

port_in 218, 241
port_out 218, 241
porting C to Handel-C 32
ports 241, 277
  interfacing with external logic 241
  port names 241, 264
  specification 277
precedence 12
preprocessor 303, 305, 306
  concatenation 306
  conditional compilation 305
  error generation 307
  file inclusion 304
  line control 306
  line splicing 307
  macros 139, 303, 307
prialt 46, 78, 79, 89
  prialt examples 79
ProASIC devices..................... 180, 183
  constraints files 281
  I/O standards supported 289, 296
  pull up resistors 280
  RAMs 210
  slew rate on output buffer 286
proc 138, 139
program structure 9
properties 278
  specification 278
protecting critical code 62
prototypes 116, 118, 138
  functions 118
  macros 116, 138
pull 280
QDR devices................................ 193
qualifiers...................................... 29
Quartus 281
  assignments 281
quartus_proj_assign specification 281
RAM 57, 186, 187, 193
  Actel 210
  Altera 211
  arrays 56
  asynchronous 187, 192, 300
  block RAM 256
  different to arrays 54
  external 188, 190, 192
  foreign code 213
  initialization 54, 56
  multi-ported 57, 61
  off-chip 188, 190, 192, 274
  on-chip 180, 210, 211, 212
  overview 54
  restrictions 97
  simultaneous access 97
  synchronous 193, 195, 204, 261, 282
  syntax 320
  targeting 192, 204
  use of 186
  writing to 54
  Xilinx 212
range 31
rate specification.................... 270, 281
rclkpos................................. 195, 282
reading from external pins....... 219, 220
recursion........................ 111, 117, 133
recursive macros111, 133, 135, 136, 137
  shared expressions 135, 136, 137
reducing logic depth 154
reference books 4
register.. 68
registered reading from external pins220
relational operators 106
releasesema()................................ 93

replicated code 76
replicators.................................. 324
reset 90, 186, 297
 global 186
 specifying reset pin 186
restrictions.......... 46, 97, 117, 127, 136
 casting 97
 functions 117, 127, 129
 on channels 46
 on RAM and ROM 97
 on shared expressions 136
retime specification 284
return...................................84, 117
 types 117
right shift.................................... 102
ROM...............................54, 186, 212
 external 212
 LUT ROM in Altera devices 256
 overview 54
same rate external clock 189
sc_type specification...................... 285
scope 8, 29, 119
 variable sharing 8
Select Clock dialog 162
select operator............................. 131
selection within switch 88
sema...62
semaphores 62, 92, 93
seq 76
sequential and parallel execution.......75
sequential replication.......................76
set 33, 160, 183, 186
 clock 160
 family 183
 intwidth 33
 part 183
 reset 186
set clock 160
set family.................................... 183
set part....................................... 183
set reset..................................... 186
shared code..... 113, 116, 127, 135, 136
shared expressions................. 135, 136
 restrictions 136
shift operators
 shift operators 102
short 32
show specification 286
side effects.............................. 18, 95
sign extension 97, 102, 133
signals..................................63, 319
signed 31, 32, 107
signed/unsigned................. 32, 96, 107
 casting 96
simulations.................................. 228
 clock required 160
 file I/O 178
 simulating buses 230
 simulating interfaces 228
simulator 178
 input file format 178
 output 286
sizeof 17
sorts 218
 interfaces 218
Spartan devices 180, 183
 constraints files 281
 I/O standards supported 289, 296
 mprams 59
 on-chip RAMs 212
 RAM timing issues 256
 slew rate of output buffer 286
specifications......................... 245, 276
 base 254

bind 254
block 256
busformat 259
clk 261
clkpulselen 282
clock position 282
clockport 262
data 264
dci 265
extinst extlib extfunc 266
extpath 268
fifolength 268
infile and outfile 269
intime and outtime 269
minperiod 167, 273
object 245
offchip 274
paranoia 169, 275
pin 276
ports 277
properties 278
pull 280
quartus_proj_assign 281
rate 281
rclkpos 282
retime 284
sc_type 285
show 286
speed 286
standard 286, 290
std_logic_vector 295
strength 296
unconstrainedperiod 297
vhdl_type 298
warn 300
wclkpos 282
wegate 300
westart and welength 300
speed 286
SSRAM 193, 195, 261, 282
pipelined access 196, 198, 200
read and write cycles 195, 198
SSRAMs 193, 195, 196, 261
timing 195, 196, 198, 200, 282
SSTL I/O standard....286, 289, 293, 294
SSTL18 Class I 286, 294
SSTL18 Class II 286, 294
SSTL2 Class I 286, 293
SSTL2 Class II 286, 293
SSTL3 Class I 286, 294
SSTL3 Class II 286, 294
standard specification 286, 289, 290
AGP 286, 290
BLVDS 286, 290
CTT 286, 290
GTL 286, 294
HSTL 286, 291
LVCMOS 286, 291, 292
LVDCI 286, 292
LVDS 286, 292
LVPECL 286, 293
LVTTL 286, 293
PCI 286, 293
SSTL 286, 293, 294
standards.......................286, 290, 294
statements 10, 25, 75, 141, 147, 321
comparison with ANSI-C 25
compound 324
syntax 321
timing 141, 147
static..................................... 69, 73
initializing static variables 73
std_logic_vector specification 295
storage class specifiers.....................64

Stratix devices................180, 183, 209
constraints files 281
embedded memory 209
I/O standards supported 289, 296
mprams 59
pull-up resistors 280
RAMs 211
strength specification.................... 296
string constants 30
strings...................................30, 315
struct 36
structure member operator............ 109
structure pointer operator 109
structure pointers...........................42
structures 36, 42, 319
storage 36
syntax 319
subtraction................................. 104
summaries 10, 12, 14, 309
keywords 309
operators 12
statements 10
types 14
supported 32, 180
devices 180
types for porting 32
switch 88, 89
termination 89
synchronization................................6
synchronous RAMs193, 195, 204, 261, 282
clocks 193, 195, 196, 261
examples 196, 200, 205, 207
generating 282
read and write cycles 195, 198
timing 196, 198, 200, 282
syntax 309
take operator.............................. 102
targeting.................177, 179, 183, 186
FPGA/PLD devices 179, 183, 209
hardware 177
ports 243
RAM and ROM 54, 186
specific tools 243
Tcl files 269, 281
timing...........................141, 154, 233
asynchronous RAM 187
buses 233, 235
constraints 269, 270, 281
efficiency 154
examples 141, 152, 233
introduction 141
SSRAM 195, 282
statements 141, 147
TriMatrix memory......................... 209
tri-state ... 218, 221, 222, 224, 232, 294
buses 218, 221, 222, 224
interfaces 218, 294
pins 232
try ... reset...................................90
trysema()....................................92
ts..........................218, 221, 222, 224
type 14, 16, 22, 32, 72
clarifier 72
conversion 17, 96, 97
mapping for C and C++ 32, 65
names 321
operators 16, 68, 70, 71
qualifiers 71
summary 14
type clarifier <>............................72
typedef.......................................69
typeof...70
types.................14, 16, 22, 29, 31, 44

architectural 44
logic 31
overview 14, 29
types in C and Handel-C 16, 22
VHDL 295
unconstrainedperiod 297
undefined...................................... 33
undivided external clock 190
unions 21, 309
unsigned.......................... 31, 32, 107
values 29
overflow 29
variables................................73, 151
auto 64
default values 73
initialization 30, 64, 71, 73
local 64
parallel access 151
width of variables 17, 29
Verilog....................................... 278
instantiating components 254
parameters 278
VHDL... 295
generics 278
instantiating component 254
types 295
vhdl_type specification................... 298
Virtex devices 180, 183
constraints files 281
I/O standards supported 286, 296
mprams 59
on-chip RAM 212
RAM timing issues 256
slew rate of output buffer 286
specifying clock input 262
specifying DCI 265
Virtex-II Pro 180, 183
void................................ 39, 84, 117
volatile 71
warn specification......................... 300
wclkpos 195, 282
we 276
wegate 188, 190, 192, 300
welength......... 187, 188, 190, 192, 300
westart 187, 188, 190, 192, 300
while loops 85, 86
width..........................17, 29, 33, 104
adjustment 17, 102
inference 33
of variables 17, 29
operator 104
wires 63, 243
naming 243
with 245
WOM (write-only memory ports) . 57, 62
work library 254
write enable 187, 193, 300
asynchronous RAM 187, 300
positioning 300
synchronous RAM 193
write strobe 187
write-only memory 62
writing to external pins 221
Xilinx 180, 183, 212
bit mapping 59
block specification 256
devices 180, 183
on-chip RAM 212
ZBT-compatible devices 193